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ELECTROMAGNETISM

PHYSICS - 4

$$\textcircled{\star} \quad \vec{\nabla} \cdot \left( \frac{\hat{r}}{r^2} \right) = 4\pi \delta^3(\vec{r})$$

$\textcircled{\star}$  Avogadro's Number में Prime की Prime हैं, अभी खाला  
 याद नहीं होता :  $N_0 = 6.023 \times 10^{23}$

$$\textcircled{\star} \quad \zeta(4) = \left( \frac{\pi^4}{90} \right)$$

(Stefan Boltzmann)

$$\zeta\left(\frac{3}{2}\right) = 2.6$$

(Bose Einstein Condensate)

$$\zeta(x+1) \Gamma(x+1) = \int_0^{\infty} \frac{x^n}{e^x - 1} dx$$

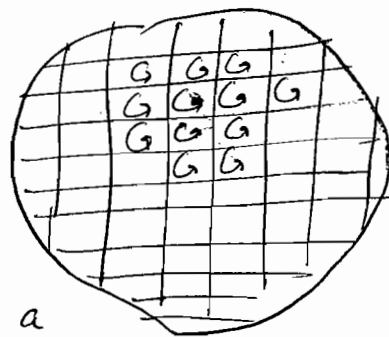
By def<sup>n</sup>

$$\zeta(x) = \sum_{n=1}^{\infty} \frac{1}{n^x}$$

$$\text{eg. } \zeta(4) = \frac{1}{1^4} + \frac{1}{2^4} + \frac{1}{3^4} + \frac{1}{4^4} + \dots$$

## Magnetic Shell

✓ Magnetic shell is a theoretical concept which can be regarded as the cause of magnetic field

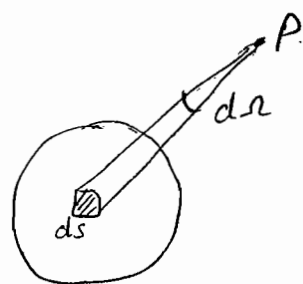


Magnetic shell is a thin sheet of magnetic material magnetized in such a way that magnetization is perpendicular to the surface of sheet everywhere.

It may be regarded as a large number of very short magnetic dipoles placed edge to edge with similar dipoles pointing in same direction.

Magnetic potential due to magnetic shell at any point P subtending solid angle  $d\Omega$  is

$$\phi = \frac{I}{4\pi} d\Omega = \frac{I}{4\pi} \left( \frac{d\vec{s} \cdot \vec{r}}{r^3} \right)$$



If the shell is divided into small elements of small area  $d\vec{s}$ , we say that each element corresponds to magnetic moment

$$\vec{m} = I d\vec{s} \quad \Rightarrow \quad I = \frac{|\vec{m}|}{|d\vec{s}|}$$

This is also called strength of the cell i.e. magnetic moment per unit area.

# ELECTRICITY & MAGNETISM (1)

\* 4Q :  $60 \times 4 = 240$

If all correct :  $0.7 \times 240 = 168$

Total : 336

16 lectures

\* 15 classes course

8 + 6 + 2  
statics dynamics current

\* 3 units :

Q6 { ① Electrostatics & Magnetostatics ⑥

{ ② Current Electricity ③

{ ③ EM Theory & Blackbody Radiation ⑥ - ⑦

Q7 ← Most scoring question

# (a) Electrostatics & Magnetostatics ⑥ Electrodynamics

- (i) Field & Potential due to Dipole, Dipole-Dipole interactions, Multipole expansion of Potential,
- (ii) Laplace & Poisson Equation & simple applications
- (iii) Method of electrical images
- (iv) Dielectric & Polarization
- (v) Boundary Value Problem of Conducting & Dielectric Sphere in Uniform Field
- (vi) Magnetostatics : Magnetized Sphere in Uniform Field  
Ferromagnetic Material & Hysteresis

## Prerequisites

Field  
Potential  
Energy

Electrostatics : stationary charges

Electro Magnetism : moving charge

# Field & Potential

$$\vec{F} = -\vec{\nabla} U$$

↑  
Potential Energy

$$\odot \vec{\nabla} U \cdot d\vec{r} = dU$$

$$\Delta U = -\int \vec{F} \cdot d\vec{r}$$

↑  
Change in Potential Energy

dividing by unit mass (or charge)

$$\vec{E} = -\vec{\nabla} V \leftarrow \text{Potential}$$

where  $\vec{E} = \frac{\vec{F}}{m}$  &  $V = \frac{U}{m}$

$$\Rightarrow \Delta V = -\int \vec{E} \cdot d\vec{r}$$

↑  
Change in Potential

or

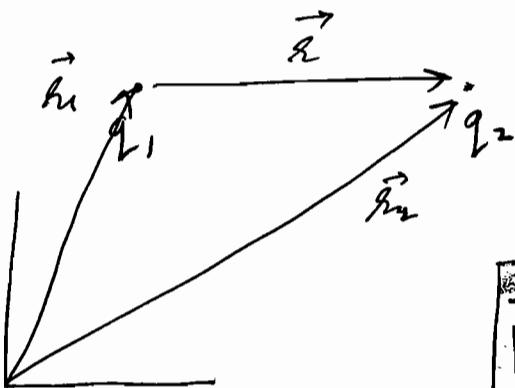
$$\vec{E} = \frac{\vec{F}}{q} \quad \& \quad V = \frac{U}{q}$$

if  $\vec{F}$  known  $\Rightarrow \vec{E}$  known  $\Rightarrow dV$  known  $\Rightarrow dU$  known

ε : Permittivity

## Coulomb's law

$$\vec{F} = \left( \frac{1}{4\pi\epsilon} \right) \frac{q_1 q_2}{r^2} \hat{r}$$



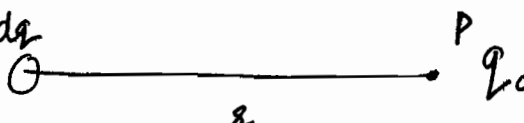
$$\vec{r} = \vec{r}_2 - \vec{r}_1$$

$\epsilon = \epsilon_0 k$   
 $k$  : dielectric const.  
 or  
 relative permittivity

$$\vec{F}_{\text{medium}} = \left( \frac{\vec{F}_{\text{air}}}{k} \right)$$

$$k \geq 1$$

$\Rightarrow E_{\text{medium}}$  reduced  $\Rightarrow \Delta V_{\text{medium}}$  reduced

Let us consider a system :- 

$$d\vec{F} = \frac{1}{4\pi\epsilon_0 k} \frac{dq q_0}{r^2} \hat{r} \quad (\text{from Coulumb's law})$$

$d\vec{E} = \frac{d\vec{F}}{q_0} =$  Field generated by  $dq$  ↑  
experimental law  
empirical observation

$$d\vec{E} = \frac{1}{4\pi\epsilon_0 k} \left(\frac{dq}{r^2}\right) \hat{r}$$

$$\vec{E} = \int d\vec{E} = \frac{1}{4\pi\epsilon_0 k} \int \frac{dq}{r^2} \hat{r}$$

This is how Coulumb law is used to find out  $\vec{E}$  at a point  $\vec{r}$ .

$$V = - \int \vec{E} \cdot d\vec{r}$$

$$U = qV$$

$dq =$	$\lambda dl$	(1-dimension)
	$\sigma da$	(2-dimension)
	$\rho dv$	(3-dimension)

Also note  $\frac{\hat{r}}{r^2} = \frac{\vec{r}}{r^3} = -\vec{\nabla} \left( \frac{1}{r} \right)$

$$d\vec{E} = \frac{1}{4\pi\epsilon_0 k} \frac{dq}{r^2} \hat{r} = -\vec{\nabla} \left( \frac{1}{4\pi\epsilon_0 k} \frac{dq}{r} \right)$$

Now  $d\vec{E} = -\vec{\nabla} V$

$$\Rightarrow \phi = \frac{1}{4\pi\epsilon_0} \left( \frac{dq}{r} \right)$$

→  $\phi$  is Potential provided it is Work Done per unit charge.

$$dW = \vec{F} \cdot d\vec{r} = q_0 \vec{E} \cdot d\vec{r}$$

$$\vec{E} \cdot d\vec{r} = \frac{dW}{q} = -\vec{\nabla} \phi \cdot d\vec{r}$$

$$= - \left( \frac{\partial \phi}{\partial x} \hat{i} + \frac{\partial \phi}{\partial y} \hat{j} + \frac{\partial \phi}{\partial z} \hat{k} \right) \cdot (dx \hat{i} + dy \hat{j} + dz \hat{k})$$

$$= -d\phi$$

$$\Rightarrow d\phi = -\vec{E} \cdot d\vec{r} = \left( \frac{dW}{q_0} \right) = dV$$

Hence  $\phi$  is Potential

$$\Rightarrow dV = \frac{1}{4\pi\epsilon_0} \left( \frac{dq}{r} \right)$$

$V_{ref}$  at  $V_{\infty} = 0$  [convention]

$$\int_0^V dV = \int \frac{1}{4\pi\epsilon_0} \frac{dq}{r} = - \int_{\infty}^r E \cdot da$$



# Gauss law

Flux of Electric Field through a closed surface is  $\left(\frac{1}{\epsilon_0}\right)$  times charge enclosed in the surface.

$$\text{Flux} = \oint_S \vec{E} \cdot d\vec{s} = \frac{1}{\epsilon_0} (q_{enc})$$

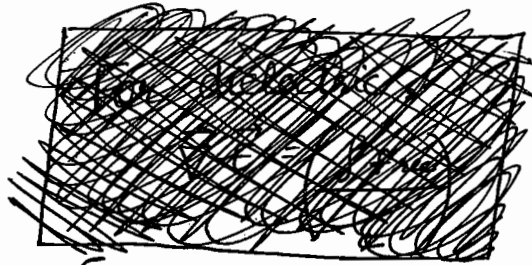
q<sub>enc</sub> is  
q<sub>total</sub> i.e.  
q<sub>free</sub> + q<sub>bound</sub>

From Fundamental law of divergence

$$\oint \vec{E} \cdot d\vec{s} = \int \vec{\nabla} \cdot \vec{E} \, dv = \frac{1}{\epsilon_0} \int \rho \, dv$$

$$\Rightarrow \int \left( \vec{\nabla} \cdot \vec{E} - \frac{\rho}{\epsilon_0} \right) dv = 0$$

$$\Rightarrow \vec{\nabla} \cdot \vec{E} = \frac{\rho}{\epsilon_0}$$



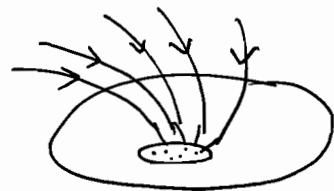
Note that the Gauss law is independent of medium

For source : divergence is +ve  $\star$   $\vec{\nabla} \cdot \vec{E}$  is a +ve thing !!

For sink : divergence is -ve



Positive divergence



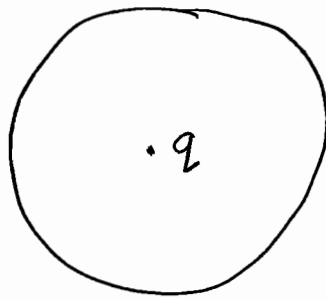
Negative divergence

→ The calculation of Gauss law is simpler when surface chosen is symmetric about E st.  $\oint E \cdot ds$  can be calculated easily. Such a closed symmetric surface is called GAUSSIAN SURFACE.

\* Note that in gravitation multiplier was  $(-4\pi G)$  while here the multiplier is  $(\frac{+1}{\epsilon_0})$

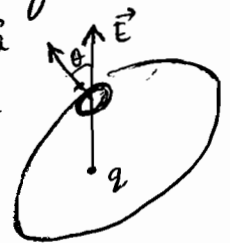
Proof

Consider a symmetric Gaussian surface around point charge  $q$ .



General Proof of Gauss law

Flux due to  $q$  at a differential element  $da$  of the gaussian surface



$$\text{Flux of } \vec{E} = \oint_S \vec{E} \cdot d\vec{s}$$

$$= \oint \frac{1}{4\pi\epsilon_0} \frac{q}{r^2} \cdot d\vec{s}$$

$$= \frac{1}{4\pi\epsilon_0} q \cdot \frac{1}{r^2} \cdot 4\pi r^2$$

$$= \left( \frac{q}{\epsilon_0} \right)$$

$$d\phi = \vec{E} \cdot d\vec{a}$$

$$= E da \cos\theta$$

$$= \frac{1}{4\pi\epsilon_0} \frac{q}{r^2} da \cos\theta$$

$$= \frac{q}{4\pi\epsilon_0} \cdot \left( \frac{da \cos\theta}{r^2} \right)$$

$$= \frac{q}{4\pi\epsilon_0} \cdot d\Omega$$

$$\Rightarrow \phi = \int d\phi = \frac{q}{4\pi\epsilon_0} \cdot 4\pi = \left( \frac{q}{\epsilon_0} \right)$$

⊙ Also  $\Omega$  subtended at an external point = 0  
 $\Rightarrow$  No contribution of external charges



⊙ There are 5 methods to find  $\vec{E}$  (that means all 4 variables of Electrostatics :-

- ① Coulumb's law
- ② Gauss law
- ③ Poisson & Laplace Equation
- ④ Method of Electrical Images

- ⑤ Graphical Method (not in course)

# Conductor

$$\rho = 0$$
$$q = 0$$

For Conducting surface: all charge resides on surface

⇒ no charge inside

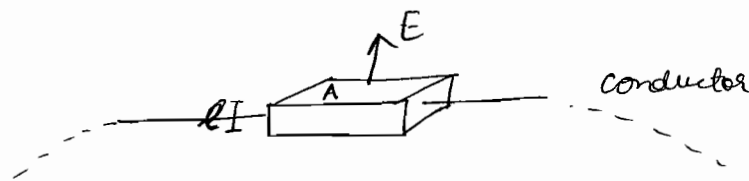
⇒  $\vec{E} = 0$  inside conductor

⇒  $V = \text{const.}$  Hence whole conductor is equipotential

Potential(inside) = Potential(surface)

$$dW = qdV = 0$$

$$\vec{E} = \left( \frac{\sigma}{\epsilon_0} \right) \hat{n}$$



Using Gauss law,

$$E \cdot A = \frac{\sigma \cdot A}{\epsilon_0}$$

$$\Rightarrow E = \left( \frac{\sigma}{\epsilon_0} \right)$$

$$\vec{E} = \left( \frac{\sigma}{\epsilon_0} \right) \hat{n}$$

General Proof of Gauss law:

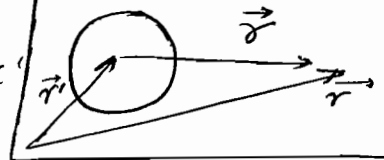
We know,  $E(\vec{r}) = \frac{1}{4\pi\epsilon_0} \int_{\text{all space}} \frac{\vec{\delta}}{\delta^2} \rho(\vec{r}') d\tau'$

where  $\vec{\delta} = (\vec{r} - \vec{r}')$

$$\Rightarrow \nabla \cdot E = \frac{1}{4\pi\epsilon_0} \int \nabla \cdot \left( \frac{\vec{\delta}}{\delta^2} \right) \rho(\vec{r}') d\tau'$$

Now  $\nabla \cdot \left( \frac{\vec{\delta}}{\delta^2} \right) = 4\pi \delta^3(\vec{\delta})$

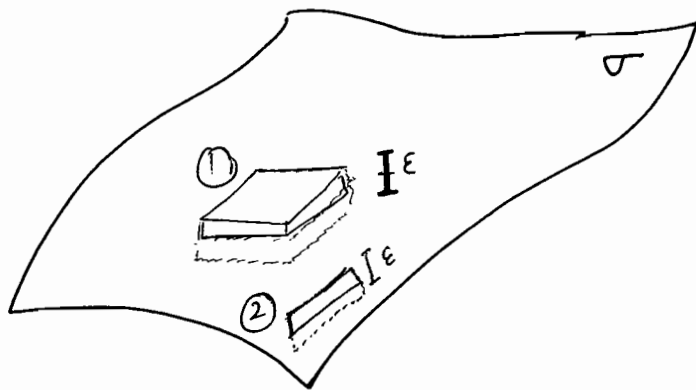
$$\Rightarrow \nabla \cdot E = \frac{1}{4\pi\epsilon_0} \int 4\pi \delta^3(\vec{\delta}) \rho(\vec{r}') d\tau' = \frac{1}{\epsilon_0} \int \delta^3(\vec{r} - \vec{r}') \rho(\vec{r}') d\tau'$$
$$= \frac{1}{\epsilon_0} \rho(\vec{r}) \quad \text{: Gauss law}$$



P-87  
Griffiths

$$\nabla^2 \left( \frac{1}{r} \right) = -4\pi \delta^3(\vec{r}) \quad \text{Correct}$$

(\*) Consider a charged boundary of surface charge  $\sigma$



Consider a match box as gaussian surface which is wafer thin, extending barely over the edge in each direction

From Gauss law,

$$\oint_S \vec{E} \cdot d\vec{a} = \frac{Q_{enc}}{\epsilon_0} = \frac{\sigma A}{\epsilon_0}$$

(if  $\sigma$  varies from point to point, choose a small  $A$ )  
 Now sides contribute nothing to flux as  $l \rightarrow 0$

$$\Rightarrow E_{above}^\perp - E_{below}^\perp = \left( \frac{\sigma}{\epsilon_0} \right) \quad \text{--- (1)}$$

$\Rightarrow$  The normal component of  $\vec{E}$  is discontinuous by an amount  $\left( \frac{\sigma}{\epsilon_0} \right)$  at any boundary. In particular, where there is no surface charge,  $E^\perp$  is continuous. eg. at the surface of a uniformly charged solid sphere.

The tangential component of  $\vec{E}$ , in contrast, is always continuous. Since  $\vec{E}$  is conservative,

$$\text{We know } \oint \vec{E} \cdot d\vec{l} = 0$$

Applying this to thin rectangular loop, where  $l \rightarrow 0$

$$E_{above}^\parallel = E_{below}^\parallel \quad \text{--- (2)}$$

Combining ① and ②,

$$\vec{E}_{\text{above}} - \vec{E}_{\text{below}} = \left( \frac{\sigma}{\epsilon_0} \right) \hat{n}$$

Meanwhile, the potential is continuous across any boundary since

$$V_{\text{above}} - V_{\text{below}} = - \int_a^b \vec{E} \cdot d\vec{l}$$

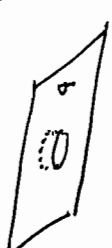
As the path shrinks to zero, so does the integral,

$$V_{\text{above}} = V_{\text{below}}$$

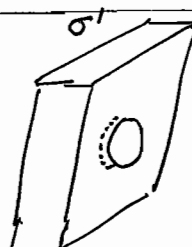
①  $\nabla \cdot \vec{E} = \left( \frac{\rho_{\text{enc}}}{\epsilon_0} \right)$  : general Gauss law. Valid even for dielectrics.

If you want concordance with Coulomb's law  $\vec{E} = \frac{1}{4\pi\epsilon_0 k} \cdot \frac{q}{r^2} \hat{r}$

We need to consider  $\rho_b = -\nabla \cdot \vec{P}$ . Then there will be no contradiction between the two.

②  conducting sheet

$$E = \frac{\sigma}{2\epsilon_0} \hat{n}$$



$$E = \left( \frac{\sigma'}{2\epsilon_0} \right)$$

But  $\sigma' = \frac{\sigma}{2}$

$$\Rightarrow E = \left( \frac{\sigma}{2\epsilon_0} \right) \hat{n}$$

How so ever thin the sheet is, we may consider it as a case of part II

Hence no contradiction.

Blackbody Radiation

1850 - 1900 : Classical Study

after 1900 : Quantum Study

Thermal Radiation typically lies in infrared region. As the body is heated more, the spectrum changes from infrared to blue light (more energy). (Concomitant with vibrational energy)

Blackbody Radiation can be regarded as either Thermodynamic system or Electromagnetic system (before 1900).  
Now we treat it as Quantum system.

Observed Fact that  $\nu \propto T$  or  $\lambda \propto \frac{1}{T}$

by observing colour of light and temperature of Blackbody.

Blackbody means any body which is at higher temperature than surrounding temperature. Mode of heat transfer is radiation. Note that every body can be blackbody while "Perfect Blackbody" is an ideal concept.

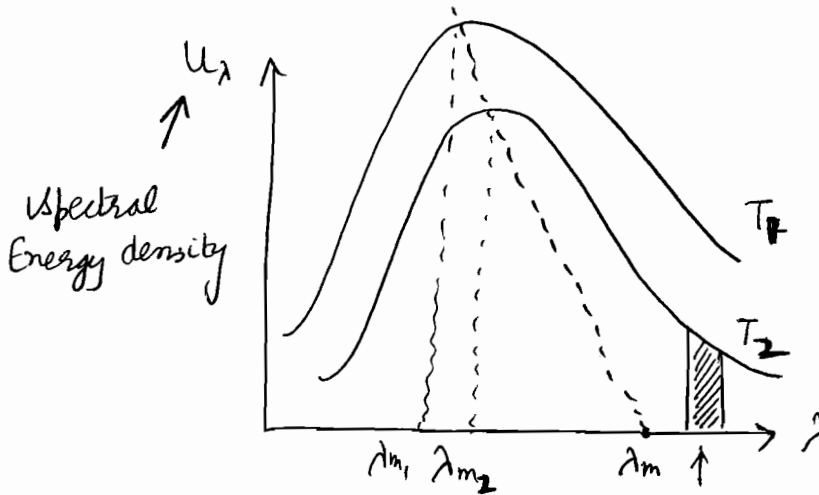
Wein observed the blackbody radiation and observed  $\lambda \propto \frac{1}{T}$  i.e.  $\lambda T = \text{const}$ . Weins displacement law

He considered it thermodynamically and gave Wein's distribution law. (system filled with adiabatically expanding assembly of particles)

Rayleigh - Jean observed it as an electromagnetic phenomenon.

Stephen Boltzmann came even before Wein's law. It says Energy radiated by blackbody [ $u \propto T^4$ ]  $T = \text{Temperature of blackbody}$

"Lummer & Pringsheim" heated various bodies and studied  $u_\lambda$  vs  $\lambda$ . Empirically they found:



$$T_1 > T_2$$

$$u = \int u_\lambda d\lambda$$

$$\text{Area} = u_\lambda d\lambda$$

= energy emitted between  $\lambda$  and  $(\lambda + d\lambda)$

Wein ←  
 lower wavelength side  
 or  
 higher frequency side

→ Rayleigh Jean  
 higher wavelength side  
 or  
 lower frequency side

$u_\lambda$  ■ : energy density

$u$  : energy =  $\int u_\lambda d\lambda$

Stephen Boltzmann's Formula

Area<sub>1</sub>  $\propto T_1^4$

Area<sub>2</sub>  $\propto T_2^4$

$$T_1 > T_2$$

$$\lambda_{m1} < \lambda_{m2}$$

$$\lambda_{m1} \propto \frac{1}{T}$$

Wein's Formula

→ From these empirical observations, all the 3 laws could be observed:

- ① Stephen Boltzmann law
- ② Wein's law
- ③ Rayleigh Jean law

Rayleigh Jeans law :  $u_{\lambda} d\lambda = \left( \frac{8\pi}{\lambda^4} d\lambda \right) kT$

We can also write  $u_{\nu} d\nu = \left( \frac{8\pi \nu^2}{c^3} d\nu \right) kT$  ✓  
Standing Waves' modes of vibration energy per mode

Wein's distribution law :  $u_{\lambda} d\lambda = \frac{A}{\lambda^5} f(\lambda T) d\lambda$

Stephen Boltzmann's law :  $\int u_{\lambda} d\lambda \propto T^4$

Both Rayleigh and Wein were not able to explain the curve. Both believed that E is continuous. Till now, no concept of Photon.

Now come Plank. He gave Planck's Blackbody Radiation Formula. He considered

$u_{\lambda} d\lambda = \left( \text{No. of modes of vibration in wavelength region } \lambda \text{ and } (\lambda+d\lambda) \text{ in unit volume} \right) \times \left( \text{Average Energy per mode} \right)$

Now we know,

$$g(\nu) d\nu = \frac{d^3 \nu d^3 p}{h^3} = \frac{V}{h^3} 4\pi p^2 dp = V \cdot \frac{4\pi \nu^2}{c^3} d\nu$$

For 2 modes of Polarization =  $V \cdot \frac{8\pi}{\lambda^4} d\lambda$

Plank took the formula from Rayleigh Jean for no. of mode and consider Avg. energy per mode =

$$\frac{\left( \frac{hc}{\lambda} \right)}{\left( e^{hc/\lambda T} - 1 \right)}$$

Planck's Radiation Formula

$$\Rightarrow u_{\lambda} d\lambda = \frac{8\pi hc}{\lambda^5} \cdot \frac{d\lambda}{e^{(hc/\lambda T)} - 1}$$

Plank's Blackbody Distribution Formula



We can derive easily from ~~Maxwell Boltzmann~~ distribution

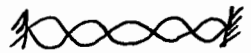
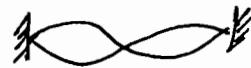
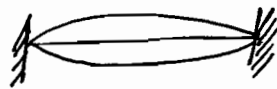
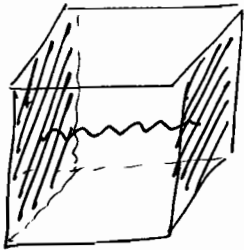
$$u = g(\nu) d\nu f(\epsilon) d\epsilon$$

$$\langle \epsilon \rangle = \frac{h\nu}{e^{h\nu/kT} - 1}$$

at  $T \uparrow$  :  $\langle \epsilon \rangle \rightarrow kT$

Hence classical blackbody radiation explained.

To find no. of Modes:



different modes of standing waves.

$$L_x = n_x \cdot \left(\frac{\lambda}{2}\right) = \left(\frac{n_x c}{2\nu}\right)$$

$$\nu = \left(\frac{n_x c}{2L_x}\right)$$

$$L_y = n_y \left(\frac{\lambda}{2}\right)$$

$$L_z = n_z \left(\frac{\lambda}{2}\right)$$

Let

$$L_x = L \cos \alpha = n_x \frac{\lambda}{2}$$

$$L_y = L \cos \beta = n_y \frac{\lambda}{2}$$

$$L_z = L \cos \gamma = n_z \frac{\lambda}{2}$$

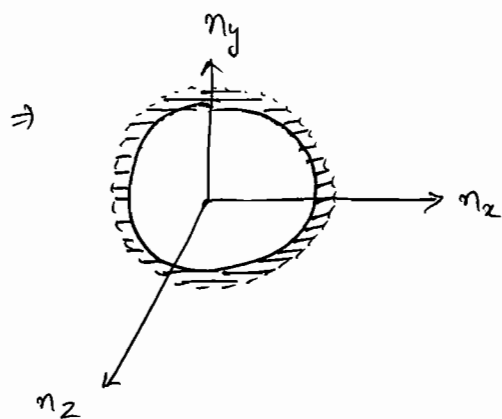
(What is  $L$ ...  
इसके हिस्से शरीर  
में करो)

We know,  $\cos^2 \alpha + \cos^2 \beta + \cos^2 \gamma = 1$

$$\frac{(n_x^2 + n_y^2 + n_z^2)}{L^2} \frac{\lambda^2}{4} = 1$$

$$\Rightarrow (n_x^2 + n_y^2 + n_z^2) = \left(\frac{2L}{\lambda}\right)^2$$

It is the equation of sphere.



\* Shaded region is the no. of modes between  $\lambda$  and  $(\lambda + d\lambda)$

$$\left[ (n_x + dn_x)^2 + (n_y + dn_y)^2 + (n_z + dn_z)^2 \right] = \left[ \frac{2L}{\lambda + d\lambda} \right]^2$$

No. of modes between wavelengths  $\lambda$  and  $(\lambda + d\lambda)$   
(Positive Quadrant only)

$$= \frac{1}{8} [4\pi R^2 dR]$$

$$= \frac{1}{8} \cdot 4\pi \frac{4L^2}{\lambda^2} \cdot \frac{2L}{\lambda^2} d\lambda$$

$$\boxed{R = \frac{2L}{\lambda}}$$

$$= V \cdot \frac{4\pi}{\lambda^4} d\lambda$$

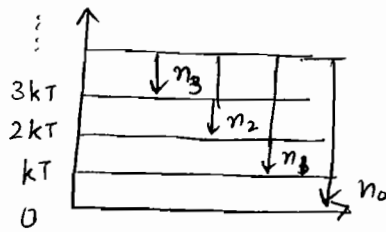
For longitudinal waves : Modes =  $V \cdot \frac{4\pi}{\lambda^4} d\lambda$

For transverse waves : Modes =  $V \cdot \frac{8\pi}{\lambda^4} d\lambda$

[Quantum Results just verifies these results]

Note that

$$\langle E \rangle = \frac{\sum \epsilon_i n_i}{\sum n_i}$$



Planck had only Maxwell Boltzmann statistics at his disposal.

Planck proposed  $E = h\nu$   
 i.e. discrete in nature. It was starting point of  
 Quantum Energy

$$n_i = A e^{-\beta \epsilon_i} \quad \leftarrow \text{Maxwell distribution}$$

$$\Rightarrow n_i = A e^{-\frac{h\nu}{kT}}$$

Put  $A = n_0$

$$\Rightarrow n_1 = n_0 e^{-\frac{h\nu}{kT}}$$

$$n_2 = n_0 e^{-\frac{2h\nu}{kT}}$$

$$\sum n_i = n_0 + n_0 e^{-\frac{h\nu}{kT}} + n_0 e^{-\frac{2h\nu}{kT}} + \dots$$

$$N = n_0 \left[ \frac{1}{1 - e^{-\frac{h\nu}{kT}}} \right]$$

$$\Rightarrow \langle E \rangle = \frac{\sum 0 \times n_0 + h\nu n_1 + 2h\nu n_2 + \dots}{\sum n_i} = \left( \frac{E}{N} \right)$$

$$E = \frac{\sum \left[ n_0 h\nu e^{-\frac{h\nu}{kT}} + 2n_0 h\nu e^{-\frac{2h\nu}{kT}} + \dots \right]}{n_0 h\nu e^{-\frac{h\nu}{kT}} \left[ 1 + 2e^{-\frac{h\nu}{kT}} + 3e^{-\frac{2h\nu}{kT}} + \dots \right]}$$

Its an APGP

$$\text{let } S = 1 + 2e^{-(h\nu/kT)} + 3e^{-(2h\nu/kT)} + \dots$$

$$e^{-h\nu/kT} S = e^{-(h\nu/kT)} + 2e^{-(2h\nu/kT)} + \dots$$

$$S[1 - e^{-h\nu/kT}] = 1 + e^{-h\nu/kT} + 2e^{-2h\nu/kT} + \dots$$

$$= \frac{1}{1 - e^{-h\nu/kT}}$$

$$\Rightarrow E = n_0 h\nu e^{-(h\nu/kT)} \cdot \frac{1}{[1 - e^{-(h\nu/kT)}]^2}$$

$$\langle E \rangle = \frac{E}{n_0 \left[ \frac{1}{1 - e^{-(h\nu/kT)}} \right]} = \frac{h\nu e^{-(h\nu/kT)}}{1 - e^{-(h\nu/kT)}}$$

$$\langle E \rangle = \frac{h\nu}{e^{h\nu/kT} - 1}$$

✓ ⊛ This is average energy per mode  
 $\langle E \rangle = E f(E) = h\nu \left( \frac{1}{e^{\frac{h\nu}{kT}} - 1} \right)$   
 Total energy =  $g(E) \cdot \langle E \rangle dE$

$$U_\nu d\nu = \frac{8\pi h \nu^3}{c^3} \frac{d\nu}{(e^{h\nu/kT} - 1)}$$

$$U_\lambda d\lambda = \frac{8\pi h c}{\lambda^5} \frac{d\lambda}{\left( e^{\frac{hc}{\lambda kT}} - 1 \right)}$$

Planck's  
 Radiation  
 Formulae

Perfect

We can derive all the 3 classical laws from here:

# Stefan Boltzmann Law

$$\int_0^{\infty} u_{\lambda} d\lambda = U = \frac{8\pi h}{c^3} \int_0^{\infty} \frac{\nu^3 d\nu}{e^{h\nu/kT} - 1}$$

Put  $\frac{h\nu}{kT} = x$

$$\nu = x \left( \frac{kT}{h} \right)$$

$$U = \frac{8\pi h}{c^3} \int_0^{\infty} \left( \frac{kT}{h} \right)^4 \frac{x^3 dx}{e^x - 1} \quad \nearrow \left( \frac{\pi^4}{15} \right)$$

$$\int_0^{\infty} \frac{x^3 dx}{e^x - 1} = \left( \frac{\pi^4}{90} \right) \times \Gamma(4)$$

$$= \frac{6\pi^4}{90}$$

$$= \left[ \frac{8\pi h}{c^3} \cdot \frac{6\pi^4}{90} \cdot \frac{k^4}{h^4} \right] T^4$$

$$U = \sigma T^4$$

U: Energy per unit volume

Hence derived.

S: Energy radiated per unit Area per unit time

$$\Rightarrow S = \frac{1}{4} U \cdot c$$

$\frac{1}{4}$ : Correction for falling normally to an area

$$S = \left[ \frac{1}{4} \cdot \frac{6 \cdot 8 \pi^5 k^4}{90 h^3 c^2} \right] T^4 = \sigma T^4$$

$$\sigma = \sigma T^4$$

$$k = 1.38 \times 10^{-23}$$

$$\sigma = 5.68 \times 10^{-8} \text{ SI units}$$

(Stefan Boltzmann's Constant)

Rayleigh Jean's law

(दीर्घ नाम, बर्तन  $\lambda$ )

$$U_\lambda d\lambda = \frac{8\pi hc}{\lambda^5} \frac{d\lambda}{e^{hc/\lambda kT} - 1}$$

In higher wavelength region,  $e^{\frac{hc}{\lambda kT}} = \left(1 + \frac{hc}{\lambda kT}\right)$

$$U_\lambda d\lambda = \frac{8\pi hc}{\lambda^5} \cdot \frac{\lambda kT}{hc} d\lambda = \frac{8\pi k}{\lambda^4} d\lambda$$

Hence Planck's law converges into Rayleigh-Jean's law at higher wavelength regions.

Wein's Distribution law

(दोटा नाम, दोटा  $\lambda$ )

For small  $\lambda$

$$U_\lambda d\lambda = \frac{8\pi hc}{\lambda^5} \frac{d\lambda}{e^{hc/\lambda kT}} = \frac{A}{\lambda^5} f(\lambda T) d\lambda$$

Hence Planck's law converges into Wein's law at lower wavelength regions.

## Wien's Displacement Formula

It says  $\lambda_m T = \text{const} = \underbrace{0.2896 \text{ cm K}}_{\substack{\uparrow \\ \text{Wien's Constant (b)}}$

$$\left. \frac{du_\lambda}{d\lambda} \right|_{\lambda_m} = 0$$

$$\left. \frac{d^2 u_\lambda}{d\lambda^2} \right|_{\lambda_m} < 0$$

$$u_\lambda = 8\pi hc \left[ \lambda^{-5} (e^{hc/\lambda kT} - 1)^{-1} \right]$$

$$-5\lambda^{-6} (e^{hc/\lambda kT} - 1)^{-1} + \lambda^{-5} - (e^{hc/\lambda kT} - 1)^{-2} e^{(hc/\lambda kT)} \left( \frac{hc}{kT} \right) \frac{1}{\lambda^2} = 0$$

$$\Rightarrow \left[ -5\lambda^{-6} + (e^{hc/\lambda kT} - 1)^{-1} e^{hc/\lambda kT} \left( \frac{hc}{kT} \right) \right] = 0$$

$$\Rightarrow 5 = \frac{\left( \frac{hc}{\lambda kT} \right) e^{hc/\lambda kT}}{e^{hc/\lambda kT} - 1}$$

Let  $\frac{hc}{\lambda kT} = x \Rightarrow$

$$\boxed{\frac{x e^x}{e^x - 1} = 5}$$

Transcendental Equation  
in  $x$

Solution:  $x \approx 4.95$

$$\frac{hc}{\lambda_m kT} = 4.95$$

$$\Rightarrow \boxed{\lambda_m T} = \frac{hc}{4.95 k} = b$$

$$= \underline{\underline{0.2896 \text{ cm K}}}$$

★ Temperatures of stars are known by studying spectra of stars. Wein's displacement law is used to study it.

○ Proof of  $\nabla \times \vec{E} = 0$

Let us prove it for the simplest configuration: a point charge

$$\vec{E} = \frac{1}{4\pi\epsilon_0} \frac{q}{r^2} \hat{r}$$

$$\Rightarrow \int_a^b \vec{E} \cdot d\vec{l} = \frac{1}{4\pi\epsilon_0} \int_a^b \frac{q}{r^2} \cdot dr = \frac{q}{4\pi\epsilon_0} \left[ \frac{1}{a} - \frac{1}{b} \right]$$

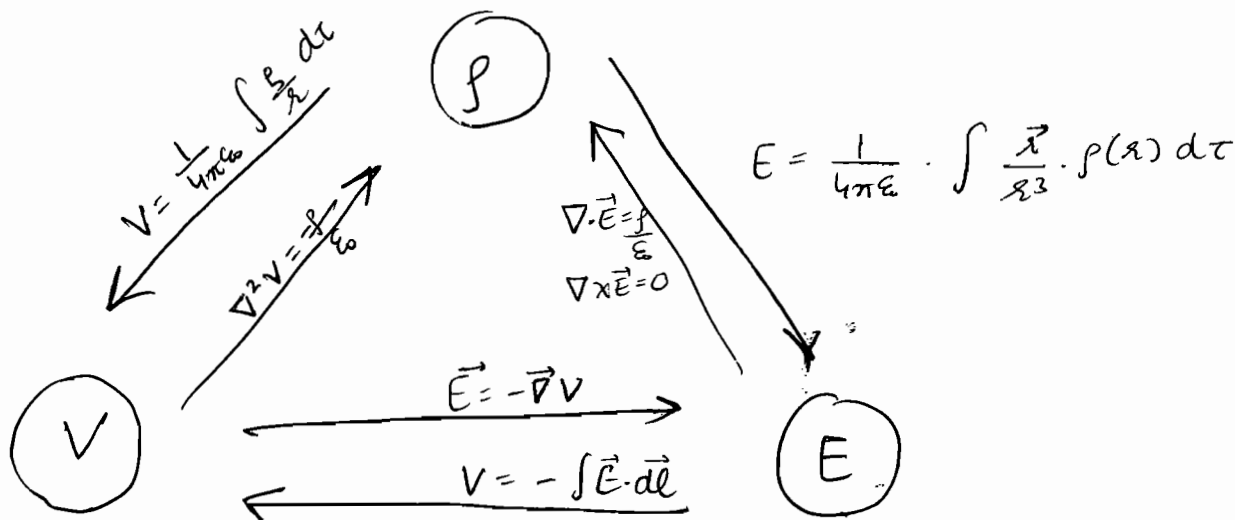
The integral around a closed path i.e.  $a=b$

$$= \int_a^a \vec{E} \cdot d\vec{l} = \oint \vec{E} \cdot d\vec{l} = \frac{q}{4\pi\epsilon_0} \left[ \frac{1}{a} - \frac{1}{a} \right] = 0$$

From Stokes theorem

$$\oint \vec{E} \cdot d\vec{l} = \int (\nabla \times \vec{E}) \cdot d\vec{A}$$

since R.H.S. = 0 for any surface  $\Rightarrow \boxed{\nabla \times \vec{E} = 0}$



★  $E_{\text{per unit volume}} = \frac{1}{2} \epsilon_0 |\vec{E}|^2$   
(energy)

This energy is stored in the electric field itself



# E&M (3)

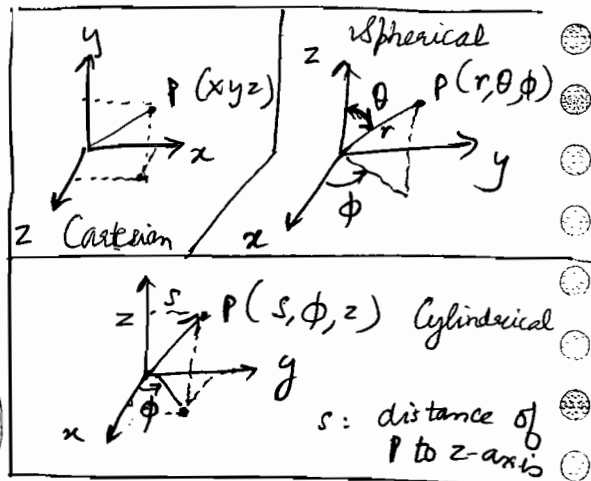
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LAPLACE & POISSON EQUATION

$$\nabla^2 V = -\frac{\rho}{\epsilon_0} \quad : \text{Poisson Eqn}$$

$$\nabla^2 V = 0 \quad \text{Laplace Eqn}$$

Called Laplace b'coz it has NOTHING else apart from a LAPLACE in LHS.



✓ We can apply  $[\nabla^2 V]$  either in rectangular (Cartesian) coordinates or spherical coordinates.

✓ Note that the 5 methods to calculate electrostatic problems can be used to solve any force-field problems. (G, E, strong, weak)

○ simple applications may include

(i)  $\rho$  is given,  $V$  is to be determined.

eg. (a)  $\underline{\underline{\rho = \rho_0 x^2}}$

$$\Rightarrow \frac{\partial^2 V}{\partial x^2} = -\left(\frac{\rho_0}{\epsilon_0}\right) x^2$$

since  $\rho = f(x) \Rightarrow V = f_1(x) \Rightarrow \frac{\partial^2 V}{\partial x^2} = \frac{d^2 V}{dx^2}$

$$\Rightarrow \frac{d^2 V}{dx^2} = -\left(\frac{\rho_0}{\epsilon_0}\right) x^2$$

$$\Rightarrow V = -\left[\frac{\rho_0}{\epsilon_0}\right] \left[ \frac{x^4}{12} + Ax + B \right]$$

★ Note that  $V = V(x)$  if  $\rho = \rho(x)$  is a correct one!!  
say if  $V = V(x, y)$ , then for sure some  $y$  term will come in  $\rho$ .

$$(b) \quad \underline{\underline{\rho = \rho_0 \left(1 - \frac{a^3}{r^3}\right)}}, \quad r \leq a$$

1 way: Put  $r = \sqrt{x^2 + y^2 + z^2}$  : Its very complicated !!!

2<sup>nd</sup> way:  $\rho = \rho(r)$   
 $\Rightarrow V = f_1(r)$

$$\nabla^2 \lambda = \frac{1}{r^2} \frac{\partial}{\partial r} \left( r^2 \frac{\partial \lambda}{\partial r} \right) + \frac{1}{r^2 \sin \theta} \frac{\partial}{\partial \theta} \left( \sin \theta \frac{\partial \lambda}{\partial \theta} \right)$$

SPHERICAL  
COORDINATES

$$+ \frac{1}{r^2 \sin^2 \theta} \left( \frac{\partial^2 \lambda}{\partial \phi^2} \right)$$

$$\frac{1}{r^2} \frac{\partial}{\partial r} \left( r^2 \frac{\partial V}{\partial r} \right) = - \left( \frac{\rho_0}{\epsilon_0} \right) \left[ 1 - \frac{a^3}{r^3} \right]$$

$$\Rightarrow \frac{\partial}{\partial r} \left[ r^2 \frac{dV}{dr} \right] = - \frac{\rho_0}{\epsilon_0} \left[ r^2 - \frac{a^3}{r} \right]$$

$$\Rightarrow r^2 \frac{dV}{dr} = - \frac{\rho_0}{\epsilon_0} \left[ \frac{r^3}{3} - \frac{a^3}{\cancel{r}} \ln r \right] + A$$

$$\Rightarrow \frac{dV}{dr} = - \frac{\rho_0}{\epsilon_0} \left[ \frac{r}{3} - \frac{a^3 \ln r}{r^2} \right] + \frac{A}{r^2}$$

$$2) \quad V = \int \frac{dV}{dr} = - \frac{\rho_0}{\epsilon_0} \left[ \frac{r^2}{6} - \left( \frac{a^3}{r^2} \ln r + \dots \right) \right] + \frac{A}{r} + B$$

↑  
IBP

(ii)  $V$  is given,  $\rho$  is to be determined.

eg.  $\rho = \underline{\underline{-\epsilon_0 \nabla^2 V}}$

11  $V = (x^2 + y^2 + z^2)$

$$\vec{E} = -\vec{\nabla}V = -[2x\hat{i} + 2y\hat{j} + 2z\hat{k}] \text{ Volt/meter}$$

$$\rho = -\epsilon_0 \nabla^2 V$$

$$= -\epsilon_0 (2+2+2) = -6\epsilon_0 \text{ Columb/m}^3$$

⊗ Note that  $\vec{E} = -\vec{\nabla}\phi$  is valid only for conservative field. Hence we have to mention, it is conservative as  $E = f(x, y, z)$  and not  $x, y, z$ .

19  $V = \underline{\underline{a - b(x^2 + y^2) - c \ln(x^2 + y^2)}}$

Note that it is conservative system

$$\Rightarrow E = -\vec{\nabla}V$$

$$\Rightarrow \nabla \cdot \vec{E} = -\nabla^2 V = \frac{\rho}{\epsilon_0} \Rightarrow \rho = -\epsilon_0 [-2b + \dots]$$

$$\Rightarrow \rho = -\epsilon_0 (\nabla^2 V)$$

iii) Verify the equation,

eg.  $V(r, \theta) = \underline{\underline{\frac{1}{4\pi\epsilon_0} \frac{\rho \cos\theta}{r^2}}}$

No derivatives involved, hence conservative....

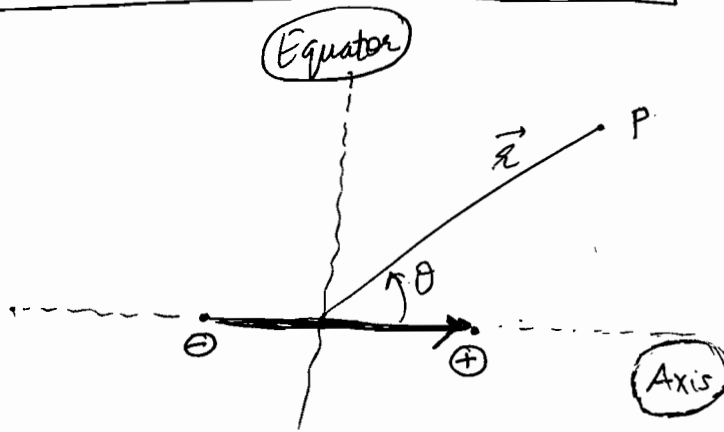
$$\vec{E} = -\vec{\nabla} V = -\frac{\partial V}{\partial r} \hat{r} + \frac{1}{r} \frac{\partial V}{\partial \theta} \hat{\theta} =$$

Spherical Coordinates

$$\vec{\nabla} \lambda = \left[ \frac{\partial \lambda}{\partial r} \hat{r} + \frac{1}{r} \frac{\partial \lambda}{\partial \theta} \hat{\theta} \right] + \frac{1}{r \sin \theta} \frac{\partial \lambda}{\partial \phi} \hat{\phi}$$

$$E_r = \frac{1}{4\pi\epsilon_0} \frac{2p \cos \theta}{r^3}$$

$$E_\theta = \frac{1}{r} \frac{\partial V}{\partial \theta} = \frac{p \sin \theta}{4\pi\epsilon_0 r^3}$$



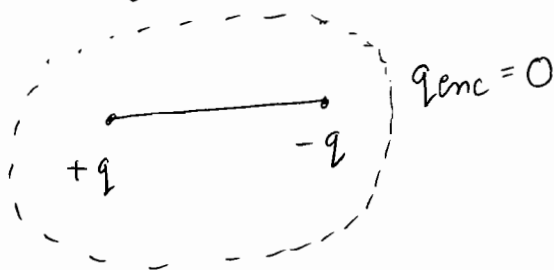
$$\vec{E} = \frac{1}{4\pi\epsilon_0} \frac{p}{r^3} (2 \cos \theta \hat{r} + \sin \theta \hat{\theta})$$

Spherical Coordinates

$$\vec{\nabla} \cdot \vec{A} = \left[ \frac{1}{r^2} \frac{\partial}{\partial r} (r^2 A_r) + \frac{1}{r \sin \theta} \frac{\partial}{\partial \theta} (\sin \theta A_\theta) + \frac{1}{r \sin \theta} \frac{\partial A_\phi}{\partial \phi} \right]$$

Now find  $\nabla^2 V = -\frac{\rho}{\epsilon_0}$

$\rho = 0 \Rightarrow \nabla^2 V$  should come out to be 0.



$$\nabla^2 V = \frac{1}{r^2} \left[ \frac{-2 \cos \theta}{r^2} \right] + \frac{1}{r \sin \theta} \frac{2 \sin \theta \cos \theta}{r^3}$$

$$= \frac{[-2 \cos \theta + 2 \cos \theta]}{r^4} = 0$$

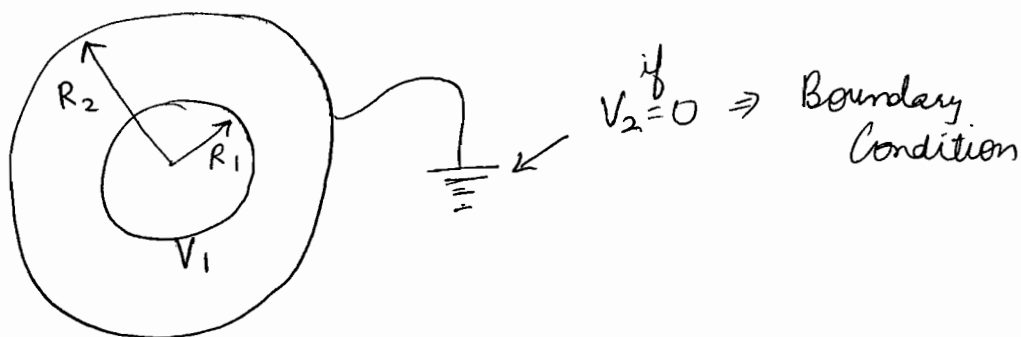
✓ The distance between charges should be of magnitude of atomic distances. Otherwise it is not a dipole.

$$\vec{E} = \frac{1}{4\pi\epsilon_0} \frac{p}{r^3} (2\cos\theta \hat{e}_r + \sin\theta \hat{\theta})$$

$$E(r, \theta) = |\vec{E}| = \frac{1}{4\pi\epsilon_0} \frac{p}{r^3} \sqrt{3\cos^2\theta + 1}$$

in language based problems.

eg. 2 ~~concentric~~ concentric spheres, Find out  $V$  in between  $R_1$  and  $R_2$ .



If non conducting  $\rho = \rho_0 \left(1 - \frac{a^3}{r^3}\right) \checkmark$

If conducting  $\rho = 0 \checkmark$

looking at symmetric geometry, we can safely conclude

$$V = f(r)$$

$$\Rightarrow \frac{d}{dr} \left( r^2 \frac{dV}{dr} \right) = 0$$

$$\Rightarrow r^2 \frac{dV}{dr} = A$$

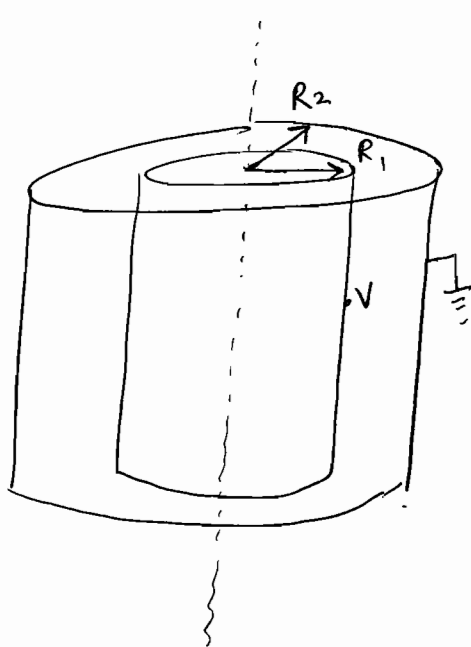
$$\Rightarrow \frac{dV}{dr} = \frac{A}{r^2} \Rightarrow V = -\frac{A}{r} + B$$

$$@ r = R_1, V = V_1 \quad \text{and} \quad @ r = R_2, V = 0$$

$$0 = -\frac{A}{R_2} + B \Rightarrow B = \frac{A}{R_2}$$

$$V_1 = -\frac{A}{R_1} + \frac{A}{R_2} \Rightarrow A = \frac{V_1}{\left(\frac{1}{R_2} - \frac{1}{R_1}\right)}$$

Hence =  $V = \frac{-A}{r} + B$



\* do not try to cross verify via Gauss law

$$V_1 = \frac{1}{4\pi\epsilon_0} \frac{Q_{enc}}{R_1}$$

$$\Rightarrow E \cdot 4\pi r^2 = \frac{Q_{enc}}{\epsilon_0}$$

$$\Rightarrow \int_r^{R_2} dV = -\int_r^{R_2} E \cdot dr$$

THIS IS WRONG AS 1<sup>st</sup> STEP IS WRONG.

all known variables.

$$V = \frac{4\pi\epsilon_0 Q_{enc}}{r}$$

for isolated sphere (no other charge like another shell in this case) and assuming E varies similarly from  $R_1$  to  $\infty$  as  $\left(\frac{1}{r^2}\right)$

$$R_1 < r < R_2 \quad \text{2nd step is correct}$$

$$E = \frac{Q_e}{4\pi\epsilon_0 r^2} \Rightarrow V(R_2) - V(r) = -\frac{Q_e}{4\pi\epsilon_0} \left(\frac{1}{r} - \frac{1}{R_2}\right)$$

$$\Rightarrow +V = \frac{Q_e}{4\pi\epsilon_0} \left(\frac{1}{r} - \frac{1}{R_2}\right)$$

$$\text{at } r=R_1, V=V_1 \Rightarrow Q_e = \frac{V_1 \cdot 4\pi\epsilon_0}{\left(\frac{1}{R_1} - \frac{1}{R_2}\right)}$$

$$\Rightarrow V = \frac{V_1}{\left(\frac{1}{R_1} - \frac{1}{R_2}\right)} \left(\frac{1}{r} - \frac{1}{R_2}\right)$$

Conducting cylinders  $\Rightarrow$  charge in between = 0

These concentric spheres, cylinders etc. form Capacitance. The Potentials in between are used to stored charge. Similarly 2 parallel plates are used to store charge.

Cylindrical

$$\nabla^2 \lambda(r, \theta, z) = \frac{1}{r} \frac{\partial}{\partial r} \left( r \frac{\partial \lambda}{\partial r} \right) + \frac{1}{r^2} \frac{\partial^2 \lambda}{\partial \theta^2} + \frac{\partial^2 \lambda}{\partial z^2}$$

Looking at geometry of problem,  $V = f(r)$

$$\Rightarrow \frac{1}{r} \frac{d}{dr} \left( r \frac{dV}{dr} \right) = 0$$

$$\Rightarrow r \frac{dV}{dr} = A \Rightarrow \frac{dV}{dr} = \frac{A}{r} \Rightarrow V = A \ln r + B$$

We can find out A and B via Boundary Conditions.

→ Can be verified from Gauss law assuming  $Q_{enc}$

@  $r_2: V = 0$

$$V = A \ln r_2 + B = 0 \Rightarrow B = -A \ln r_2$$

$$\Rightarrow V = A \ln \left( \frac{r}{r_2} \right)$$

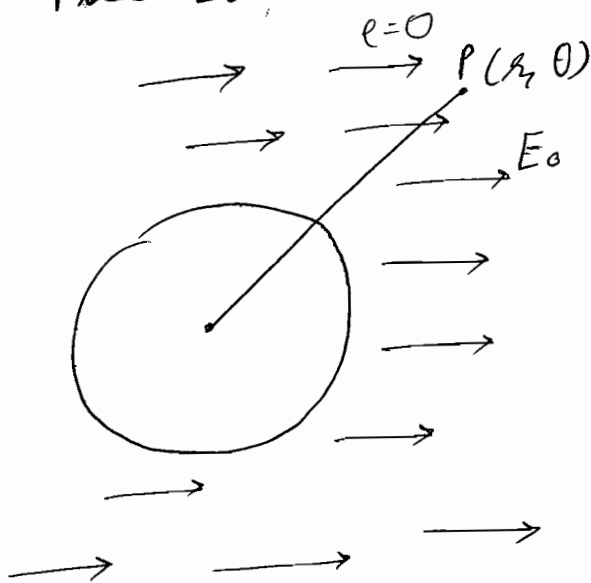
@  $r_1, V = V_1 \Rightarrow$

$$V = \frac{V_1}{\ln \left( \frac{r_2}{r_1} \right)} \ln \left( \frac{r_2}{r} \right)$$

⊛ Till now, we have dealt with simple problems of single dimension. Boundary Value Problems are difficult as they deal with multiple dimension, hence separation of variables are used!!

(v) Boundary Value Problem of Conducting Sphere in

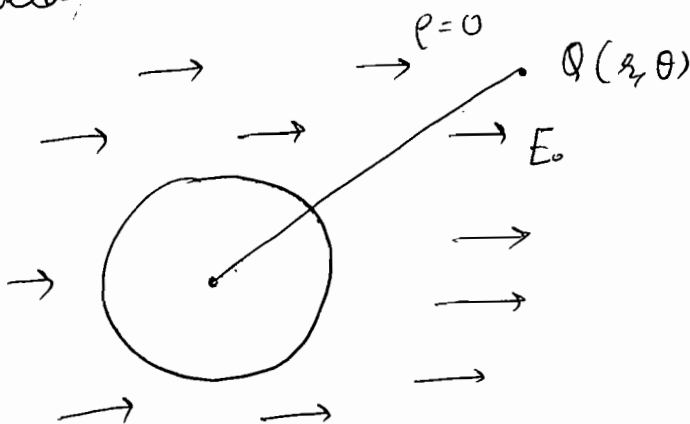
Uniform Field  $\vec{E}_0$



Find  $V_p$  and  $E_p$

(vi) Boundary Value Problem of Dielectric Sphere in

Uniform Field



Find  $V_Q$  and  $E_Q$ .

For conducting sphere placed in  $E_0$ .

$\rho_{\text{outside}} = 0$   
 $\rho_{\text{inside}} = 0$   
 $V_{\text{inside}} = V_{\text{surface}}$   
 $\nabla^2 V = 0$

$\rightarrow$  Before solving specifically for a conductor or dielectric, let us find the general solution of Laplace equation in spherical coordinates.  
 Specific solutions will be nothing but boundary values.

$\star V = f(r, \theta)$

Conducting sphere

$$\nabla^2 V = 0 \quad \frac{1}{r^2} \frac{\partial}{\partial r} \left( r^2 \frac{\partial V}{\partial r} \right) + \frac{1}{r^2 \sin \theta} \frac{\partial}{\partial \theta} \left( \sin \theta \frac{\partial V}{\partial \theta} \right) + \frac{1}{r^2 \sin^2 \theta} \frac{\partial^2 V}{\partial \phi^2} = 0$$

Separation of variables are used to solve such types of 3-d problems.

$V(r, \theta, \phi) = R(r) \Theta(\theta) \Phi(\phi)$  (say)

$$\frac{\partial^2 V}{\partial r^2} = \Theta \Phi \left( \frac{d^2 R}{dr^2} \right)$$

$$\frac{\partial^2 V}{\partial \theta^2} = R \Phi \left( \frac{d^2 \Theta}{d\theta^2} \right)$$

$$\frac{\partial^2 V}{\partial \phi^2} = \left( \frac{d^2 \Phi}{d\phi^2} \right) R \Theta$$

done on book: P-158

We are looking for solutions, that are products of functions, each of which depends on only one of the coordinates !!



Multiplying by  $r^2 \sin^2 \theta$ ,

$$\sin^2 \theta \frac{\partial}{\partial r} \left( r^2 \frac{\partial V}{\partial r} \right) + \sin \theta \frac{\partial}{\partial \theta} \left( \sin \theta \frac{\partial V}{\partial \theta} \right) + \frac{\partial^2 V}{\partial \phi^2} = 0$$

Separating the variables,

$$\Rightarrow \Theta \Phi \sin^2 \theta \frac{\partial}{\partial \theta} \left( \frac{\partial^2 \Phi}{\partial \phi^2} \frac{\partial R}{\partial r} \right) + \dots$$

Divide by  $R \Theta \Phi$ , we get

$$\Rightarrow \frac{1}{R} \sin^2 \theta \frac{d}{dr} \left( r^2 \frac{dR}{dr} \right) + \frac{1}{\Theta} \sin \theta \frac{d}{d\theta} \left( \sin \theta \frac{d\Theta}{d\theta} \right)$$

$$= -\frac{1}{\Phi} \frac{d^2 \Phi}{d\phi^2} = m^2 \text{ (say)}$$

it has to be a const. as LHS is not a function of  $\phi$

$$\Rightarrow \frac{d^2 \Phi}{d\phi^2} = -\Phi m^2$$

$$\Rightarrow \boxed{\Phi = e^{\pm i m \phi} = C_1 \sin m \phi + C_2 \cos m \phi}$$

dividing by  $\sin^2 \theta$ ,  $\star$  Assuming azimuthal symmetry; that  $v$  is not a function of  $\phi$  s.t.  $m=0$ .

$$\frac{1}{R} \frac{\cancel{\sin^2 \theta}}{\cancel{\sin^2 \theta}} \frac{d}{dr} \left( r^2 \frac{dR}{dr} \right) = -\frac{1}{\Theta} \frac{\sin \theta}{\sin^2 \theta} \frac{d}{d\theta} \left( \sin \theta \frac{d\Theta}{d\theta} \right)$$

Since both are equated  $\Rightarrow$  they are pure numbers

let us take const. =  $n(n+1)$  where  $n$  is an integer.

$$\frac{1}{R} \frac{d}{dr} \left( r^2 \frac{dR}{dr} \right) = n(n+1)$$

$$\Rightarrow \frac{d}{dr} \left( r^2 \frac{dR}{dr} \right) = + R n(n+1)$$

$$\Rightarrow r^2 \frac{d^2 R}{dr^2} + 2r \frac{dR}{dr} = nR(n+1)$$

let  $R = r^\alpha$  (or  $e^{\alpha r}$ )

$$\frac{dR}{dr} = \alpha r^{\alpha-1}$$

$$\frac{d^2 R}{dr^2} = \alpha(\alpha-1) r^{\alpha-2}$$

$$\Rightarrow r^2 \alpha(\alpha-1) r^{\alpha-2} + 2r \alpha r^{\alpha-1} = n r^\alpha (n+1)$$

$$\Rightarrow r^\alpha [ \alpha^2 + \alpha - n(n+1) ] = 0$$

$$\Rightarrow r^\alpha [ \alpha(\alpha+1) - n(n+1) ] = 0$$

$$\Rightarrow \boxed{\alpha = n} \quad \text{or} \quad \boxed{\alpha = -(n+1)}$$

$$\Rightarrow \text{[scribble]}$$

$$\Rightarrow \boxed{R = A_n r^n + B_n r^{-(n+1)}}$$

↑  
→ Here we cannot assume integer  $n \in \mathbb{R}^+$

→ From Power Series solution, we will get to know that  $n$  is an integer

that's why we took  $n(n+1)$

$$\begin{aligned} \alpha &= \frac{-1 \pm \sqrt{1^2 + 4n(n+1)}}{2} \\ &= \frac{-1 \pm \sqrt{4n^2 + 4n + 1}}{2} \\ &= \frac{-1 \pm (2n+1)}{2} \\ &= n, -(n+1) \end{aligned}$$

(★) Note that assumption of azimuthal symmetry is almost always valid wherever  $\vec{E}$  is taken along  $z$ -axis.

$$\frac{1}{\Theta} \frac{d}{d\Theta} \left( \sin\Theta \frac{d\Theta}{d\Theta} \right) = -n(n+1)$$

do not ignore this minus sign.

$$\Rightarrow \frac{1}{\sin\Theta} \frac{d}{d\Theta} \left( \sin\Theta \frac{d\Theta}{d\Theta} \right) + n\Theta(n+1) = 0$$

## Legendre's Differential Equation

Take  $\cos\Theta = x$

$$-\sin\Theta d\Theta = dx$$

note that कि अगर m लगे तो associate legendre आरगी और अगर m नही लगे तो legendre आरगी!!

$$\frac{d}{d\Theta} = \frac{d}{dx} \cdot \frac{dx}{d\Theta} = -\sin\Theta \frac{d}{dx}$$

$$\frac{1}{\sin\Theta} \frac{d}{d\Theta} = -\frac{d}{dx}$$

Both for both eigen value निकालने के लिए m नही लेना है!!

$$\frac{d}{dx} \left[ -\sin^2\Theta \frac{d\Theta}{dx} \right] + n(n+1)\Theta = 0$$

$$\Rightarrow \frac{d}{dx} \left[ (1-x^2) \frac{d\Theta}{dx} \right] + n(n+1)\Theta = 0$$

$$\Rightarrow (1-x^2) \frac{d^2\Theta}{dx^2} - 2x \frac{d\Theta}{dx} + n(n+1)\Theta = 0$$

solved by:  $\Theta_{(x)} = \sum_{k=0}^{\infty} a_k x^k$

Power series solution

solution:  $\Theta = \Theta_n(x) = P_n(x) = \frac{1}{2^n n!} \frac{d^n (x^2-1)^n}{dx^n}$  : Legendre's polynomials

$$\begin{aligned}
 P_0(x) &= 1 \\
 P_1(x) &= x \\
 P_2(x) &= \frac{3x^2 - 1}{2} \\
 &\vdots
 \end{aligned}
 \qquad
 \begin{aligned}
 P_0(x) &= 1 \\
 P_1(x) &= x \\
 P_2(x) &= \frac{3x^2 - 1}{2} \\
 &\vdots
 \end{aligned}$$

$$\begin{aligned}
 V_{R, \theta, \phi} &= R(\theta) \Phi(\phi) \\
 &= \sum_{n=0}^{\infty} \left( A_n r^n + \frac{B_n}{r^{n+1}} \right) P_n(\cos \theta) e^{\pm im\phi}
 \end{aligned}$$

Now we need to determine  $A_n$  and  $B_n$  using Boundary Conditions !!

★  $\vec{E}$  is a vector quantity having 3 components when  $V$  is a scalar quantity. Yet  $\vec{E} = -\vec{\nabla}V$   
 This is extraordinary because 1 scalar function contains information about all the 3 components. Answer is that components of  $E$  are not independent but connected by three equations:  $\vec{\nabla} \times \vec{E} = 0$

Therefore, in reality what Poisson equation does is to reduce a vector problem into 1 scalar problem in which there is no need to fuss with the components.

★ We usually use  $\infty$  as reference point. st.  $V_{\infty} = 0$  but this might not always work.

eg  $\left[ V(z) \text{ for a plane} = - \int_{\infty}^z \frac{1}{2\epsilon} \sigma \cdot dz = \frac{\sigma}{2\epsilon} (z - \infty) \right]$

Remedy is simply to choose any other point as reference point eg. we might choose the origin.

★ Only thing worth knowing for conductors is that  $\vec{E}_{\text{inside}} = 0$  and  $V_{\text{inside}} = V_{\text{surface}} = \text{const.}$

# Vector Calculus of Curvilinear Coordinates

HW.

(Appendix A : Griffith)

- $d\vec{l} = dx \hat{x} + dy \hat{y} + dz \hat{z}$  : Cartesian
- $d\vec{l} = dr \hat{r} + r d\theta \hat{\theta} + r \sin\theta d\phi \hat{\phi}$  : Spherical
- $d\vec{l} = ds \hat{s} + s d\phi \hat{\phi} + dz \hat{z}$  : Cylindrical

• Writing in terms of Coefficient

	$C_x$	$C_y$	$C_z$
Cartesian	1	1	1
spherical	1	$r$	$r \sin\theta$
cylindrical	1	$s$	1

such that for a general coordinate system,

$$\vec{dl} = C_x dx \hat{x} + C_y dy \hat{y} + C_z dz \hat{z}$$

## Gradient

We know  $dT = \vec{\nabla} T \cdot d\vec{l}$

$$= (\vec{\nabla} T)_x C_x dx + (\vec{\nabla} T)_y C_y dy + (\vec{\nabla} T)_z C_z dz$$

$$\Rightarrow \vec{\nabla} T = \frac{1}{C_x} \left( \frac{\partial T}{\partial x} \right) \hat{x} + \frac{1}{C_y} \left( \frac{\partial T}{\partial y} \right) \hat{y} + \frac{1}{C_z} \left( \frac{\partial T}{\partial z} \right) \hat{z}$$

✓ दिशा में  $E$  दिशा फैलाता है।

## Divergence

$$\vec{\nabla} \cdot \vec{A} = \frac{1}{C_x C_y C_z} \left[ \frac{\partial}{\partial x} (C_y C_z A_x) + \frac{\partial}{\partial y} (C_z C_x A_y) + \frac{\partial}{\partial z} (C_x C_y A_z) \right]$$

## Curl

$$\begin{aligned} \vec{\nabla} \times \vec{A} = & \frac{1}{C_y C_z} \left[ \frac{\partial}{\partial y} (C_z A_z) - \frac{\partial}{\partial z} (C_y A_y) \right] \hat{x} \\ & + \frac{1}{C_z C_x} \left[ \frac{\partial}{\partial z} (C_x A_x) - \frac{\partial}{\partial x} (C_z A_z) \right] \hat{y} \\ & + \frac{1}{C_x C_y} \left[ \frac{\partial}{\partial x} (C_y A_y) - \frac{\partial}{\partial y} (C_x A_x) \right] \hat{z} \end{aligned}$$

$$\Leftrightarrow \frac{1}{C_x C_y C_z} \begin{vmatrix} C_x \hat{x} & C_y \hat{y} & C_z \hat{z} \\ \frac{\partial}{\partial x} & \frac{\partial}{\partial y} & \frac{\partial}{\partial z} \\ C_x A_x & C_y A_y & C_z A_z \end{vmatrix}$$

⊛ 1<sup>st</sup> और 3<sup>rd</sup> row  
 में coefficients  
 multiplied हैं और  
 $\frac{1}{C_x C_y C_z}$  to of course after

Laplacian  $\Leftrightarrow \vec{\nabla} \cdot \vec{\nabla}$

## Laplacian

$$\nabla^2 T = \vec{\nabla} \cdot \vec{\nabla} T = \frac{1}{C_x C_y C_z} \left[ \frac{\partial}{\partial x} \left( \frac{C_y C_z}{C_x} \left( \frac{\partial T}{\partial x} \right) \right) + \frac{\partial}{\partial y} \left( \frac{C_z C_x}{C_y} \left( \frac{\partial T}{\partial y} \right) \right) + \frac{\partial}{\partial z} \left( \frac{C_x C_y}{C_z} \left( \frac{\partial T}{\partial z} \right) \right) \right]$$

→ H-atom problem

# E&M (4)

17/07/2012

o For conducting sphere placed in Electric Field,

$$V_{E, \theta} = \sum_{n=0}^{\infty} \left( A_n r^n + \frac{B_n}{r^{n+1}} \right) P_n(\cos \theta)$$

⊛ THESE ARE STANDARD RESULTS OF SOLUTION OF A LAPLACE EQUATION IN SPHERICAL COORDINATES.

where  $P_n(x) = \frac{1}{2^n n!} \frac{d^n}{dx^n} (x^2-1)^n$

Also  $\int_{-1}^1 P_m(x) P_n(x) dx = 0$  for  $m \neq n$

o Legendre Polynomials are orthogonal i.e. Cross product = 0 \*inner

$$\int_{-1}^1 P_m(x) P_n(x) dx = \frac{2}{2n+1} \text{ for } m=n$$

Hence Orthogonal Polynomials

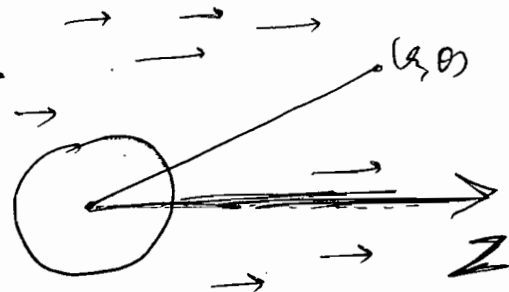
Q/24 & 26 :- same derivation for separation of variables

Boundary Value Problem of Conducting Sphere in Uniform Field (Assume field in z-axis)

(i) at  $r=R$ ,  $V$  is equipotential i.e.  $V = \text{const.} = 0$  [by suitable choice of scale]

(ii) at  $r \gg R$ ,  $V = - \int_0^r E_0 dx$   
 Potential @  $0 = 0$   
 As complete volume inside sphere is equipotential  
 $\Rightarrow E \cdot r \cos \theta$   
 Potential drops along direction of  $E$   
 taking effect of  $E_0$  and not conductor

ONLY 2 CONDITIONS



$$V = \left( A_0 + \frac{B_0}{r} \right) + \left( A_1 r + \frac{B_1}{r^2} \right) \cos \theta + \dots$$

From (ii)

looking @ Boundary Value,  $V = -E_0 r \cos \theta$   
 hence retaining only 2 terms, because 3 and above will have  $\cos^2 \theta$  dependence!!

$$\Rightarrow V = \left( A_0 + \frac{B_0}{r} \right) + \left( A_1 r + \frac{B_1}{r^2} \right) \cos \theta$$

@  $r \gg R$   $V = A_0 + A_1 r \cos \theta = -E_0 r \cos \theta$

$$\begin{cases} A_1 = -E_0 \\ A_0 = 0 \end{cases}$$

From (i)

$$V_R = 0 = \frac{B_0}{R} + \left( -E_0 R + \frac{B_1}{R^2} \right) \cos \theta$$

Comparing coefficients, (Note that we can vary  $\theta$  and it will still be 0)

$$B_0 = 0$$

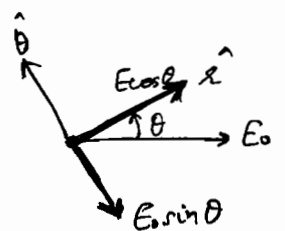
$$\frac{B_1}{R^2} - E_0 R = 0 \Rightarrow B_1 = +E_0 R^3$$

$$\Rightarrow V(r, \theta) = -E_0 r \cos \theta + \frac{E_0 R^3}{r^2} \cos \theta$$

$$\vec{E} = E_x \hat{x} + E_\theta \hat{\theta}$$

$$E_x = -\left( \frac{\partial V}{\partial x} \right) = E_0 \cos \theta + \frac{2E_0 R^3 \cos \theta}{r^3}$$

$$E_\theta = -\frac{1}{r} \left( \frac{\partial V}{\partial \theta} \right) = E_0 \sin \theta + \frac{E_0 R^3 \sin \theta}{r^3}$$



$$E_0 R^3 \propto \frac{\rho}{4\pi\epsilon_0}$$

Original field =  $E_0 \cos \theta \hat{x} - E_0 \sin \theta \hat{\theta}$

Note that here we are not looking for any other possibilities. We just want 1 function that satisfied B.C. Uniqueness Theorem will ensure that there are no other solutions.



$$\text{Modified } \vec{E} = \frac{E_0 R^3}{r^3} [2 \cos \theta \hat{x} + \sin \theta \hat{\theta}]$$

We can also take modified potential and find out  $\vec{E}_{\text{modified}}$

Hence it behaves like a dipole.

⊗ Note that modified field falls off as  $\left(\frac{1}{r^3}\right)$ , hence at  $r \gg R$ , only effect of  $E_0$ .

Surface Charge Density  $\sigma$

We know @ surface of conductor,

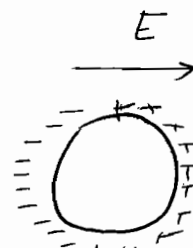
$$\vec{E} = \frac{\sigma}{\epsilon_0} \hat{n}$$

For sphere  $\vec{E} = \frac{\sigma}{\epsilon_0} \hat{x} = \vec{E}_r$

$$\frac{\sigma}{\epsilon_0} = \frac{E \cos \theta + E_0 \cos \theta}{2 \cos \theta} \frac{R^3}{R^3} = \underline{3 E_0 \cos \theta}$$

$$\Rightarrow \sigma = 3 \epsilon_0 E_0 \cos \theta$$

Upper surface:  $\oplus$  ve charge  
Lower surface:  $\ominus$  ve charge



$$Q = \iint \sigma dA = \iint \sigma r^2 \sin \theta d\phi d\theta$$

$Q$  should come out to be zero because it behaves like a dipole. Also it is induced charge, net charge conserved.

hence  $Q = 0$ .

○ At  $r = R$ , Note that  $E_\theta = 0$  ✓  
 $E_x = 3 E_0 \cos \theta$  ✓

○ For conductor,  $V_{\text{in}} = V_{\text{surface}}$  ✓

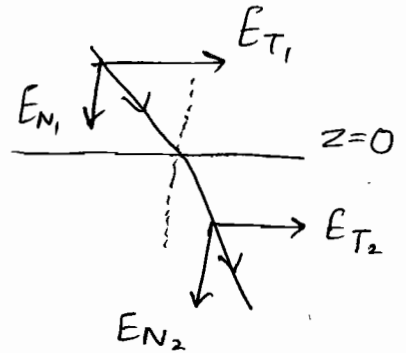
NOTE THAT CONDUCTOR के अंदर का ANALYSIS नहीं करना !!

Boundary value problem with dielectric sphere placed in  $\vec{E}$

## Dielectric Sphere

### Boundary Conditions

At interface of 2 dielectric media, tangential vector of  $\vec{E}$  remains conserved.



①  $E_{t1} = E_{t2}$  [ No Force acting hence Tangential Component remains conserved ]

&

②  $\epsilon_1 E_{n1} = \epsilon_2 E_{n2}$   
 i.e.  $k_1 \epsilon_0 E_{n1} = k_2 \epsilon_0 E_{n2}$   
 $\Rightarrow k_1 E_{n1} = k_2 E_{n2}$

$D_{n1} = D_{n2}$

→ Due to Polarization in dielectrics we do not know  $q_{enc}$  due to bound charge  $\therefore$  cannot apply normal Gauss law

→ Applying Gauss law for dielectrics

$\vec{\nabla} \cdot \vec{D} = \rho_f \Rightarrow \oint \vec{D} \cdot d\vec{s} = q_f$

$\Rightarrow D_1 \cdot A - D_2 \cdot A = 0 \Rightarrow D_1 = D_2$

$k_1 \left( \frac{\partial V}{\partial r} \right)_1 = k_2 \left( \frac{\partial V}{\partial r} \right)_2$

let  $V_{in} = V_{r < R}$   
 $V_o = V_{r > R}$

• DO not worry about bound charges & Laplace Equation  
 $\vec{E} = -\vec{\nabla} V \Rightarrow \vec{D} = -\epsilon \vec{\nabla} V$  (linear)  
 $\nabla \cdot \vec{D} = \rho_f \Rightarrow -\epsilon \nabla^2 V = \rho_f = 0$   
 $\Rightarrow \nabla^2 V = 0$

③  $\nabla^2 V_{in} = 0$

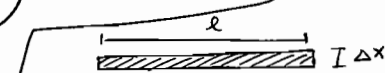
$\Rightarrow V_{in} = \sum \left( A_n r^n + \frac{B_n}{r^{n+1}} \right) P_n(\cos\theta)$

④  $\nabla^2 V_{out} = 0$

$V_o = \sum \left( A_n r^n + \frac{B_n}{r^{n+1}} \right) P_n(\cos\theta)$

⑤ @  $r=R$   $V_i = V_o$

⑥ @  $r=R$   $k_1 E_{n1} = k_2 E_{n2}$



$\Delta x \rightarrow 0$   
 $\vec{\nabla} \times \vec{E} = 0$   
 $\Rightarrow \int \vec{E} \cdot d\vec{l} = 0$   
 $\Rightarrow E_{T1} l + E_{n1} \Delta x - E_{T2} l - E_{n2} \Delta x = 0$   
 $\Rightarrow E_{T1} l - E_{T2} l = 0$   
 $\Rightarrow E_{T1} = E_{T2}$

Proof of  $E_{T1} = E_{T2}$

$$\text{i.e. } k_1 \left( \frac{\partial V_i}{\partial r} \right) \Big|_{r=R} = k_2 \left( \frac{\partial V_o}{\partial r} \right) \Big|_{r=R}$$

$$(7) \quad r \gg R \quad V_o = -E_o r \cos \theta$$

$$(8) \quad r=0 \quad V_{in} = \text{finite}$$

$$\Rightarrow B_n' \text{ are } 0$$

$$\Rightarrow V_{in} = \sum \left( A_n' r^n + \frac{B_n'}{r^{n+1}} \right) P_n(\cos \theta)$$

$$\Rightarrow V_{in} = \sum A_n' r^n P_n(\cos \theta)$$

$n$  will go upto 0 and 1

$$\Rightarrow V_{in} = A_0' P_0(\cos \theta) + A_1' r P_1(\cos \theta)$$

$$\Rightarrow V_{in} = A_0' + A_1' r \cos \theta \quad \checkmark$$

$$\text{Choosing } V_{o0} = A_0' = 0$$

$$\Rightarrow V_{in} = A_1' r \cos \theta$$

Now

$$V_o = \sum \left( A_n r^n + \frac{B_n}{r^{n+1}} \right) P_n(\cos \theta)$$

$$\begin{aligned} \text{At } r \gg R \quad -E_o r \cos \theta &= \sum A_n r^n P_n(\cos \theta) + \frac{B_0}{r} + \frac{B_1}{r^2} \cos \theta \\ &= \left( A_0 + A_1 r \cos \theta + \frac{B_0}{r} + \frac{B_1}{r^2} \cos \theta \right) \quad \left[ \text{only dependence till } \cos \theta \right] \end{aligned}$$

Using (1)

$$\Rightarrow A_0 = 0 \quad (\text{comparing coefficients})$$

$$\Rightarrow A_1 = -E_o$$

ONLY 4  
CONDITIONS  
(5, 6, 7, 8)

Using (8)

Note that basic expressions are same. Only difference is due to different boundary conditions & hence different constants.

$$\Rightarrow V_{out} = -E_0 r \cos\theta + \frac{B_0}{r} + \frac{B_1}{r^2} \cos\theta$$

Using (5) at  $r=R$

$$A_0' + A_1' r \cos\theta = -E_0 r \cos\theta + \frac{B_0}{r} + \frac{B_1}{r^2} \cos\theta$$

Comparing Coefficients

$$\Rightarrow B_0 = 0 \quad (\text{since } A_0' = 0)$$

@  $r=R$   $V_{in} = V_{out}$  ✓  
&  
 $E_{t1} = E_{t2}$   
give same equation!!

$$A_1' = -E_0 + \frac{B_1}{R^3}$$

Using (6) For radial coordinates, for a sphere  $\left(\frac{\partial V}{\partial n}\right) = \left(\frac{\partial V}{\partial r}\right)$

$$\text{@ } r=R \quad k_1 \left(\frac{\partial V_{in}}{\partial r}\right) = k_2 \left(\frac{\partial V_{out}}{\partial r}\right)$$

Always remember this step....

$$\Rightarrow k_1 A_1' \cos\theta = k_2 \left( -E_0 r \cos\theta + 2 \frac{B_1}{r^3} \cos\theta \right)$$

$$\Rightarrow k_1 A_1' = -k_2 E_0 - \frac{2k_2 B_1}{R^3}$$

Solving the 2 eqns,

we get  $V, E,$  induced charge

$$-k_1 E_0 + \frac{k_1 B_1}{R^3} = -k_2 E_0 - \frac{2k_2 B_1}{R^3}$$

$$\Rightarrow B_1 = \frac{(k_1 - k_2) E_0 R^3}{(k_1 + 2k_2)}$$

$$A_1' = -E_0 + E_0 \frac{(k_1 - k_2)}{(k_1 + 2k_2)}$$

$$A_1' = \frac{-3k_2 E_0}{(k_1 + 2k_2)}$$

$$\Rightarrow V_i = \frac{-3k_2 E_0 R \cos \theta}{(k_1 + 2k_2)}$$

$$V_o = -E_0 R \cos \theta + \frac{(k_1 - k_2) E_0 R^3 \cos \theta}{(k_1 + 2k_2) r^2}$$

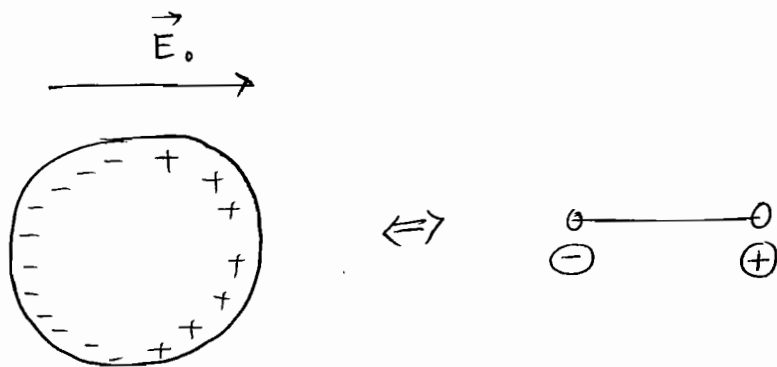
Dielectric sphere is behaving like a dipole.

$$\text{Modified Potential } V_m(\text{outside}) = \frac{(k_1 - k_2) E_0 R^3 \cos \theta}{(k_1 + 2k_2) r^2}$$

Taking analogy with dipole potential

$$\text{Hence } \frac{p}{4\pi\epsilon_0} = \frac{(k_1 - k_2) E_0 R^3}{(k_1 + 2k_2)}$$

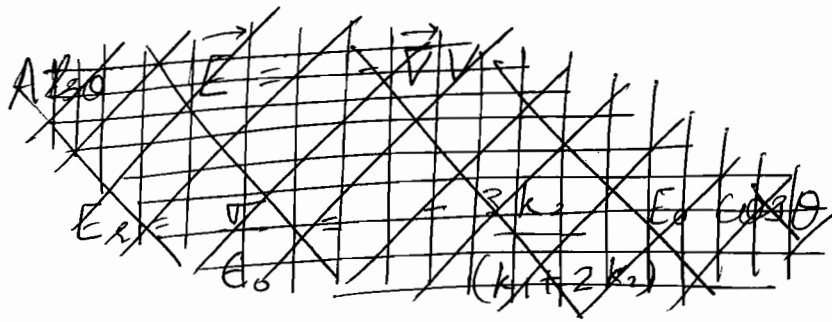
$$\Rightarrow p_{\text{equivalent}} = \frac{4\pi\epsilon_0 (k_1 - k_2) E_0 R^3}{(k_1 + 2k_2)}$$



Using volume as  $\frac{4}{3}\pi R^3$  and using Polarization as dipole moment per unit volume,

$$\vec{P} = \frac{p}{\frac{4}{3}\pi R^3} = 3\epsilon_0 E_0 \frac{(k_1 - k_2)}{(k_1 + 2k_2)}$$

⊙ Due to displacement of charge inside dielectric, caused due to applied external field, causes Polarization.



✓ For Cylindrical Coordinates, solving Laplace Equation:

$$V = V(r, \phi, z)$$

$$\nabla^2 V = 0$$

$$\frac{1}{r} \frac{\partial}{\partial r} \left( r \frac{\partial V}{\partial r} \right) + \frac{1}{r^2} \frac{\partial^2 V}{\partial \phi^2} + \left( \frac{\partial^2 V}{\partial z^2} \right) = 0 \quad \checkmark$$

$$V(r, \phi, z) = R(r) \Phi(\phi) Z(z)$$

$$\nabla^2 V = 0 \quad \text{and dividing by } R \Phi Z$$

$$\Rightarrow \frac{1}{R} \frac{d}{dr} \left( r \frac{dR}{dr} \right) + \frac{1}{\Phi r^2} \left( \frac{d^2 \Phi}{d\phi^2} \right) + \frac{d^2 Z}{dz^2} = 0$$

$$\frac{d^2 Z}{dz^2} + m^2 Z = 0$$

$$\Rightarrow \boxed{Z = e^{\pm imz}}$$

Making  $m=0$ , for 2 coordinates i.e. assuming cylindrical symmetry and no dependence on  $z$

$$\frac{1}{R} \frac{d}{dr} \left[ r \frac{dR}{dr} \right] = - \frac{1}{\Phi} \frac{d^2 \Phi}{d\phi^2} = \underline{\underline{n^2}} \quad (\text{say})$$

$$\frac{-1}{\Phi} \frac{d^2 \Phi}{d\phi^2} = n^2 \Rightarrow$$

$$\Phi = e^{\pm i n \phi}$$

$$\Phi(\phi) = \sum A_n \cos n\phi + B_n \sin n\phi$$

$$r \frac{d}{dr} \left( r \frac{dR}{dr} \right) = n^2 R$$

take  $R = r^\alpha$

$$r^2 \frac{d^2 R}{dr^2} + r \frac{dR}{dr} = n^2 R$$

$$\Rightarrow r^2 \alpha(\alpha-1) r^{\alpha-2} + r \alpha r^{\alpha-1} = n^2 r^\alpha$$

$$\Rightarrow \alpha(\alpha-1) + \alpha = n^2$$

$$\Rightarrow \underline{\alpha = \pm n}$$

$$R(r) = \sum_{n=0}^{\infty} C_n r^n + D_n r^{-n}$$

Its not taking into account for  $n=0$

Handling  $n=0$  case separately ....

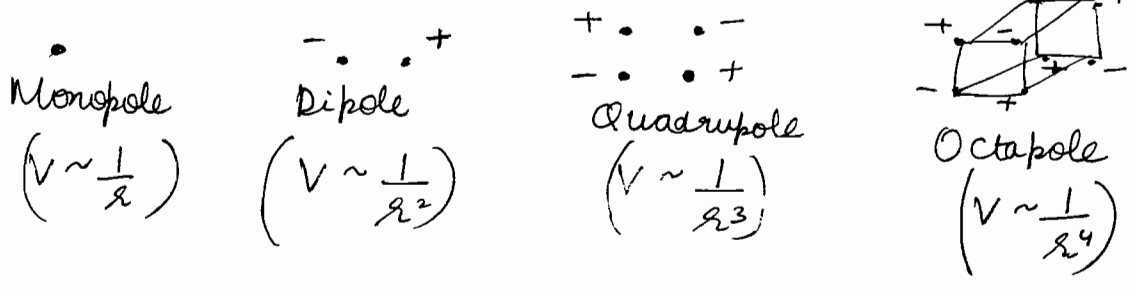
$$\Rightarrow \underline{r \frac{d}{dr} \left( r \frac{dR}{dr} \right) = 0}$$

[Put this form as 0 and note the opened form.  $\frac{d}{dr}$  में टिप्पणी खराब होगा]

$$\Rightarrow r \frac{dR}{dr} = A \Rightarrow R = A \ln r + B$$

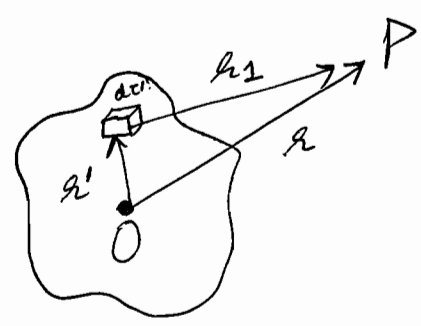
Hence  $R(r) = A \ln r + B + \sum_{n=1}^{\infty} (C_n r^n + D_n r^{-n})$  ✓

At large distances, potentials fall off as



From Multipole Expansion,

$$V(r) = \frac{1}{4\pi\epsilon_0} \sum_{n=0}^{\infty} \frac{1}{r^{n+1}} \int (r')^n P_n(\cos\theta') \rho(\vec{r}') d\tau$$



For monopole, placing charge at Origin O,

$$V_{\text{monopole}}(r) = \frac{1}{4\pi\epsilon_0} \cdot \frac{Q}{r}$$

For a point charge at the origin,  $V_{\text{monopole}}(r)$  represents the exact potential everywhere, not merely a first approximation at a large  $r$ .

(in this case, potential due to higher terms vanish)  
 If point charge is not placed at origin, then  $V(r)$  consists of other terms also.  
 But note that for a dipole, the higher terms do not vanish but become progressively smaller. So the formula

$$V_{\text{dipole}}(r) = \frac{1}{4\pi\epsilon_0} \frac{1}{r^2} \int r' \rho(r') d\tau' \cos\theta' = \frac{1}{4\pi\epsilon_0} \frac{p \cos\theta}{r^2}$$

(refer PTO PTO)

represents only on the 1<sup>st</sup> non zero term & is not exact, therefore.



Of course, as we go further and further away from the dipole, the  $V_{\text{dipole}}(r)$  becomes better and better approximation.

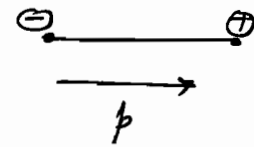
By the same token, at a fixed  $r$ , the dipole approximation improves as you shrink the separation  $d$ .

To construct a dipole whose potential is exactly given by  $V_{\text{dipole}}(r)$ , we will have to let  $d \rightarrow 0$ .

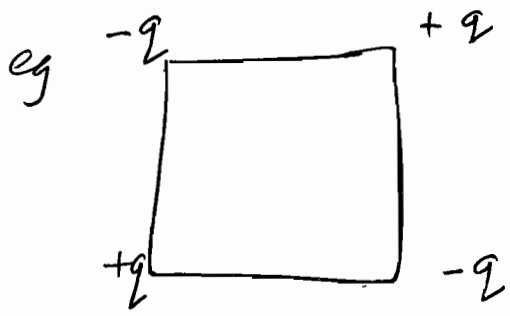
Correspondingly to conserve  $p$ , we will need to let  $q \rightarrow \infty$  i.e. A physical dipole becomes a pure dipole in the artificial limit  $q \rightarrow \infty$  and  $d \rightarrow 0$ .

BUT FOR ALL PRACTICAL PURPOSE, WE USE  $V_{\text{dipole}}(r)$  FOR PHYSICAL DIPOLES WITH SMALL BUT FINITE separation  $d$ .

✓ **Dipoles Add like vectors.**



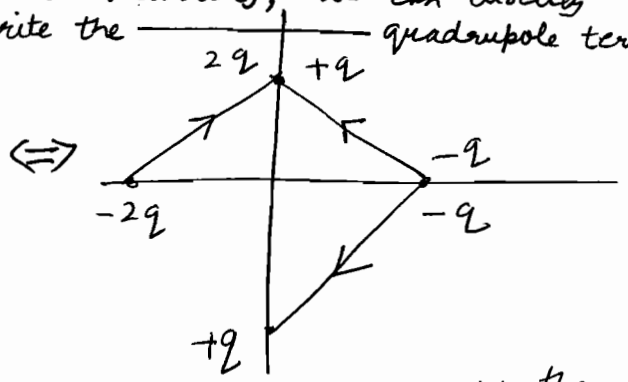
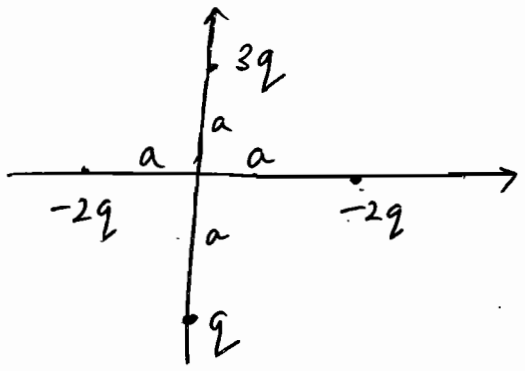
$$\vec{p}_{\text{total}} = \sum_i \vec{p}_i$$



$$\vec{p}_{\text{total}} = \vec{p}_1 + \vec{p}_2 = 0$$

→ This is a quadrupole and potential is dominated by quadrupole term in multipole expansion.

So, if all dipoles add up to zero vectorially, we can directly write the quadrupole term.

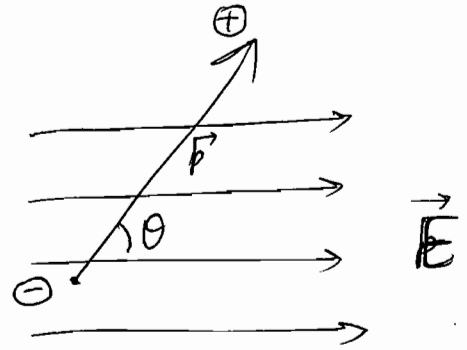


We can now add them vectorially.

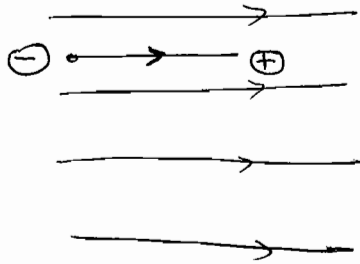
★ Multipole Expansion depends drastically on choice of origin.

# Potential Energy for a dipole

We can see  $\tau = \vec{p} \times \vec{E}$

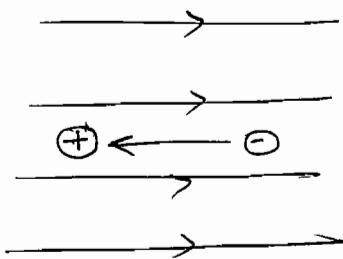


Now consider



We can see that it is stable equilibrium  
 $\Rightarrow U_{\min}$  at  $\theta = 0$

Now consider



We can see that it is in unstable equilibrium  
 $\Rightarrow U_{\max}$  at  $\theta = 180^\circ$

$$\text{Also } U_{\max} - U_{\min} = \int_0^\pi \tau_{\text{ext}} d\theta = + pE \int_0^\pi \sin\theta d\theta$$

$$= \underline{\underline{2pE}}$$

$$\tau = -pE \sin\theta$$

$$\tau_{\text{ext}} = pE \sin\theta$$

$$\Rightarrow dW = pE \sin\theta d\theta$$

$$\Rightarrow W = -pE \cos\theta \Big|_0^\pi$$

$$= pE \cdot 2$$

$$= \underline{\underline{2pE}}$$

Choose  $U_{\max} = pE$   
 $U_{\min} = -pE$

Now  $U_f \rightarrow (-pE) = \int_0^\theta pE \sin\theta d\theta$   
 $= pE \cdot (1 - \cos\theta)$

$\Rightarrow U_f = -pE \cos\theta$   
 $\Rightarrow \boxed{U(\theta) = -\vec{p} \cdot \vec{E}}$

★ Torque on a dipole in non uniform field :

$$\mathbf{N} = \vec{p} \times \vec{E} + \vec{r} \times \vec{F}$$

where  $\vec{F} = (\vec{p} \cdot \nabla) \vec{E}$

★ Also from Multipole Expansion, the dipole term is:

$$V_{\text{dipole}} = \frac{1}{4\pi\epsilon_0} \frac{1}{r^2} \int r' \cos\theta' \rho(\vec{r}') d\tau'$$

Now  $r' \cos\theta' = \vec{r}' \cdot \hat{r}$

$$\Rightarrow V_{\text{dipole}} = \frac{1}{4\pi\epsilon_0} \frac{\hat{r} \cdot \int \vec{r}' \rho(\vec{r}') d\tau'}{r^2}$$

We define dipole moments as  $\vec{p} = \int \vec{r}' \rho(\vec{r}') d\tau'$

$$\Rightarrow V_{\text{dipole}} = \frac{1}{4\pi\epsilon_0} \frac{\vec{p} \cdot \hat{r}}{r^2}$$

lets note :

- ① A point charge placed at Origin constitutes a "pure" monopole. If its not at Origin, then other terms of multipole expansion will also appear.
- ② The dipole moment  $\vec{p}$  does not change when we shift the origin only if net charge is zero. Otherwise it depends upon choice of origin

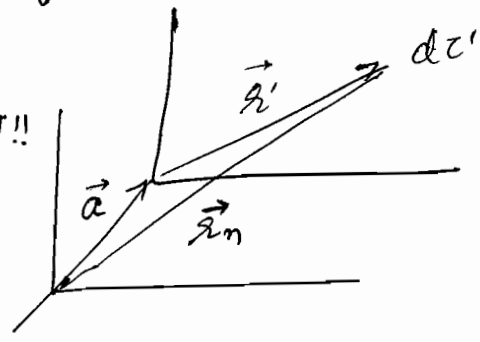
$$\vec{p}_{\text{new}} = \int \vec{r}'_m \rho(\vec{r}') d\tau'$$

*dq तो same ही रहेगा !!*

$$= \int (\vec{r}' + \vec{a}) \rho(\vec{r}') d\tau'$$

$$= \int \vec{r}' \rho(\vec{r}') d\tau' + \vec{a} \int \rho(\vec{r}') d\tau'$$

$$= \underline{\underline{\vec{p}}} + \vec{a} Q$$



For cylindrical coordinates, solving Laplace gives :-

$$V(r, \phi, z) = \left[ (A \ln r + B) + \sum_{n=1}^{\infty} (C_n r^n + D_n r^{-n}) \right] \left[ \sum A_n \cos n\phi + B_n \sin n\phi \right] e^{\pm imz}$$

## Dipole

Single charge is called Monopole (1)

$$V(r) = \frac{1}{4\pi\epsilon_0} \left( \frac{q}{r} \right) \quad l=0$$

If there are 2 charges forming a dipole (2)

$$V(r, \theta) = \frac{1}{4\pi\epsilon_0} \frac{p \cos \theta}{r^2} \quad l=1$$

Quadrupole (4)  $l=2$

Octapole (8)  $l=3$

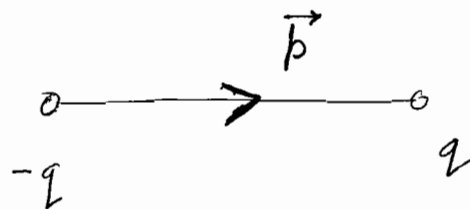
Hexapole (16)  $l=4$

In general  $2^l$  poles in Multipole system.

- If no  $\vec{E}_{\text{external}}$ , all multipoles are arranged randomly.
- When  $\vec{E}_{\text{ext}}$  is applied, charges are displaced and multipoles are formed.

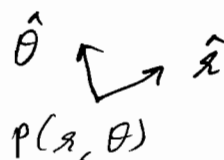
# Dipole

⊙ Equal & opposite charges separated by atomic dimensions.

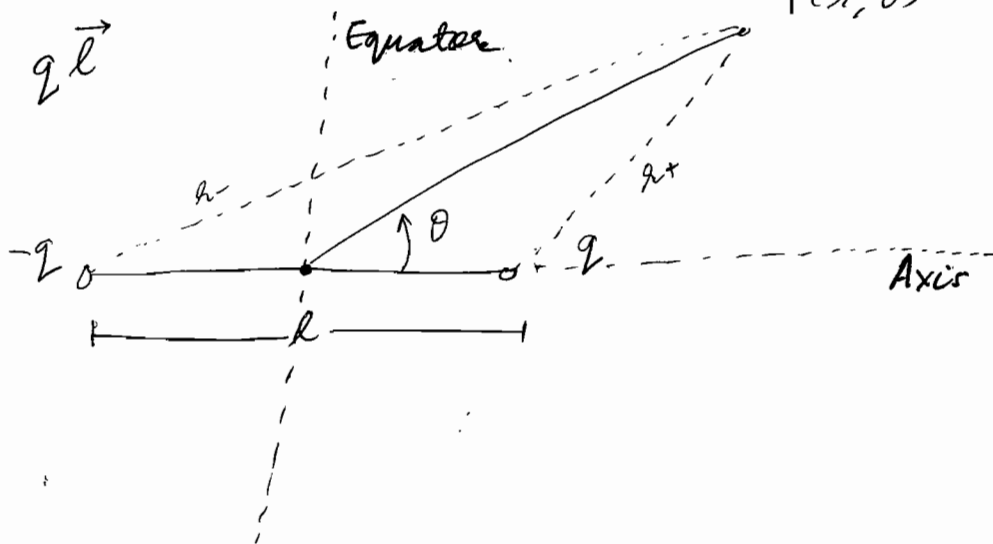


⊙  $\vec{p}$  from  $-q$  to  $+q$ .

$$\vec{p} = q\vec{l}$$



⊙ let us consider



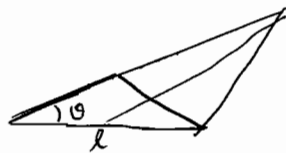
Note that  $l \ll r \Rightarrow r_+ \approx r_-$

$$V(r, \theta) = \frac{1}{4\pi\epsilon_0} \left[ \frac{q}{r_+} - \frac{q}{r_-} \right] = \frac{q}{4\pi\epsilon_0} \left[ \frac{\theta_- - \theta_+}{r_+ r_-} \right]$$

Approximating  $\angle Pq_-q_+ = \theta$

$$\checkmark (r_-) - (r_+) \approx l \cos \theta$$

$$\checkmark r_- r_+ \approx r^2$$



$$\Rightarrow V(r, \theta) = \frac{q}{4\pi\epsilon_0} \frac{l \cos \theta}{r^2} = \frac{1}{4\pi\epsilon_0} \left( \frac{p \cos \theta}{r^2} \right) \checkmark$$

⊙ (Note that in diffraction, we used to measure  $(90^\circ - \theta)$ .  $\therefore \sin \theta$  there)

$$\circ \vec{E} = -\vec{\nabla}V$$

$$= - \left[ \frac{\partial V}{\partial r} \hat{r} + \frac{1}{r} \frac{\partial V}{\partial \theta} \hat{\theta} \right]$$

$$= + \frac{2p \cos \theta}{4\pi\epsilon_0 r^3} \hat{r} + \frac{p \sin \theta}{4\pi\epsilon_0 r^3} \hat{\theta}$$

$$\boxed{\vec{E} = \frac{p}{4\pi\epsilon_0 r^3} [2 \cos \theta \hat{r} + \sin \theta \hat{\theta}]}$$

o Note that at equatorial point,  $V = 0$

Note that  $V(r, \theta) = \frac{1}{4\pi\epsilon_0} \frac{p \cos \theta}{r^2} = \frac{1}{4\pi\epsilon_0} \frac{\vec{p} \cdot \vec{r}}{r^3}$

$$\vec{E} = -\frac{1}{4\pi\epsilon_0} \vec{\nabla} \left( \frac{\vec{p} \cdot \vec{r}}{r^3} \right)$$

Take 1 scalar as  $\vec{p} \cdot \vec{r}$   
other scalar as  $\left(\frac{1}{r^3}\right)$

We know  $\boxed{\vec{\nabla}(fg) = (\vec{\nabla}f)g + f(\vec{\nabla}g)}$

$$\Rightarrow \vec{E} = -\frac{1}{4\pi\epsilon_0} \left[ \frac{1}{r^3} \vec{\nabla}(\vec{p} \cdot \vec{r}) + \vec{p} \cdot \vec{r} \vec{\nabla}\left(\frac{1}{r^3}\right) \right]$$

$$= -\frac{1}{4\pi\epsilon_0} \left[ \frac{1}{r^3} \cdot [\vec{p}] + \vec{p} \cdot \vec{r} \left[ \frac{-3\vec{r}}{r^5} \right] \right]$$

$$\boxed{\vec{E} = \frac{1}{4\pi\epsilon_0} \left[ \frac{3(\vec{p} \cdot \vec{r})\vec{r}}{r^5} - \frac{\vec{p}}{r^3} \right]}$$

We know

$$\frac{d\vec{p}}{dr} = 0$$

$$\vec{\nabla}(\vec{p} \cdot \vec{r}) = \vec{p}$$

$$\vec{\nabla}\left(\frac{1}{r^3}\right) = \frac{-3\vec{r}}{r^5}$$

⊛  $\vec{\nabla}(\vec{A} \cdot \vec{r}) = \vec{A}$  for any  $\vec{A}$  (const.)

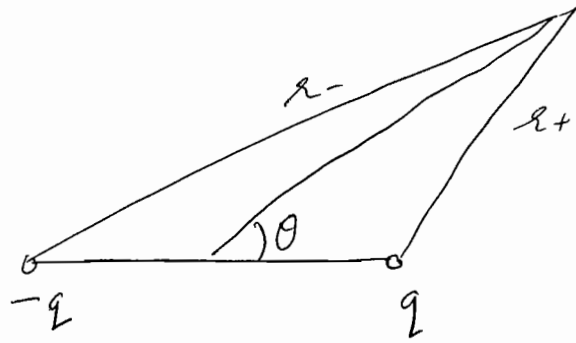
⊛  $\vec{\nabla}\left(\frac{1}{r^3}\right) = \sum \frac{\partial}{\partial x} \frac{1}{(x^2+y^2+z^2)^{3/2}} = \sum \frac{-3}{2} \cdot (x^2+y^2+z^2)^{-5/2} \cdot 2x = \left( \frac{-3\vec{r}}{r^5} \right)$

$$r_+ = \sqrt{r^2 + \left(\frac{l}{2}\right)^2 - r l \cos \theta}$$

$$= r \sqrt{1 + \left(\frac{l}{2r}\right)^2 - \left(\frac{l}{r}\right) \cos \theta}$$

$$= r \left[ 1 - \underbrace{\left\{ \frac{l}{r} \cos \theta - \left(\frac{l}{2r}\right)^2 \right\}}_{\epsilon_1} \right]^{1/2}$$

$$\frac{1}{r_+} = \frac{1}{r} \left[ 1 - \epsilon_1 \right]^{-1/2}$$



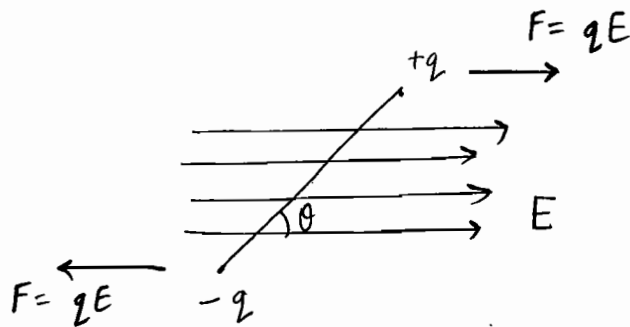
$$r_- = \sqrt{r^2 + \left(\frac{l}{2}\right)^2 + r l \cos \theta}$$

$$= r \left[ 1 + \underbrace{\left\{ \left(\frac{l}{2r}\right)^2 + \left(\frac{l}{r} \cos \theta\right) \right\}}_{\epsilon_2} \right]^{1/2}$$

$$r_- = r \left[ 1 + \epsilon_2 \right]^{+1/2}$$

$$\frac{1}{r_-} = \frac{1}{r} \left[ 1 + \epsilon_2 \right]^{-1/2}$$

$$V(r, \theta) = \frac{q}{4\pi\epsilon_0} \left[ \frac{1}{r_+} - \frac{1}{r_-} \right]$$



$$\underline{F_{\text{net}} = 0}$$

$$\tau = |\vec{r} \times \vec{F}| = q E l \sin \theta = p E \sin \theta$$

$$\underline{\tau = \vec{p} \times \vec{E}}$$

$$dU = +dw_{\text{ext}} = +\tau_{\text{ext}} d\theta = +[p E \sin \theta d\theta]$$

$$U - U_0 = -p E \cos \theta$$

$$\Rightarrow U - U_0 = -pE \cos \theta = -\vec{p} \cdot \vec{E}$$

★ इस पे दिमाग नहीं लगा ना !!

@  $\theta = 90^\circ$ ,  $U = U_0$  ✓

Take  $\cos \theta = +pE \sin \theta$   
& just mention,  
taking  $U|_{\theta=90^\circ} = 0$   
& do definite integral  $\int_{90}^{\theta}$

If the field is non-uniform,

$$\vec{F} = (\vec{p} \cdot \vec{\nabla}) \vec{E}$$

$$= \left( p_x \frac{\partial}{\partial x} + p_y \frac{\partial}{\partial y} + p_z \frac{\partial}{\partial z} \right) \cdot [E_x \hat{i} + E_y \hat{j} + E_z \hat{k}]$$

$$= \left[ p_x \frac{\partial E_x}{\partial x} + p_y \frac{\partial E_x}{\partial y} + p_z \frac{\partial E_x}{\partial z} \right] \hat{i}$$

$$\left[ p_x \frac{\partial E_y}{\partial x} + p_y \frac{\partial E_y}{\partial y} + p_z \frac{\partial E_y}{\partial z} \right] \hat{j}$$

$$\left[ p_x \frac{\partial E_z}{\partial x} + p_y \frac{\partial E_z}{\partial y} + p_z \frac{\partial E_z}{\partial z} \right] \hat{k}$$

[Note that for  $\vec{E} = E_0 \hat{x}$ ,  $\vec{F} = 0$ ]

$$\vec{d} = \vec{r}(\theta) - \vec{r}(0)$$

Derivation

$$\vec{F} = -q\vec{E}_- + q\vec{E}_+ = q[\vec{E}_+ - \vec{E}_-] = q d\vec{E}$$

$$= q [\Delta E_x \hat{i} + \Delta E_y \hat{j} + \Delta E_z \hat{k}]$$

$$= \sum q \Delta E_x \hat{i} = \sum q \cdot \vec{\nabla} E_x \cdot \vec{d} \hat{i}$$

~~Handwritten scribbles and crossed-out work.~~

$$= \sum q \vec{d} \cdot \vec{\nabla} E_x \hat{i}$$

$$= \sum \vec{p} \cdot \vec{\nabla} E_x \hat{i}$$

$$= (\vec{p} \cdot \vec{\nabla}) \vec{E}$$

$$\vec{F} = (\vec{p} \cdot \vec{\nabla}) \vec{E}$$

Main point was:  $\Delta E_x = \vec{\nabla} E_x \cdot \vec{d}$



$$dE_x = (\vec{E} \cdot \vec{\nabla}) dx$$

$$dE_y = (\vec{E} \cdot \vec{\nabla}) dy$$

$$dE_z = (\vec{E} \cdot \vec{\nabla}) dz$$

$$\Rightarrow \vec{F} = q [(\vec{E} \cdot \vec{\nabla}) dx \hat{i} + (\vec{E} \cdot \vec{\nabla}) dy \hat{j} + (\vec{E} \cdot \vec{\nabla}) dz \hat{k}]$$

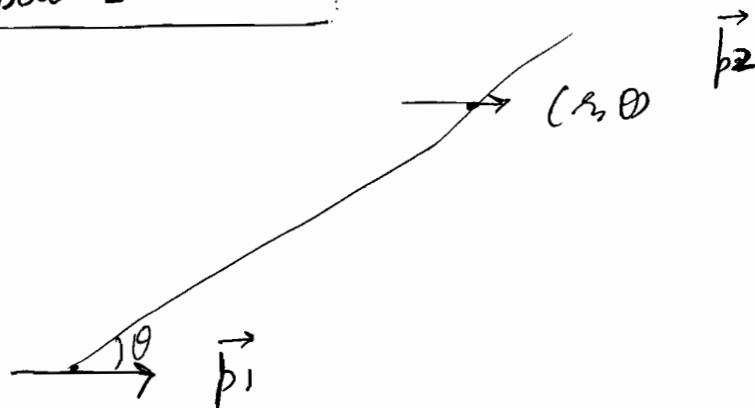
$$\boxed{F = q (\vec{d} \cdot \vec{\nabla}) \vec{E} = (\vec{p} \cdot \vec{\nabla}) \vec{E}}$$

Note that whether force is 0 or non-zero  $\tau$  will act [unless placed parallel to  $\vec{E}$ ]

$$\rightarrow \tau = \vec{r} \times \vec{F}$$

$$\rightarrow U = -\vec{p} \cdot \vec{E}$$

### Dipole Dipole Interaction



$$U = -\vec{p}_2 \cdot \vec{E}_1 = -\vec{p}_1 \cdot \vec{E}_2$$

$$\vec{E}_1 = \frac{1}{4\pi\epsilon_0} \left[ \frac{3(\vec{p}_1 \cdot \vec{r})}{r^5} \vec{r} - \frac{\vec{p}_1}{r^3} \right]$$

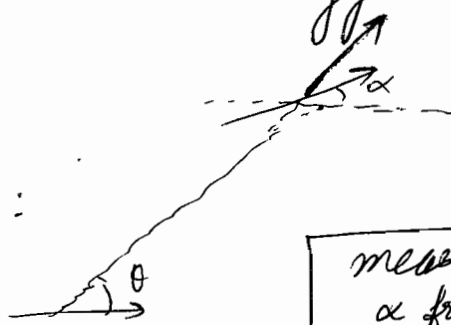
$$U = \frac{1}{4\pi\epsilon_0} \left[ \frac{(\vec{p}_1 \cdot \vec{p}_2)}{r^3} - \frac{3(\vec{p}_1 \cdot \vec{r})(\vec{p}_2 \cdot \vec{r})}{r^5} \right]$$

Most stable configuration, i.e. minimum energy

$p_1$  and  $p_2$ :  $\alpha$  angle

⊛ remember from multipole expansion:

( $\theta$ ) or ( $-\theta$ ) won't matter  
as dependence is upon  
 $\cos\theta$



measure  
 $\alpha$  from  $\hat{r}$

$$U = \frac{1}{4\pi\epsilon_0} \left[ \frac{p_1 p_2 \cos\alpha}{r^3} - \frac{3 p_1 \cos\theta p_2 \cos(\theta - \alpha)}{r^3} \right]$$

$$= \frac{1}{4\pi\epsilon_0 r^3} p_1 p_2 [ \cos\alpha - 3 \cos\theta \cos(\theta - \alpha) ]$$

$$= \frac{1}{4\pi\epsilon_0 r^3} p_1 p_2 \left[ \cos\alpha - \frac{3}{2} [ \cos(2\theta - \alpha) + \cos(\alpha) ] \right]$$

$$= \frac{1}{4\pi\epsilon_0 r^3} p_1 p_2 \left[ \frac{\cos\alpha}{2} + \frac{3 \cos(2\theta - \alpha)}{2} \right]$$

$$\frac{dU}{d\alpha} = 0 \quad -\sin\alpha + 3 \sin(2\theta - \alpha) = 0$$

$$\Rightarrow \sin(2\theta - \alpha) = \frac{1}{3} \sin\alpha$$

⊛

Note that upon calculation, we can see that  
 $\alpha$  points in the direction of  $E_1$

write  $\alpha = \theta + \phi$  &  $\phi = \tan^{-1} \left( \frac{\sin\theta}{2 \cos\theta} \right)$  ✓

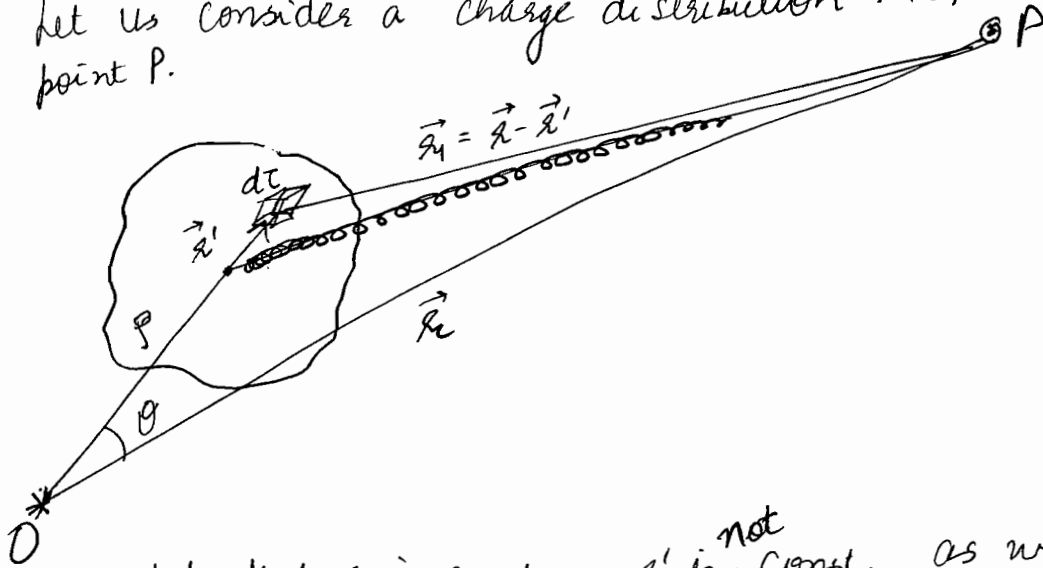


# E&M (6)

19/1/2012

## Multipole Expansion

→ let us consider a charge distribution. To find out Potential at point P.



Note that  $r$  is const.  $r'$  is <sup>not</sup> const. as we take different  $dq$ .

$$dV = \frac{1}{4\pi\epsilon_0} \frac{dq}{r_1}$$

$$V = \int dV = \int \frac{1}{4\pi\epsilon_0} \frac{\rho d\tau}{r_1}$$

$$r_1 = \sqrt{r^2 + r'^2 - 2rr'\cos\theta} = r \sqrt{\left(\frac{r'}{r}\right)^2 - 2\frac{r'}{r}\cos\theta + 1}$$

$$\frac{1}{r_1} = \frac{1}{r} (1 + \epsilon)^{-1/2}$$

$$\Rightarrow r_1 = r \sqrt{1 + \epsilon}$$

$$\epsilon = \left(\frac{r'}{r}\right) \left(\frac{r'}{r} - 2\cos\theta\right)$$

$$\frac{1}{r_1} = \frac{1}{r} \left[ 1 - \frac{1}{2}\epsilon + \frac{3}{8}\epsilon^2 - \frac{5}{16}\epsilon^3 + \dots \right]$$

$$= \frac{1}{r} \left[ 1 - \frac{1}{2} \left\{ \left(\frac{r'}{r}\right)^2 - 2\frac{r'}{r}\cos\theta \right\} + \frac{3}{8} \left\{ \left(\frac{r'}{r}\right)^4 + \frac{4r'}{r}\cos^2\theta - 4\left(\frac{r'}{r}\right)^2\cos\theta \right\} \right]$$

(पूरा खोल दिया....  
no approximation  
is used)  $\frac{5}{16} \{ \dots \}$

⊛ Only open till  $\left(\frac{3}{8}\right)$  term.... do not go for cubic.... till here we will get  $P_0, P_1$  and  $P_2$ ..... We can say that rest follow!!!

$$= \frac{1}{r_2} \left[ 1 \left( \frac{r'}{r} \right)^0 + \cos \theta \left( \frac{r'}{r} \right)^1 + \left( \frac{r'}{r} \right)^2 \left( \frac{1}{2} + \frac{3}{2} \cos^2 \theta \right) + \left( \frac{r'}{r} \right)^3 \left( \dots \right) \right]$$

$$\frac{1}{r_1} = \frac{1}{r} \sum_{n=0}^{\infty} \left( \frac{r'}{r} \right)^n P_n(\cos \theta)$$

Surprisingly the coefficients of  $\left( \frac{r'}{r} \right)^n$  come out as Legendre poly.

$\left( \frac{1}{r_1} \right)$  is called generating function of Legendre's polynomial.

$$\int dV = \frac{1}{4\pi\epsilon_0} \int \frac{\rho(r')}{r_1} = \frac{1}{4\pi\epsilon_0} \int \frac{\rho(r')}{r} \sum_{n=0}^{\infty} \left( \frac{r'}{r} \right)^n P_n(\cos \theta)$$

$$V = \frac{1}{4\pi\epsilon_0} \sum_{n=0}^{\infty} \frac{1}{r^{n+1}} \int \rho(r') (r')^n P_n(\cos \theta) d\tau'$$

**Multipole Expansion** This equation is exact and no approximation is used to derive it. But primarily it is used in approximation scheme: the lowest non-zero term in the expansion provides approximate potential at large  $r$ , and the successive terms tell us how to improve the approximation if more precision is required.

$$V(r, \theta) = \frac{1}{4\pi\epsilon_0} \int \frac{\rho(r')}{r} d\tau' + \frac{1}{4\pi\epsilon_0} \int \frac{\rho(r') r' \cos \theta}{r^2} d\tau' + \frac{1}{4\pi\epsilon_0} \int \frac{\rho(r') r'^2 (3\cos^2 \theta - 1)}{2r^3} d\tau'$$

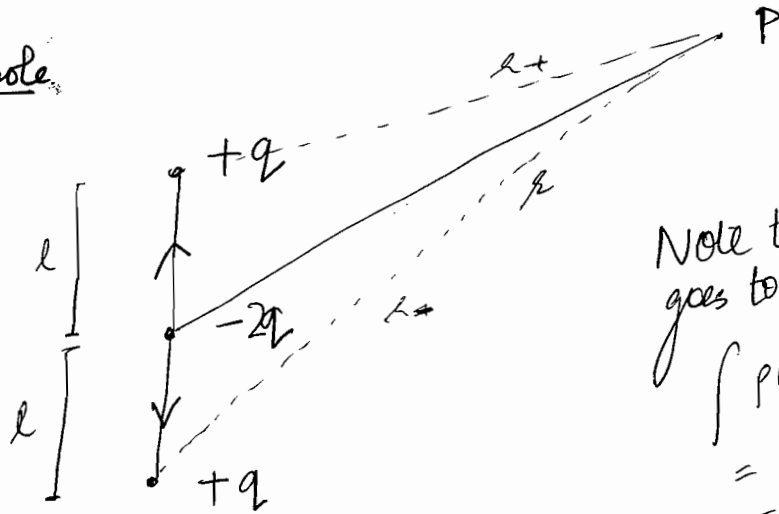
Monopole ( $n=0$ )  
 (choose charge @ the origin)  
 Dipole ( $n=1$ )  
 Quadrupole ( $n=2$ )

Hence a charge distribution can behave as Monopole, Dipole, Quadrupole, their combinations.

At  $r \gg r'$  : Monopole, At  $r \approx r'$  : Multipole  
 At  $r > r'$  : Dipole

EM waves are produced when charge particles are accelerated

### Linear Quadrupole



Note that dipole term goes to 0 as

$$\int \rho(x') x' \cos \theta \, dx' = q l \cos \theta - q l \cos \theta = 0$$

$$V(r, \theta) = \frac{1}{4\pi\epsilon_0} \frac{q l^2}{r^3} (3 \cos^2 \theta - 1) \quad \checkmark \text{ Quadrupole term of multipole expansion will come}$$

$$V(r, \theta) = \frac{1}{4\pi\epsilon_0} \left[ \frac{q}{r_+} + \frac{q}{r_-} - \frac{2q}{r} \right]$$

$$= \frac{q}{4\pi\epsilon_0 r} \left[ \frac{1}{\sqrt{1+\epsilon_1}} + \frac{1}{\sqrt{1+\epsilon_2}} - \frac{2}{1} \right]$$

$\frac{1}{\sqrt{1+\epsilon}} \approx 1 - \frac{\epsilon}{2} + \frac{3}{8}\epsilon^2 - \dots$   
 terms are dipole terms  
 they get cancelled or addition hence no dipole term

$$r_+ = r \sqrt{1 + \left(\frac{l}{r}\right)^2 - \frac{2l}{r} \cos \theta}$$

$$r_- = r \sqrt{1 + \left(\frac{l}{r}\right)^2 + \frac{2l}{r} \cos \theta}$$

$$\approx \frac{q}{4\pi\epsilon_0 r} \left[ (3 \cos^2 \theta - 1) \frac{l^2}{r^2} \right] \checkmark$$

### Dielectrics & Polarization

- Molecules are always in constant motion. i.e. Atoms are in motion; nucleus & electrons are in motion.
- Molecular level dipoles are randomly aligned in absence of  $\vec{E}_{ext}$ .

When  $\vec{E}_{ext}$  is applied, dipoles align in a particular direction.

Every dielectric can be modified.

Quantifiable quantity is Intensity of Polarization  $\vec{P}$

$\rightarrow \vec{P} =$  (No of dipoles per unit volume) in the direction of alignment of dipoles.

$k$ : dielectric const. electrical property of matter

$\epsilon_0$ : Permittivity of free space

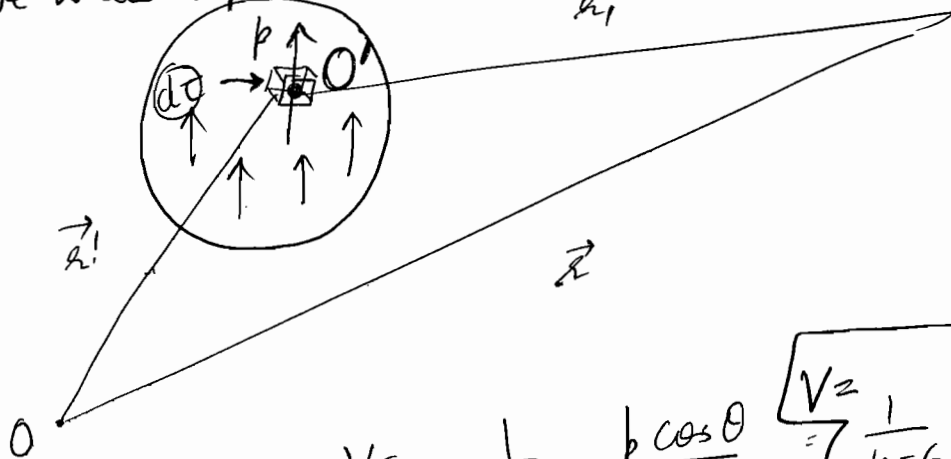
$\epsilon$ : Permittivity of a medium

$$\epsilon = \epsilon_0 k$$

$$\vec{p} = \vec{P} \cdot d\tau$$

$\rightarrow$  If  $d\tau$  is very small s.t. it contains a single dipole.

Since  $Q_{total} = 0 \Rightarrow$  we have Monopole term = 0  
 The only term we are interested in is dipole term. So we write dipole term taking the charge element as origin itself.



By taking the origin itself to O' we can write

$$V = \frac{1}{4\pi\epsilon_0} \frac{p \cos \theta}{r_1^2} \quad \boxed{V = \frac{1}{4\pi\epsilon_0} \frac{\vec{p} \cdot \vec{r}_1}{r_1^3}}$$

Now we will vary the origin itself

Due to  $d\tau$ ,  $dV = \frac{1}{4\pi\epsilon_0} \frac{\vec{P} \cdot \vec{r}_1}{r_1^3} = \frac{1}{4\pi\epsilon_0} \frac{\vec{P} \cdot d\tau \cdot \vec{r}_1}{r_1^3}$

We know

$$\boxed{\vec{\nabla} \cdot (\phi \vec{A}) = \phi \vec{\nabla} \cdot \vec{A} + \vec{A} \cdot \vec{\nabla} \phi}$$

1 scalar  
3 terms  
1 vector  
3 terms  
New

$$\nabla \cdot \left( \frac{1}{\epsilon_1} \vec{P} \right) = \frac{1}{\epsilon_1} \nabla \cdot \vec{P} + \vec{P} \cdot \nabla \left( \frac{1}{\epsilon_1} \right) \quad \text{--- (1)}$$

we can write

$$V(\epsilon_2, \theta) = \frac{1}{4\pi\epsilon_0} \int \vec{P} d\tau \cdot \nabla \left( \frac{1}{\epsilon_1} \right)$$

Note that  $r = \sqrt{(x-x')^2 + (y-y')^2 + (z-z')^2}$

Grad w.r.t. final point

$$\nabla \left( \frac{1}{r} \right) = \left( -\frac{\vec{r}}{r^3} \right)$$

final point is variable  
(x', y', z') : fixed

Grad. w.r.t. source point

$$\nabla \left( \frac{1}{r} \right) = \left( \frac{\vec{r}}{r^3} \right)$$

initial point is variable (x, y, z)  
fixed  $\Rightarrow$  one (-) sign of differentiation in numerator

Now using (1),

$$V(\epsilon_2, \theta) = \frac{-1}{4\pi\epsilon_0} \int_V \frac{1}{\epsilon_1} \nabla \cdot \vec{P} - \nabla \cdot \left( \frac{\vec{P}}{\epsilon_1} \right) d\tau$$

$$= \frac{-1}{4\pi\epsilon_0} \int_V \frac{1}{\epsilon_1} (\nabla \cdot \vec{P}) d\tau + \frac{1}{4\pi\epsilon_0} \int_V + (\nabla \cdot \frac{\vec{P}}{\epsilon_1}) d\tau$$

From reverse of divergence theorem

$$= \frac{+1}{4\pi\epsilon_0} \int_V - \frac{(\nabla \cdot \vec{P})}{\epsilon_1} d\tau + \frac{1}{4\pi\epsilon_0} \oint_S \frac{+\vec{P} \cdot d\vec{s}}{\epsilon_1}$$

Hence  $-\nabla \cdot \vec{P} = \rho_b$  defining ✓  
 Bound Volume Charge Density

$+(\vec{P} \cdot \hat{n}) = \sigma_b$  Bound Surface Charge Density ✓

$$V(\zeta, \theta) = \frac{1}{4\pi\epsilon_0} \int_V \frac{(\rho_b d\tau)}{r_1} + \frac{1}{4\pi\epsilon_0} \oint_S \frac{\sigma_b ds}{r_1}$$

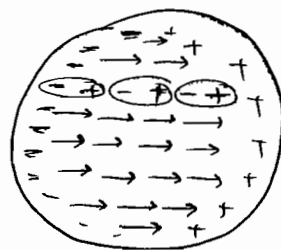
where  $\rho_b = -\vec{\nabla} \cdot \vec{P}$  ✓  
 $\sigma_b = \vec{P} \cdot \hat{n}$  ✓

✓ If matter is uniformly polarized  $\Rightarrow P$  is same everywhere  
 $\Rightarrow \vec{\nabla} \cdot \vec{P} = 0$

Hence  $\rho_b$  arises only when non-uniform Polarization.

✓ Note that the charge here is not conduction charge like current. We cannot light bulb from this charge flow. ( $\because$  its fixed charge distribution & not a flowing charge)

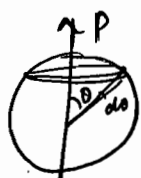
✓ Surface charge density will be there whenever Polarization occurs.



✓ Gas & liquids are always uniformly polarized, hence  $-\vec{\nabla} \cdot \vec{P} = \rho_b = 0$  for fluids.

$$q_b = \int_V \rho_b d\tau + \int_S \sigma_b ds$$

Note that  $\oint \sigma_b ds = 0$  ✓



$$Q_s = \oint \sigma_b ds = \int_0^\pi P \cos\theta \cdot 2\pi R^2 \sin\theta d\theta = 0$$



Q29)

Do not derive...

-42

Just use  $-\vec{\nabla} \cdot \vec{P}$

Write 2-3 lines of Physics

### Polarizability :-

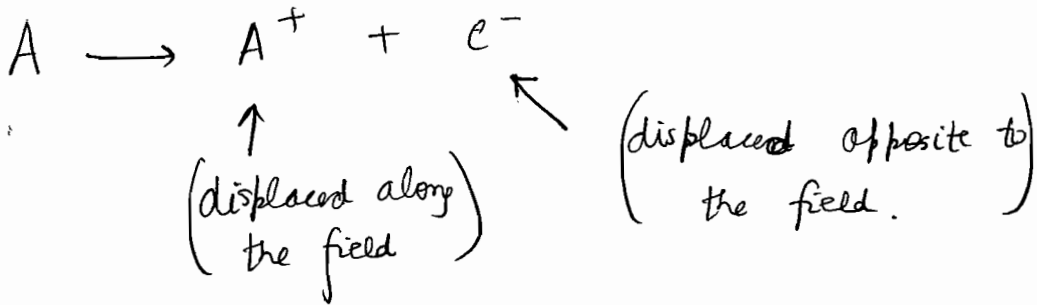
Dipole ~~moment~~ Moment of a molecules due to field experienced by a molecule.

$$\vec{p}_m = \alpha \vec{E}_0$$

$\uparrow$  Constant of Proportionality  
 $\nwarrow$  applied field

$\alpha$  is due to

- valency electrons ( $\alpha_e$ )
- ionic polarizability ( $\alpha_{io}$ )



### • Orientation Polarizability ( $\alpha_{orientation}$ )

→ atomic polarizability = [electronic + ionic + orientation] polarizabilities

hence,  $\alpha = [\alpha_{elect} + \alpha_{ionic} + \alpha_{orientation}]$

$$\vec{p}_m = \alpha \vec{E}_0 \checkmark$$

note that this derived for general dielectric (sphere) and not just linear

### Field due to Uniform Polarization :-

$$\vec{E}_m = \vec{E}_{else} - \frac{\vec{P}}{3\epsilon_0}$$

inside the dielectric

✓ It behaves as a dipole with  $|\vec{p}| = |\vec{P}| \times \text{Volume}$  Outside the dielectric

$\vec{P} = \left(\frac{4}{3}\pi R^3\right) \vec{P}$

$$\vec{E}_{\text{Polarization}} = \frac{1}{4\pi\epsilon_0} \int \frac{\nabla_b ds}{r^3} \vec{r}$$

$$= -\left(\frac{P}{3\epsilon_0}\right) \checkmark$$

### Classius - Mossotte Relation

Relates 'α' with no. of dipoles per unit volume for fluids.

Let there be n dipoles per unit volume of dielectric medium

$$\Rightarrow \vec{P} = n p_m = n \alpha \vec{E}_{\text{ext}}$$

$$= n \alpha \left[ \vec{E}_{\text{total}} + \frac{\vec{P}}{3\epsilon_0} \right]$$

$$\Rightarrow \vec{P} \left[ 1 - \frac{n \alpha}{3\epsilon_0} \right] = n \alpha \vec{E}_{\text{total}} \quad \text{--- (1)}$$



We know from Gauss law,

$$\nabla \cdot \vec{E} = \frac{\rho}{\epsilon_0}$$

$$\nabla \cdot \vec{E} = \frac{\rho_{\text{conduction}} + \rho_{\text{bound}}}{\epsilon_0} = \left( \frac{\rho_{\text{bound}} + \rho_{\text{free}}}{\epsilon_0} \right)$$

$$\Rightarrow \nabla \cdot (\epsilon_0 \vec{E}) = -(\vec{\nabla} \cdot \vec{P}) + \rho_{\text{free}}$$

$$\Rightarrow \nabla \cdot (\epsilon_0 \vec{E} + \vec{P}) = \rho_{\text{free}}$$

let  $\vec{D} = \epsilon_0 \vec{E} + \vec{P}$

$$\Rightarrow \nabla \cdot \vec{D} = \rho_{\text{free}}$$

Gauss law in dielectric medium

For air,  $P = 0$

$$\vec{D} = \epsilon_0 \vec{E}$$

For any dielectric

$$\underline{\underline{\vec{D} = \epsilon_0 \vec{E} + \vec{P} = \epsilon \vec{E} = k \epsilon_0 \vec{E}}}$$

Putting in ①

$$\frac{|\vec{P}|}{|\epsilon_0 \vec{E}|} = (k-1) \epsilon_0$$

let  $\frac{P}{\epsilon_0 E} = \chi$  : dielectric susceptibility

$$\boxed{1 + \chi = k}$$

$$\Rightarrow 1 - \frac{n\alpha}{3\epsilon_0} = \frac{n\alpha}{(k-1)\epsilon_0} \checkmark$$

$$\Rightarrow 1 \neq n\alpha \left[ \frac{1}{(k-1)\epsilon_0} + \frac{1}{3\epsilon_0} \right] = \frac{n\alpha}{3\epsilon_0} \left[ \frac{3}{k-1} + 1 \right]$$

$$\Rightarrow 1 = \frac{n\alpha}{3\epsilon_0} \left( \frac{k+2}{k-1} \right) \checkmark$$

$$\Rightarrow \boxed{\frac{(k-1)}{(k+2)} = \frac{n\alpha}{3\epsilon_0}}$$

Classius Mossotte Relationship

Microscopic property

macroscopic property

$$\alpha = \frac{3}{4\pi N} \left( \frac{\epsilon - 1}{\epsilon - 2} \right)$$

### Derivation of Classius Mossotte

Let a dielectric object kept in field  $\vec{E}_{ext}$ .

Now  $\vec{P} = \text{[scribble]} n\alpha \vec{E}_{ext} \text{ --- (1)}$

Also for linear dielectric  $\vec{P} = \chi_e \epsilon_0 \vec{E}_{total} \text{ --- (2)}$

$$\Delta \vec{E}_{total} = \vec{E}_{ext} - \frac{\vec{P}}{3\epsilon_0} \text{ --- (3)}$$

$$\Rightarrow \frac{\vec{P}}{\chi_e \epsilon_0} = \frac{\vec{P}}{n\alpha} - \frac{\vec{P}}{3\epsilon_0}$$

$$\Rightarrow \frac{1}{\chi_e \epsilon_0} + \frac{1}{3\epsilon_0} = \frac{1}{n\alpha}$$

$$\Rightarrow \frac{n\alpha}{3\epsilon_0} = \frac{\chi_e \epsilon_0}{3\epsilon_0 + \chi_e \epsilon_0} = \frac{\chi_e}{3 + \chi_e} = \underline{\underline{\left( \frac{\epsilon_r - 1}{\epsilon_r + 2} \right)}}$$

$$\rho_b = \nabla \cdot \vec{P} = 0$$

$$\sigma_b = \vec{P} \cdot \hat{z} = P \cos \theta$$

$$\vec{E} = \frac{1}{4\pi\epsilon_0} \int \frac{\sigma_b ds}{r^2} \hat{r}$$

$$ds = 2\pi r \sin \theta r d\theta$$

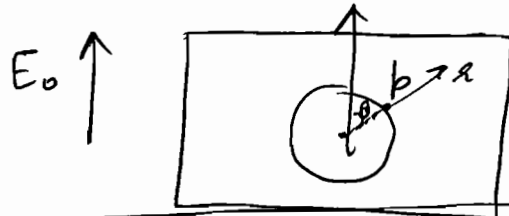
$$\vec{E} = \frac{1}{4\pi\epsilon_0} \int_0^\pi \frac{(P \cos \theta) \cdot \cos \theta ds}{r^2}$$

$$= \frac{P}{4\pi\epsilon_0} \int_0^\pi \frac{\cos^2 \theta \cdot 2\pi r^2 \sin \theta d\theta}{r^2}$$

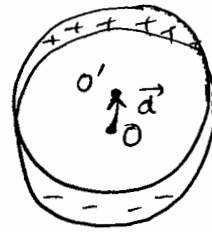
$$= \frac{P}{2\epsilon_0} \int_0^\pi \cos^2 \theta \sin \theta d\theta$$

$$= \frac{P}{2\epsilon_0} \cdot \frac{2}{3} = \frac{\vec{P}}{3\epsilon_0}$$

$$\vec{E} = \frac{\vec{P}}{3\epsilon_0}$$



To analyze uniformly polarized sphere, we can take the following analogy:



2 spheres of charge (opposite) are overlapped but displaced slightly. The leftover charge at top & bottom are equivalent to Polarization surface charge density.

At any point in region of overlap, we can calculate from Gauss law

$$\vec{E} = -\frac{1}{4\pi\epsilon_0} \cdot \frac{q \vec{d}}{R^3}$$

We can express it in terms of Polarization of sphere

$$\text{as } \vec{P} = q \vec{d} = \left(\frac{4}{3}\pi R^3\right) \vec{P}$$

$$\Rightarrow \vec{E} = -\left(\frac{P}{3\epsilon_0}\right)$$

For outside points,  $V = \frac{1}{4\pi\epsilon_0} \left(\frac{\vec{P} \cdot \hat{z}}{R^2}\right)$

# Electric Field in Matter

✓ ⊛ On a broad scale, materials are divided into two types: Conductors and Insulators (or dielectrics)

Conductors are substances that have "unlimited" supply of charges that are free to move.

In dielectrics, in contrast, all charges are attached to specific atom or molecule - they are not free to move but can move a bit within the atom or molecule. Such microscopic displacements are not as dramatic as in a conductor, but their cumulative effects account for characteristic behaviour of dielectrics.

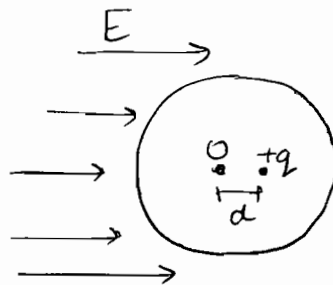
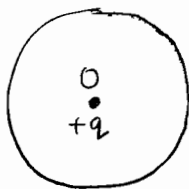
✓ ⊛ When placed in an Electric Field (weak), a neutral atom acquires a tiny dipole moment  $\vec{p}$  which points in the same direction as  $\vec{E}$ . (Note that if the field is large enough, it can pull the atom apart completely, "ionizing" it and the dielectric becomes a conductor)

$$\vec{p} = \alpha \vec{E}_{\text{applied}}$$

$\alpha$ : atomic polarizability

✓ ⊛ Consider a model of atom (volume =  $v$ ) of a point nucleus (+ $q$ ) surrounded by a uniformly spherical cloud (- $q$ ) of radius  $a$ . Calculate atomic polarizability of such an atom.

Ans/



In presence of electric field  $\vec{E}$ , the nucleus will be shifted towards right and the electron cloud to the left. Assume e-cloud maintains spherical shape.

Now at equilibrium, Force upon nucleus balance.

$$\Rightarrow E q = \frac{1}{4\pi d^2} \cdot \frac{Q^2}{\epsilon_0} \cdot d^3 \quad \Rightarrow \quad \frac{Q d}{4\pi \epsilon_0 a^3} = E \quad \Rightarrow \quad p = 4\pi \epsilon_0 \cdot a^3 E$$

$\Rightarrow \frac{p}{E} = 3 v \epsilon_0$

✓ If the material (dielectric) already has dipole moment, then the tiny dipoles experience a torque and align along electric field.

∴ the two mechanisms (induced dipole moment and torque) produce the same basic result: a lot of dipoles pointing along the direction of the field - the material becomes polarized. A convenient measure of this effect is  $\underline{P}$ , the dipole moment per unit volume. Its called POLARIZATION.

✓ Solving for potential, we get equivalent of dielectric of Polarization  $P$  is

$$\begin{aligned} \text{surface charge of } & \vec{P} \cdot \hat{n} \\ \text{Volume charge of } & -\vec{\nabla} \cdot \vec{P} \end{aligned}$$

✓ A specified charge density  $\sigma_0(\theta)$  is glued over surface of a spherical shell of radius  $R$ . Find the resulting potential inside and outside of sphere.

Interior

$$V(r, \theta) = \sum_{n=0}^{\infty} \left( A_n r^n + \frac{B_n}{r^{n+1}} \right) P_n(\cos \theta)$$

जहाँ  $\vec{E}$   
शोरत  
 $\vec{E}!!$

So that  $V(r, \theta)$  does not blow up at  $r=0$ , we have  $B_n = 0$

$$\Rightarrow V(r, \theta) = \sum_{n=0}^{\infty} A_n r^n P_n(\cos \theta)$$

Outside

$$V(r, \theta) = \sum_{n=0}^{\infty} \left( A_n r^n + \frac{B_n}{r^{n+1}} \right) P_n(\cos \theta)$$

So that  $V(r, \theta)$  goes to 0 at  $\infty$ , we have  $A_n = 0$  ( $n > 0$ )

$$\Rightarrow V(r, \theta) = \sum_{n=0}^{\infty} \frac{B_n}{r^{n+1}} P_n(\cos \theta)$$

① The Potential must be continuous at  $r=R$

$$\sum_{n=0}^{\infty} A_n R^n P_n(\cos\theta) = \sum_{n=0}^{\infty} B_n R^{-(n+1)} P_n(\cos\theta)$$

Comparing coefficients of  $P_n(\cos\theta)$

$$B_n = A_n R^{2n+1}$$

② The radial derivative of  $V$  suffers a discontinuity at the surface

(From Gauss law,  $E_{above} - E_{below} = \frac{\sigma}{\epsilon_0} \hat{n}$ )

$$\Rightarrow \left. \frac{\partial V_{out}}{\partial r} - \frac{\partial V_{in}}{\partial r} \right|_{r=R} = -\frac{1}{\epsilon_0} \sigma_0(\theta)$$

$$\Rightarrow -\sum_{n=0}^{\infty} (n+1) \frac{B_n}{R^{n+2}} P_n(\cos\theta) - \sum_{n=0}^{\infty} n A_n R^{n-1} P_n(\cos\theta) = -\left(\frac{\sigma_0(\theta)}{\epsilon_0}\right)$$

$$\sum (2n+1) A_n R^{n-1} P_n(\cos\theta) = \frac{\sigma_0(\theta)}{\epsilon_0}$$

Using Fourier not reqd.

$$A_n = \frac{1}{2\epsilon_0 R^{n+1}} \int_0^\pi \sigma_0(\theta) P_n(\cos\theta) \sin\theta d\theta$$

Specific Case

$$\sigma_0(\theta) = k \cos\theta$$

All  $A_n$  are zero except  $n=1 \Rightarrow 3A_1 \cos\theta = \frac{k \cos\theta}{\epsilon_0} \Rightarrow A_1 = \left(\frac{k}{3\epsilon_0}\right)$

$$\Rightarrow A_1 = \frac{k}{2\epsilon_0} \int_0^\pi \cos^2\theta \sin\theta d\theta = \frac{k}{2\epsilon_0} \cdot \left(\frac{2}{3}\right) = \left(\frac{k}{3\epsilon_0}\right)$$

$$\Rightarrow B_1 = \left(\frac{k R^3}{3\epsilon_0}\right)$$



$$\Rightarrow V_{\text{inside}} = \frac{k}{3\epsilon_0} r \cos\theta \quad (r \leq R)$$

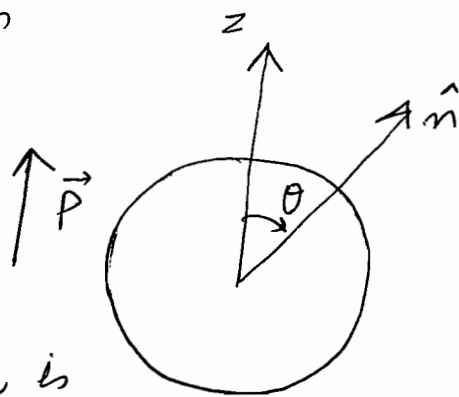
$$V_{\text{outside}} = \frac{k}{3\epsilon_0} \frac{R^3}{r^2} \cos\theta \quad (r \geq R)$$

## Field due to uniformly polarized sphere of radius R

Let us choose  $\hat{z}$  as the direction of  $\vec{P}$

Now  $\rho_b = 0$  (uniform  $\vec{P}$ )

$$\sigma_b = \vec{P} \cdot \hat{n} = P \cos\theta$$



Now the potential due to  $\sigma_b$ , is given as

$$V_{\text{inside}} = \frac{P}{3\epsilon_0} r \cos\theta \quad (r \leq R)$$

$$V_{\text{outside}} = \frac{P}{3\epsilon_0} \frac{R^3}{r^2} \cos\theta \quad (r \geq R)$$

Note that these results can be directly lifted as the uniform polarization ensures that Laplace equation is valid both inside and outside the sphere.

$$\vec{E}_{\text{in}} = -\vec{\nabla} V_{\text{in}} = -\frac{P}{3\epsilon_0} \hat{z} = \underline{\underline{-\frac{1}{3\epsilon_0} \vec{P}}} \quad r < R$$

$$\begin{aligned} \vec{E}_{\text{out}} &= -\vec{\nabla} V_{\text{out}} = \frac{P R^3}{3\epsilon_0} \frac{2}{r^3} \cos\theta (\hat{r}) + \frac{P R^3}{3\epsilon_0 r^2} \sin\theta \hat{\theta} \\ &= \frac{P R^3}{3\epsilon_0 r^3} (2 \cos\theta \hat{r} + \sin\theta \hat{\theta}) \end{aligned}$$

Now using  $p_{\text{total}} = P \cdot \frac{4}{3} \pi R^3$

$$= \frac{p_{\text{total}}}{4\pi\epsilon_0 r^3} (2 \cos\theta \hat{r} + \sin\theta \hat{\theta})$$

$\therefore$  outside the sphere, it acts as dipole of dipole moment  $p_{\text{total}} = P \cdot V$

$r > R$

3.3  
 (★) Note that for most dielectrics,  $\nabla \cdot \mathbf{P} = 0 \Rightarrow \rho_b = 0$   
 Hence  $\int \sigma_b \cdot d\vec{a} = 0 \quad \therefore$  the overall dielectric is neutral.

But in the case of non uniform Polarization,  $\rho_b$  is not zero and neither is charge due to  $\sigma_b$

i.e.  $Q_{\text{surface}} + Q_{\text{volume}} = 0$

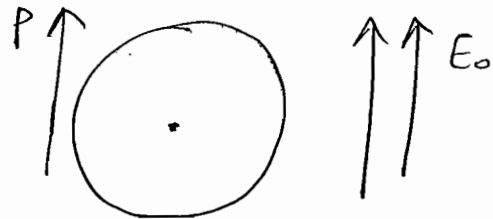
but individually they may not be zero.

$$\begin{aligned} \text{Also, } Q_{\text{vol}} &= \int_V \rho_b \, d\tau = - \int_V \nabla \cdot \vec{P} \, d\tau = - \oint \vec{P} \cdot d\vec{a} \\ &= - \oint \sigma_b \, ds = - Q_{\text{surface}} \end{aligned}$$

$$\Rightarrow Q_{\text{volume}} + Q_{\text{surface}} = 0$$

Field inside dielectric in presence of  $\vec{E}$

$$\begin{aligned} E_{\text{inside}} &= E_{\text{outside}} + E_{\text{polarization}} \\ &= \left( E_0 - \frac{P}{3\epsilon_0} \right) \hat{z} \end{aligned}$$



(Note that  $\vec{E}$  obeys superposition principle always... even for conductor

Gauss law for dielectrics

$$E_{\text{in}} = E_{\text{applied}} + E_{\text{conductor, charge}} = 0$$

From Gauss law

$$\nabla \cdot \vec{E} = \left( \frac{\rho}{\epsilon} \right)$$

Now in a dielectric  $\rho$  may be of two types: bound charge ( $\rho_b = -\nabla \cdot \vec{P}$ ) and other free charges that might

be present. Let's call them  $\rho_f$ . Free charge may consist of electrons in a conductor that may be present in dielectric medium or ions embedded in dielectric medium, or whatever charge that is not a result of polarization.

$$\Rightarrow \boxed{\nabla \cdot \vec{E} = \frac{\rho}{\epsilon_0}}$$

Note that Gauss law remains same but it can be expressed in a different form for dielectric medium

$$\Rightarrow \vec{\nabla} \cdot \vec{E} = \frac{\rho_b + \rho_f}{\epsilon_0} = \frac{-\nabla \cdot \vec{P} + \rho_f}{\epsilon_0}$$

$$\Rightarrow \epsilon_0 \vec{\nabla} \cdot \vec{E} + \vec{\nabla} \cdot \vec{P} = \rho_f$$

$$\Rightarrow \vec{\nabla} \cdot (\epsilon_0 \vec{E} + \vec{P}) = \rho_f$$

Define Electric Displacement Vector  $\vec{D}$  as

$$\vec{D} = \epsilon_0 \vec{E} + \vec{P}$$

$$\Rightarrow \boxed{\vec{\nabla} \cdot \vec{D} = \rho_f}$$

Alternate 'form' of Gauss law for dielectric medium.

In integral form,  $\oint \vec{D} \cdot d\vec{a} = Q_{\text{free enclosed}}$

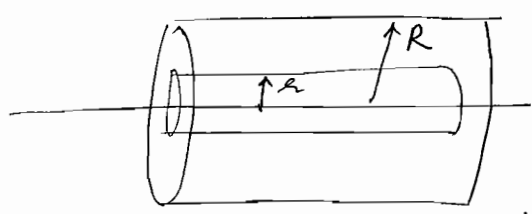
In typical situations, we usually know  $\rho_{\text{free}}$  but not  $\rho_{\text{bound}}$ ,  $\therefore$  this form of Gauss law is particularly useful for dielectrics.

✓ ~~Q~~ A long straight wire carrying uniform line charge  $\lambda$ , is surrounded by rubber insulation out to a radius  $R$ . Find electric displacement.

A1 From Gauss law

$$D \cdot 2\pi r \cdot l = \frac{\lambda l}{\epsilon_0}$$

$$\vec{D} = \frac{\lambda}{2\pi r} \hat{s}$$



Now note that this formula holds both within and outside the insulation. In the latter region,  $\vec{P} = 0$

$$\Rightarrow \vec{D} = \epsilon_0 \vec{E} + \vec{P} = \epsilon_0 \vec{E}$$

$$\Rightarrow \vec{E}_{outside} = \frac{\vec{D}}{\epsilon_0} = \frac{\lambda}{2\pi \epsilon_0 r} \hat{s}$$

Inside, the electric field cannot be determined, since we do not know  $\vec{P}$ .

### Some Properties of $\vec{D}$

① No Coulombs law for  $\vec{D}$

② Curl of  $\vec{D} \neq 0$

$$\nabla \times \vec{D} = \epsilon_0 (\nabla \times \vec{E}) + \nabla \times \vec{P} = 0 + \nabla \times \vec{P} = (\nabla \times \vec{P}) \neq 0$$

③ There is no Potential corresponding to  $\vec{D}$ .

④ Boundary Condition

$$D_{above}^\perp - D_{below}^\perp = \sigma_{free}$$

$$D_{above}'' - D_{below}'' = P_{above}'' - P_{below}'' \quad (\text{from def}'' \text{ of } \vec{D})$$

from integral form of Gauss Law for dielectrics  
we can get from def'' also  $\sigma_b = \dots$

## LINEAR DIELECTRICS

For linear dielectrics,

$$\vec{P} \propto \vec{E}$$

$$\Rightarrow \vec{P} = \epsilon_0 \chi_e \vec{E}$$

$\chi_e$ : electric susceptibility (dimensionless quantity)

Note that  $\vec{E}$  is the total field, it may be due to

- ① free charges or
- ② polarization itself

If for instance, we put a piece of dielectric into an external field  $\vec{E}_0$ , we cannot compute  $\vec{P}$  directly from this equation as the external field will polarize the material, and this polarization will produce its own field, which will contribute to total field ... which in turn modifies polarization and so on till an equilibrium  $E$  is achieved. The simplest approach is to begin with displacement  $\vec{D}$ , at least in those cases where  $\vec{D}$  can be deduced directly from free charge dist<sup>n</sup>.

In linear media,

$$\vec{D} = \epsilon_0 \vec{E} + \vec{P} = \epsilon_0 \vec{E} + \chi_e \epsilon_0 \vec{E} = \epsilon_0 (1 + \chi_e) \vec{E}$$

define permittivity of material  $\epsilon$ , s.t.

$$\epsilon = \epsilon_0 (1 + \chi_e)$$

$$\Rightarrow \boxed{\vec{D} = \epsilon \vec{E}}$$

define relative permittivity or dielectric constant of the medium,  $\epsilon_r$

$$\epsilon_r = \frac{\epsilon}{\epsilon_0} = (1 + \chi_e)$$

⊛ Note that while Polarization is usually in the direction of applied external field while the electric field due to Polarization is opposite to  $\vec{P}$  i.e. usually opposite to applied field. Therefore, like a conductor, a dielectric tries to minimize field inside it as much as it can, but the cancellation is partial (unlike conductors where cancellation is complete and  $\vec{E}_{\text{inside}} = 0$ ). Dielectric is therefore, a poor conductor.

We have calculated field for a dielectric as

$$\vec{E}_{inside} = \vec{E}_0 - \frac{\vec{P}}{3\epsilon_0}$$

$E_0$ :  $E_{applied}$   
 $E_{inside}$ :  $E_{total}$

For linear dielectric,  $\vec{P} = \epsilon_0 \chi_e \vec{E}_{total}$

$$\Rightarrow \vec{E} = \vec{E}_0 - \frac{\epsilon_0 \chi_e E}{3\epsilon_0}$$

$$\Rightarrow \vec{E} \left(1 + \frac{\epsilon_0 \chi_e}{3\epsilon_0}\right) = \vec{E}_0$$

$$\Rightarrow \vec{E} = \frac{\vec{E}_0}{\left(1 + \frac{\epsilon_0 \chi_e}{3\epsilon_0}\right)} = \frac{\vec{E}_0}{1 + \frac{\chi_e}{3}} = \left(\frac{3\vec{E}_0}{\chi_e + 3}\right) = \underline{\underline{\left(\frac{3\vec{E}_0}{\epsilon_r + 2}\right)}}$$

$$\boxed{\vec{E} = \left(\frac{3}{\epsilon_r + 2}\right) \vec{E}_0}$$

Also, for a linear dielectric,

$$\rho_b = -\vec{\nabla} \cdot \vec{P}$$
$$= -\vec{\nabla} \cdot (\epsilon_0 \chi_e \vec{E})$$

$$= -\vec{\nabla} \cdot \left(\epsilon_0 \chi_e \frac{\vec{D}}{\epsilon}\right)$$

$$= -\vec{\nabla} \cdot \left(\frac{\epsilon_0 \chi_e}{(1+\chi)\epsilon_0} \vec{D}\right)$$

$$= -\frac{\chi_e}{1+\chi_e} \vec{\nabla} \cdot \vec{D}$$

$$= -\left(\frac{\chi_e}{1+\chi_e}\right) \rho_{free}$$

$$\boxed{\rho_b = -\left(\frac{\chi_e}{1+\chi_e}\right) \rho_{free}}$$

for linear dielectric

## Boundary Conditions for linear dielectrics

We know,

$$D_{\text{above}}^{\perp} - D_{\text{below}}^{\perp} = \sigma_{\text{free}}$$

$$\Rightarrow \boxed{\epsilon_{\text{above}} E_{\text{above}}^{\perp} - \epsilon_{\text{below}} E_{\text{below}}^{\perp} = \sigma_{\text{free}}}$$

$$\Rightarrow \epsilon_{\text{above}} \left( \frac{\partial V_{\text{above}}}{\partial n} \right) - \epsilon_{\text{below}} \left( \frac{\partial V_{\text{below}}}{\partial n} \right) = -\sigma_f$$

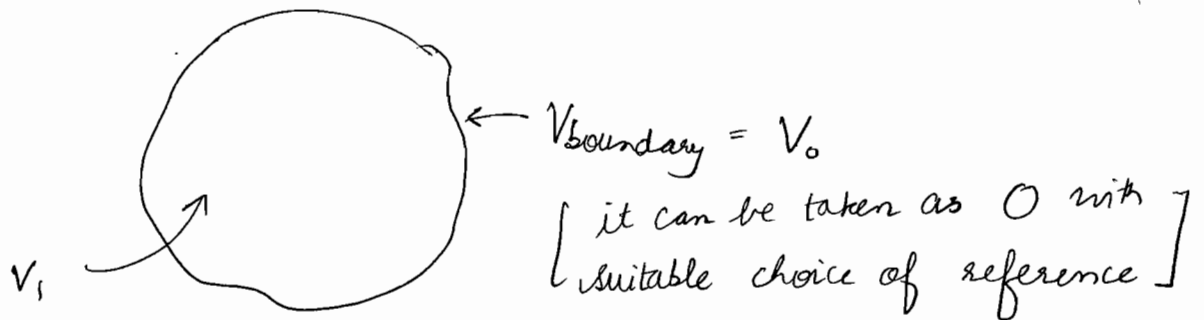
Also

$$\boxed{V_{\text{above}} = V_{\text{below}}} \text{ at } \underline{\text{surface only}}$$

## Method of Electrical Images

Not all field problems can be solved easily with method of electrical images.

This method is based upon 'Uniqueness Theorem in Electrostatics' i.e. for a particular boundary conditions, the solution of Laplace & Poisson equation is unique. i.e. one and only one. ✓



Now for given  $V_0$ ,  $V_1$  will be unique!!!

### Proof by Contradiction

$$\nabla^2 V = 0$$

✓

Let there be 2 solutions in the region under consideration whose boundary is at Const. Potential (say 0)

The solution of differential equations gives the average values of the functions. They do not tolerate local maxima or local minima. Extreme values of functions lie on the boundary.

Now, Maxima of  $V_1 = 0$   
Minima of  $V_1 = 0$   
 $\Rightarrow V_1 = 0$

Similarly, Maxima of  $V_2 = 0$   
Minima of  $V_2 = 0$   
 $\Rightarrow V_2 = 0$

$$\Rightarrow V_1 = V_2$$



If  $V_1$  and  $V_2$  are solutions, hence linear combination of  $V_1$  and  $V_2$  are also solutions.

$\Rightarrow \alpha_1 V_1 + \alpha_2 V_2$  are solutions of  $\nabla^2 V = 0$

$$\nabla^2 (\alpha_1 V_1 + \alpha_2 V_2) = \alpha_1 \nabla^2 V_1 + \alpha_2 \nabla^2 V_2 = 0 + 0 = 0$$

Hence  $V_1 - V_2$  will be a solution,

$$\nabla^2 (V_2 - V_1) = \nabla^2 V_2 - \nabla^2 V_1 = 0 - 0 = 0$$

Its maxima & minima (of  $V_1 - V_2$ ) will lie on the boundary. It will not tolerate local minima & local maxima.

$$\text{Maxima of } (V_1 - V_2) = 0$$

$$\text{Minima of } (V_1 - V_2) = 0$$

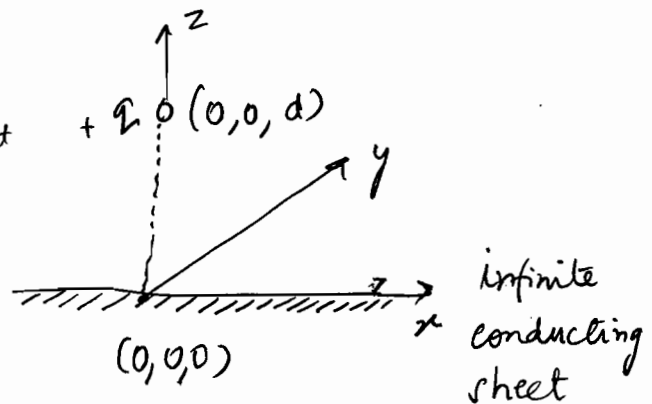
$$\Rightarrow V_1 - V_2 = 0$$

$$\Rightarrow V_1 = V_2$$

Note that apart from  $\vec{E}, \vec{F}, U, V, \sigma$  can also be calculated from the 5 methods.  
There are only two standard problems:

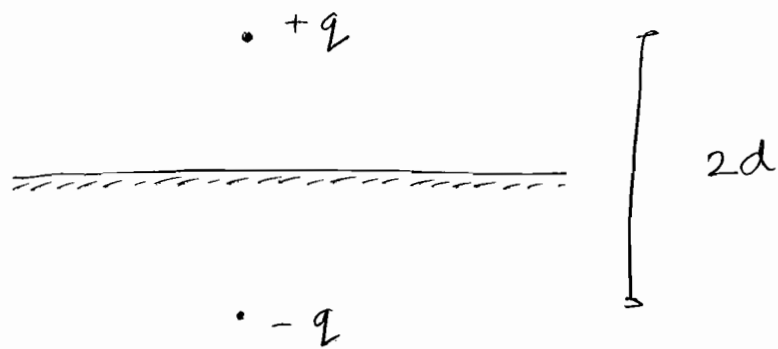
### ① Infinitely Conducting Sheet

At  $z=0$ , Potential is always constant  
ie.  $V_{z=0} = 0$



Now, according to method of electrical images,

the given problem is equivalent to



Note that main difficulty is to replicate the same boundary conditions

$$F = \frac{1}{4\pi\epsilon_0} \left( \frac{q^2}{4d^2} \right)$$

Force between 'sheet & charge' is attractive.

$$V(x, y, z) = \frac{1}{4\pi\epsilon_0} \left[ \frac{q}{\sqrt{x^2 + y^2 + (z-d)^2}} - \frac{q}{\sqrt{x^2 + y^2 + (z+d)^2}} \right] \quad \text{--- (1)}$$

$$V(x, y, 0) = 0 - 0 = 0 \quad \text{[Boundary Condition]}$$

$$dw = \frac{1}{2} \int F dz$$

← Note that Work is half as the other charge is just virtual

From (1)

$$\vec{E} = -\vec{\nabla} V$$

From  $E(x, y, z)$  we can find  $E(x, y, 0)$

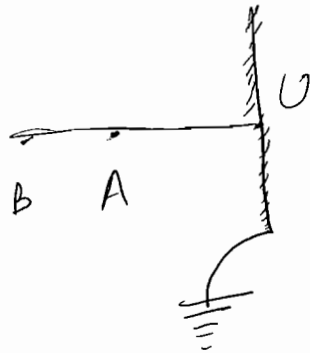
$$\checkmark E(x, y) = \frac{\sigma}{\epsilon_0}$$

$$\boxed{\sigma = \epsilon_0 E(x, y)}$$



$$Q = \int_{-\infty}^{\infty} \int \sigma \, dx \, dy \quad \checkmark$$

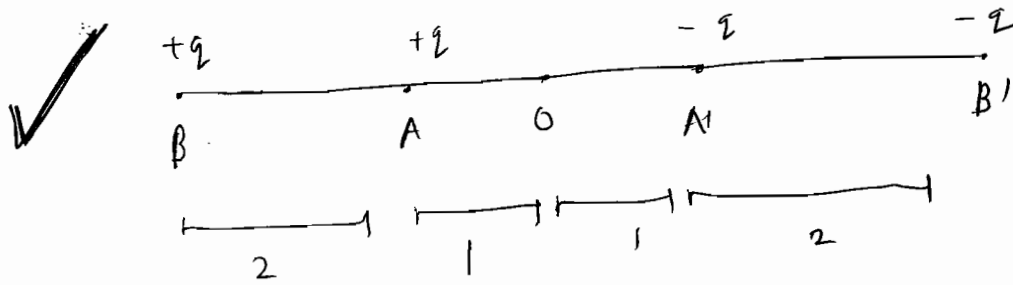
Q1)



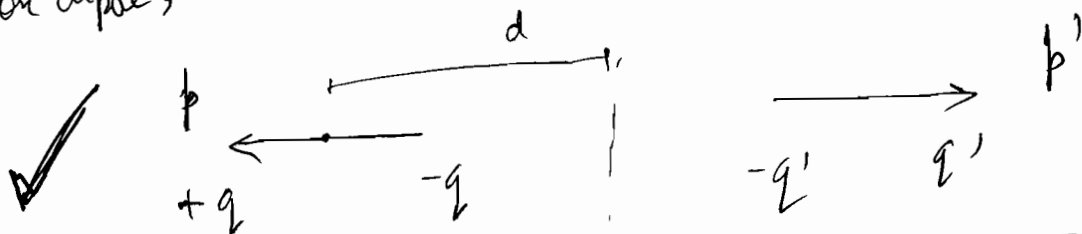
OA = 1m    AB = 2m

$q_A = q_B = 2 \times 10^{-3} \text{ C}$

Now the problem is equivalent to.



Q2) For dipole,



$$U = \frac{1}{4\pi\epsilon_0} \left[ \frac{p_1 \cdot p_2}{r^3} - \frac{3(p_1 \cdot r)(p_2 \cdot r)}{r^5} \right] \quad \boxed{r = 2d}$$

⊛ There are two points worth noting in Laplace's equation:  $\nabla^2 V = 0$

①  $V(x)$  is the average of  $V(x+a)$  and  $V(x-a)$   
i.e. Laplace equation is a kind of averaging instruction, it tells us to assign to the point  $x$ , the average of the values to the left and right of  $x$ .

② Laplace equation tolerates no local maxima or minima, extreme values of  $V$  must be at the end points.

< These validations are for any dimension  $(x, y)$  or  $(x, y, z)$  >

⊛ Solutions of Laplace equations are called Harmonic functions.



This cannot be potential function as it has a local minima.

Laplace equation will pick the most featureless function possible. eg. If you put a ping-pong ball on a stretched rubber sheet (following the potential function), it will roll over to one side and fall off - it will not find a "pocket" somewhere to settle into, for Laplace equation allows no such dents. eg. just as a straight line is the shortest distance between two points, so a harmonic function in 2-d minimizes the surface area spanning the given boundary line.

(Note that we are talking of Laplace solutions (and not Poisson's))

⊛ From these observations, <sup>we</sup> ~~get~~ EARN SHAW'S THEOREM:

[A charged particle cannot be held in a stable equilibrium by electrostatic field alone.]

This is because stable equilibrium requires a valley type of Potential function.

∴ electrostatic forces cannot balance nuclear fusion in Tokamak.

Fortunately it's possible to confine plasma magnetically.

## Uniqueness Theorem

Laplace equation requires a suitable set of boundary conditions to determine  $V$ . But what are suitable boundary conditions, sufficient to determine the potential and yet not so strong as to generate inconsistencies.

If 2- or 3 dimensions partial equations are given, it's not easy to say what would constitute acceptable boundary conditions. The proof that a proposed set of boundary conditions will suffice is usually presented in the form of a Uniqueness theorem. There are two basic uniqueness theorems:

① The solution to Laplace's equation in some volume  $T$  is uniquely determined if potential  $V$  is specified on the boundary surface of  $T$ .

Proof let two solutions  $V_1$  and  $V_2$  satisfying

$$\nabla^2 V_1 = 0, \quad \nabla^2 V_2 = 0$$

Consider  $V_3 = V_1 - V_2$

$$\Rightarrow \nabla^2 V_3 = \nabla^2 V_1 - \nabla^2 V_2 = 0 - 0 = 0 \quad (\text{Laplace eqn})$$

Now at boundary  $V_1$  and  $V_2$  are equal and specified

$$\Rightarrow V_3 = 0 \quad \text{at boundary surface}$$

But Laplace equation allows no local maxima or minima - all extrema occur at the boundary so the maxima & minima of  $V_3$  are both 0

$$\Rightarrow V_3 \text{ is } 0 \text{ everywhere}$$

$$\Rightarrow \boxed{V_1 = V_2}$$

QED !!

## Application

Potential inside an enclosure, having no charges, and completely surrounded by conducting material: Potential on cavity wall is same  $V_0$

One Potential function that satisfies Laplace equation and has constant value  $V_0$  at the boundary is

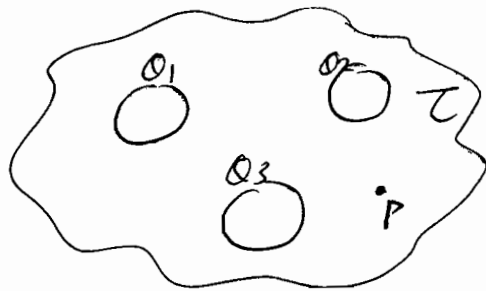
$$V = V_0$$

The uniqueness theorem guarantees, that this is the only solution.

Also imagine a solution that satisfies (a) Laplace equation (b) has correct value at boundaries

This is the right solution.

- ② In a volume  $T$  surrounded by conductors and containing a specified charge density  $\rho$ , the electric field is uniquely determined if total charge on each is specified (conductor)

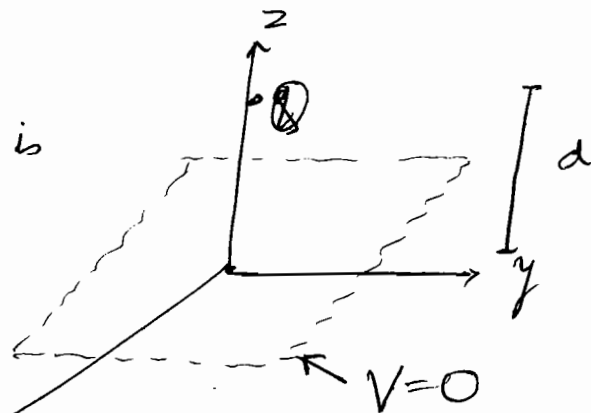


## Method of Images

- ① Suppose a point charge  $Q$  is held a distance  $d$  above an infinite grounded conducting plane. For the region above the plane,  $V$  is not just

$$\frac{1}{4\pi\epsilon_0} \left( \frac{Q}{r} \right) \because q \text{ is induced on } x$$

Some negative charge in conductor which we don't know is how much or how distributed.



From a mathematical point of view, our problem is to solve Poisson's equation in region  $z > 0$ , with a single point charge  $q$  at  $(0, 0, d)$ , subject to boundary condition:

- (i)  $V=0$  when  $z=0$
- (ii)  $V \rightarrow 0$  where  $r \rightarrow \infty$

Corollary to 1<sup>st</sup> Uniqueness theorem

If there is some charge density  $\rho$

$$\nabla^2 V_1 = \frac{-\rho}{\epsilon_0}$$

$$\nabla^2 V_2 = \frac{-\rho}{\epsilon_0}$$

$$\nabla^2 (V_1 - V_2) = 0$$

$$\nabla^2 V_3 = 0$$

⊛ Note that properties of Laplace (no maxima/minima) apply on  $V_3$  only & not on  $V_1$  and  $V_2$ .

Now  $V_3$  satisfied Laplace equation & has value zero on all boundaries, so  $V_3 = 0 \Rightarrow V_1 = V_2$

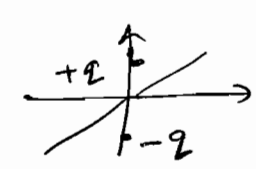
Corollary: Potential in a volume  $T$  is uniquely determined if (a) charge density throughout the region is specified (b) value of  $V$  on all boundary is specified.

$\therefore$  1<sup>st</sup> uniqueness theorem valid for Poisson eqn too

$\therefore$  from corollary of 1<sup>st</sup> Uniqueness theorem, there is only 1 function that meets these requirements.

If by a trick, we can discover such a function, then its got to be the right answer.

Trick study a completely different problem





It follows that

- ①  $V=0$  when  $z=0$
- ②  $V=0$  when  $r \rightarrow \infty$

which are the same conditions as previous problem.

$$\therefore V(x, y, z) = \frac{1}{4\pi\epsilon} \left[ \frac{q}{\sqrt{x^2 + y^2 + (z-d)^2}} - \frac{q}{\sqrt{x^2 + y^2 + (z+d)^2}} \right]$$

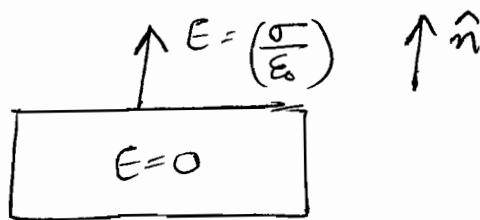
in the upper region  $z \geq 0$

[Note that in lower region  $V_1 \neq V_2$  but we don't care]

Induced Surface Charge

Conductor  $\vec{E} = \frac{\sigma}{\epsilon_0} \hat{n}$

$$\sigma = E \epsilon_0 = -\epsilon_0 \left( \frac{\partial V}{\partial n} \right) = -\epsilon_0 \left( \frac{\partial V}{\partial z} \right) \Big|_{z=0}$$



$$\Rightarrow \underline{\underline{\sigma(x, y) = \frac{-q d}{2\pi [x^2 + y^2 + d^2]^{3/2}} = \frac{-q d}{2\pi (x^2 + y^2 + d^2)^{3/2}}}}$$

$$\Rightarrow \underline{\underline{\text{Total Induced Charge } Q = \int \sigma \, da = \int_0^{2\pi} \int_0^{\infty} \sigma \, r \, dr \, d\phi = -q}}$$

important to do by Polar Coordinates

Force & Energy

Since the potential in the vicinity of  $q$  is the same as that of analogous problem, so the the field and  $\therefore$  the force is also same

$$\Rightarrow F = -\frac{1}{4\pi\epsilon_0} \cdot \frac{q^2}{(2d)^2} \hat{z}$$

Note the plane is grounded. So no free charge <sup>4.4</sup> apart from  $-q$  that is added by charge  $q$ . the remaining  $+q$  charge goes to ground.

There field below is zero, as  $\sigma_{\text{below}} = 0$

Free charge in a grounded conductor flows to the Earth.

$\Rightarrow$  Only half of the energy is there in this problem as compared to analogous problem.

$$W_{\text{actual}} = -\frac{1}{4\pi\epsilon_0} \frac{q^2}{4d} = \frac{1}{2} W_{\text{analogous}}$$

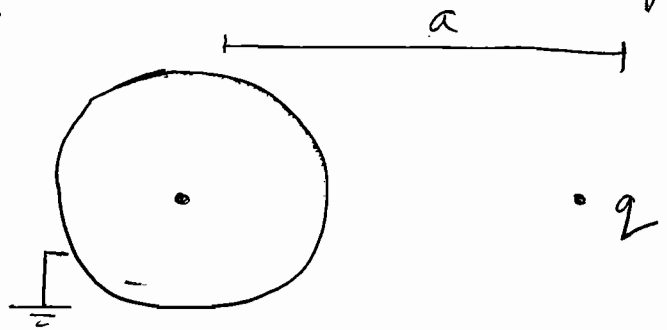
$$W_{\text{analogous}} = -\frac{1}{4\pi\epsilon_0} \frac{q^2}{2d}$$

$W_{\text{actual}}$  can be calculated as

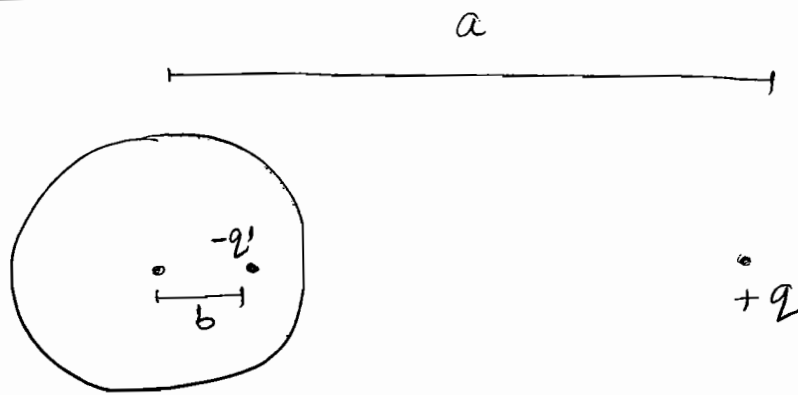
$$\begin{aligned} W &= \int_0^d F_{\text{ext.}} \cdot dx \\ &= \int_0^d \frac{1}{4\pi\epsilon_0} \frac{q^2}{(2z)^2} dz \\ &= \frac{1}{4\pi\epsilon_0} \cdot \frac{q^2}{4} \left[ \frac{1}{\infty} - \frac{1}{d} \right] = \underline{\underline{-\frac{1}{4\pi\epsilon_0} \left( \frac{q^2}{4d} \right)}} \end{aligned}$$

\* Note that this method is not limited to a single point charge; any stationary charge dist<sup>n</sup> near a grounded conducting plane can be treated in the same way by placing opposite charges at mirror image s.t.  $V=0$  at the plane.

(2) Aim is to find out Potential outside the sphere.



# Analogous Problem



$$-q' = \left(\frac{R}{a}\right) q$$

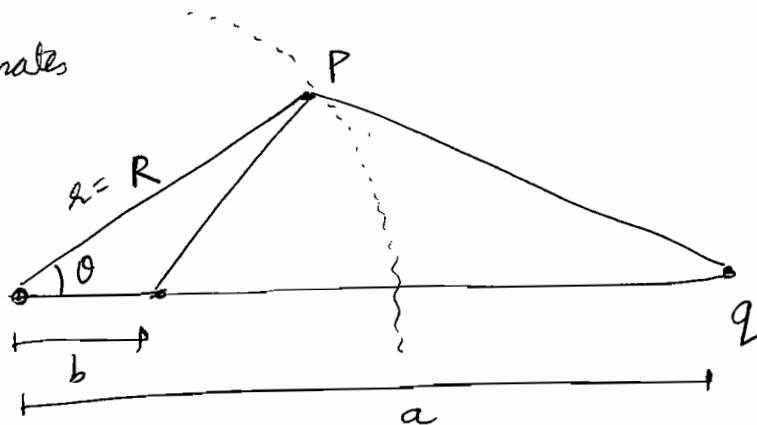
$$b = \left(\frac{R^2}{a}\right)$$

• "a" in denominator in both the cases.

दोनों एक दूसरे से दूर जाएँगे "a" के बढ़ने पर !!

Note that  $b < R$ . So the image charge is safely inside the sphere; we cannot put image charges in the region where we are calculating the  $V$ ; that would change  $\rho$  and we will solve Poisson equation with wrong  $\rho$ .

Use Polar Coordinates



$$V_P = \frac{1}{4\pi\epsilon_0} \left[ \frac{q}{\sqrt{R^2 + a^2 - 2Ra \cos\theta}} - \frac{qR}{a \sqrt{b^2 + R^2 - 2bR \cos\theta}} \right]$$

$$= \frac{1}{4\pi\epsilon_0} \left[ \frac{q}{\sqrt{R^2 + a^2 - 2Ra \cos\theta}} - \frac{q}{\sqrt{\frac{a^2 \cdot R^4}{R^2 a^2} + R^2 \frac{a^2}{R^2} - 2 \frac{R^2 \cdot Ra^2 \cos\theta}{a R^2}}} \right]$$

$$= \frac{1}{4\pi\epsilon_0} \left[ \frac{q}{\sqrt{R^2 + a^2 - 2aR \cos\theta}} - \frac{q}{\sqrt{R^2 + a^2 - 2aR \cos\theta}} \right]$$

= 0  
Use similar procedure to calculate  $V(r)$  by using  $r$  in place of  $R$

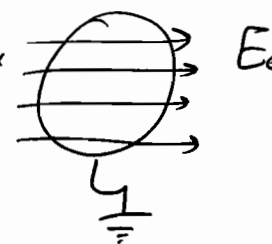
$$\vec{\nabla} \cdot \vec{D} = \rho$$

$$\vec{D} = \epsilon \vec{E}$$

$$\vec{\nabla} \cdot \vec{E} = \frac{\rho}{\epsilon}$$

$$\nabla^2 V = -\frac{\rho}{\epsilon}$$

25) Coulumb  $\rightarrow$  Gauss  $\rightarrow$   $\nabla \cdot E$  form  $\rightarrow$  Laplace

$$V(r, \theta) = -E_0 r \cos \theta \left(1 - \frac{a^3}{R^3}\right)$$


32)  $\vec{P} = n \vec{p}$

2nd standard problem is : Conducting Sphere and Point Charge

2011 Question

$$V(r, \theta) = -E_0 r \cos \theta \left(1 - \frac{a^3}{R^3}\right)$$

$$\vec{E} = -\vec{\nabla} V$$

$$E_r = -\left(\frac{\partial V}{\partial r}\right)$$

We know,

$$E = \frac{\sigma}{\epsilon_0} \hat{r}$$

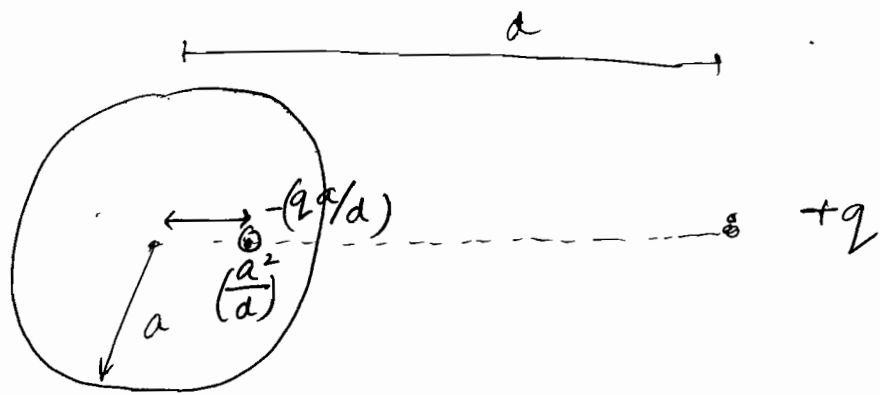
$$-\left(\frac{\partial V}{\partial r}\right) = \frac{\sigma}{\epsilon_0}$$

← Do not go into detail of  $\sigma_{\text{sphere}}$  !  
Physicist IAS बनना है, नही!

$$\sigma = -\frac{q}{4\pi R^2} (a^2 - R^2) (R^2 + a^2 - 2aR \cos \theta)$$

$$Q_{\text{induced}} = -\frac{qR}{a} = q' \checkmark$$

(2) Sphere



$$V(r=a) = 0$$

Replace the sphere by another point charge  $-q'$  equal to  $-\left(\frac{qa}{d}\right)$  placed at  $\left(\frac{a^2}{d}\right)$  from centre.

$$\left[ \frac{q}{d-a} + \frac{q'}{a-b} \right] = 0$$

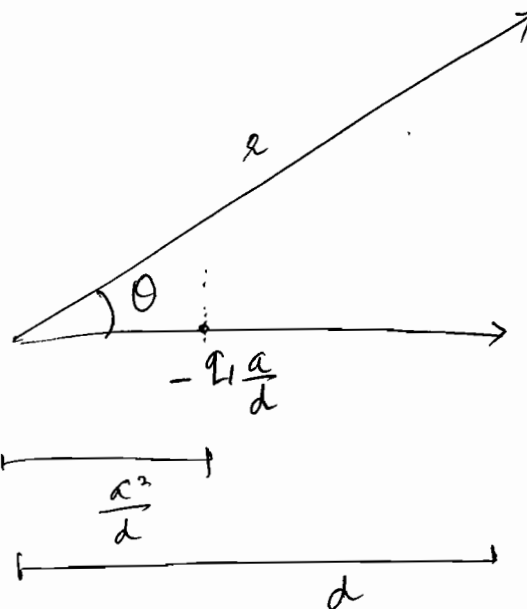
$$b = \frac{a^2}{d}$$

$$q' = -\frac{qa}{d}$$

$$\Rightarrow \frac{q}{da} = \frac{qa}{d(a-b)}$$

$$\Rightarrow da - db = da - a^2$$

$$\Rightarrow \boxed{b = \frac{a^2}{d}}$$



$$V(r, \theta) = \frac{1}{4\pi\epsilon_0} \left[ \frac{q}{\sqrt{r^2 + d^2 - 2rd\cos\theta}} - \frac{(qa/d)}{\sqrt{r^2 + (a^2/d)^2 - 2ra^2/d\cos\theta}} \right]$$

$$\vec{E} = -\vec{\nabla}V$$

$$E_r = \left( \frac{\partial V}{\partial r} \right)_{r=a} = \left( \frac{\sigma}{\epsilon_0} \right)$$

$$Q = \int_S \sigma dA$$

○ If not grounded  
then how to solve  
by method of images?



• q

Place a charge  $= -q'$  at centre of sphere.

Net charge inside conducting sphere = 0

Note that in previous question, earthing की  
द्वारा Net charge inside = 0

# E&M (8)

21/01/2012

2<sup>nd</sup> Method

$$E = \frac{1}{2} \epsilon_0 \int_{\text{space}} E^2 d\tau$$

$$= \frac{\epsilon_0}{2} \int \frac{Q^2}{(4\pi\epsilon_0)^2} \frac{r^2}{R^6} d\tau + \frac{\epsilon_0}{2} \int \frac{Q^2}{(4\pi\epsilon_0)^2} \frac{1}{r^4} d\tau$$

Now  $d\tau = 4\pi r^2 dr$

$$\Rightarrow E = \frac{3}{5} \frac{Q^2}{(4\pi\epsilon_0) R}$$

Self Energy

1<sup>st</sup> Method

$$\begin{aligned} &= \int V dq \\ &= \int_0^R \frac{1}{4\pi\epsilon_0} \cdot \frac{\rho \cdot \frac{4}{3}\pi r^3}{r} \cdot 4\pi r^2 dr \\ &= \frac{3}{20} \frac{Q^2}{\pi\epsilon_0 R} \end{aligned}$$

$$= \frac{3}{5} \cdot \left( \frac{1}{4\pi\epsilon_0} \right) \frac{Q^2}{R}$$

Note similarity between it

And  $\frac{3}{5} \frac{G M^2}{R}$

Using Gauss law, we can calculate

$V(r < R)$  inside sphere of charge  $q$

$$= \frac{1}{4\pi\epsilon_0} \frac{q}{R} \left( \frac{3}{2} - \frac{r^2}{R^2} \right)$$

Using this we can find out, self energy

$$W = \int dw = \int V dq = \int V \rho d\tau$$

$$= \int \frac{1}{4\pi\epsilon_0} \frac{q}{R} \cdot \frac{q}{4\pi R^3} \cdot 4\pi r^2 dr \left( \frac{3}{2} - \frac{r^2}{R^2} \right)$$

$$= \int \frac{1}{4\pi\epsilon_0} \frac{q}{R} \cdot \frac{3q}{R} \left( \frac{3}{2} - \frac{r^2}{R^2} \right)$$

$$= \frac{3q^2}{4\pi\epsilon_0 R^2} \left[ \frac{3}{2} R - \frac{R^3}{R^2} \right]$$

$$= \frac{3q^2}{4\pi\epsilon_0 R^2} \cdot \frac{1}{2} R = \left( \frac{3Q^2}{8\pi\epsilon_0 R} \right)$$

$$= \int \frac{1}{4\pi\epsilon_0} \frac{3q^2}{R^4} \cdot \left[ \frac{3}{2} r^2 - \frac{r^4}{R^2} \right] dr$$

$$= \frac{3q^2}{\pi\epsilon_0 R^4} \cdot \left[ \frac{3}{2} \frac{R^3}{3} - \frac{R^5}{5R^2} \right]$$

$$= \frac{3q^2}{\pi\epsilon_0 R} \left[ \frac{1}{2} - \frac{1}{5} \right] = \underline{\underline{\frac{3}{20} \frac{Q^2}{\pi\epsilon_0 R}}}$$


# ★ Lorentz Force


# MAGNETOSTATICS

5.1

✓  $\vec{F}_{\text{mag}} = q(\vec{v} \times \vec{B})$  Lorentz Force law

○ if  $v=0$ , no  $\vec{F}_{\text{mag}} \Rightarrow$  stationary charges experience 0 force

○  Attractive Force if current in same direction

○  Repulsive Force if current in opposite direction

✓ In the presence of both electric & magnetic fields,

$$\vec{F} = q(\vec{v} \times \vec{B}) + q\vec{E}$$

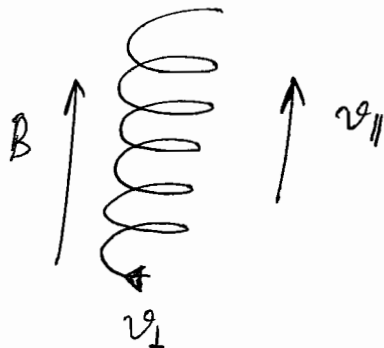
Note that these are empirical laws.

✓ Typical motion under magnetic field is circular with radius as  $\left(\frac{mv_{\perp}}{qB}\right)$

This is the motion in cyclotron - the first of the modern particle accelerators.

Its also a measure of momentum of a particle by measurement of radius.

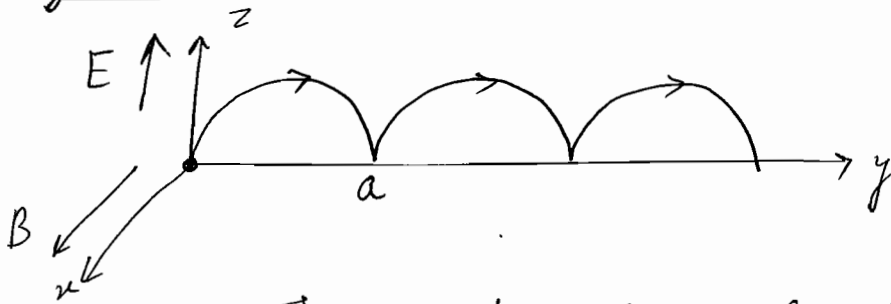
If it also has  $v_{\parallel}$  component, then helical motion





# Cycloid Motion

$m > q$



If  $\vec{E}$  and  $\vec{B}$  are at right angles and initial particle is at rest at origin. The path followed will be a cycloid.

Initially magnetic force = 0 and electric field accelerates the charge in z-direction. As it picks up speed, the magnetic force pulls the charge to the right. The faster it goes, stronger the magnetic force becomes, eventually it curves the particle back towards y-axis. At this point, charge moves against ~~for~~ electric force and therefore decelerates, magnetic force decreases, and electric force takes over, bringing the charge to rest at point a. Now the entire process repeats.

Equation of motions comes out :

$$(y - R\omega t)^2 + (z - R)^2 = R^2$$

$$R = \frac{E}{\omega B}$$

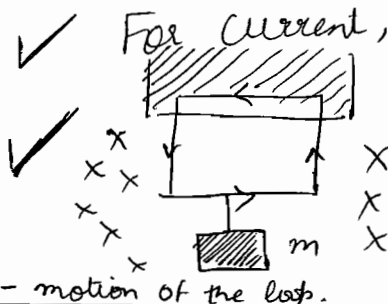
$$\omega = \frac{qB}{m}$$

Note that overall motion is  $\perp$  to  $\vec{E}$  and  $\vec{B}$ .

✓ Magnetic Forces do no work.

Even when Magnetic cranes lift the junk car, still magnetic field does zero work. In every case, when work is perceived to have been done, there is some other agency that does the work.

For current, magnetic force  $F_{mag} = \int I (d\vec{l} \times \vec{B})$



say Magnetic Force > Gravitational Force  $\Rightarrow$  mass m rises up. The work is done by Battery and not magnetic field. Magnetic Force simply redirects the horizontal force of battery into vertical.

✓ ⊛ 1 Ampere = 1 Coulomb / second

⊛ Surface Current Density:  $\vec{k}$

$k = \frac{dI}{dl_{\perp}}$  : Current per unit width perpendicular to the flow

If (mobile) surface charge density is  $\sigma$  and the velocity is  $v$ .

$dI = \left(\frac{dq}{dt}\right) = \sigma \cdot dl_{\perp} \left(\frac{dx}{dt}\right)$



$\Rightarrow \left(\frac{dI}{dl_{\perp}}\right) = \sigma v$

$\Rightarrow \vec{k} = \sigma \vec{v}$

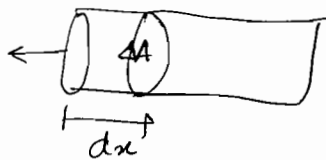
$\vec{F}_{magnetic} = \int dq (\vec{v} \times \vec{B}) = \int \sigma da (\vec{v} \times \vec{B})$

$\vec{F}_{mag} = \int (\vec{k} \times \vec{B}) da$

⊛ Volume Current Density

$J = \frac{dI}{da_{\perp}}$  : Current per unit area perpendicular to the flow.

$dI = \left(\frac{dq}{dt}\right) = \rho(AA) \frac{dx}{dt}$



$\Rightarrow \frac{dI}{dA} = \rho v \Rightarrow \vec{J} = \rho \vec{v}$

Ohm's law  $\vec{J} = \sigma \vec{E}$

$\vec{F}_{mag} = \int dq (\vec{v} \times \vec{B}) = \int \rho \cdot d\tau \cdot (\vec{v} \times \vec{B}) = \int (\vec{J} \times \vec{B}) d\tau$

$\Rightarrow \vec{F}_{mag} = \int (\vec{J} \times \vec{B}) d\tau$

## Continuity Equation

Now current crossing a surface  $S$  can be written as

$$I = \int_S \mathbf{J} \cdot d\mathbf{a}_\perp = \int_S \vec{J} \cdot d\vec{a}$$

Total charge leaving the volume per unit time is given by

$$\oint_S \vec{J} \cdot d\vec{a} = \int_{\text{Volume}} (\nabla \cdot \vec{J}) d\tau$$

Since the charge is conserved, whatever flow out must come at the expense of that remaining inside.

$$\Rightarrow \int_{\text{Volume}} \nabla \cdot \vec{J} d\tau = -\frac{d}{dt} q_{\text{inside}} = -\frac{d}{dt} \int \rho d\tau$$

$$\Rightarrow \int \nabla \cdot \vec{J} d\tau = \int -\left(\frac{\partial \rho}{\partial t}\right) d\tau$$

$$\Rightarrow \boxed{\nabla \cdot \vec{J} = -\left(\frac{\partial \rho}{\partial t}\right)}$$

## Biot-Savart Law

$$\mathbf{B}(\mathbf{r}) = \frac{\mu_0}{4\pi} \int \frac{I d\vec{\ell} \times \vec{r}}{r^3}$$

for a steady current  $I$ .

Note that stationary charges produce electric fields that are constant in time, hence the term electrostatics.

Steady currents produce magnetic fields that are constant in time, hence the term magnetostatics.

For most practical applications, magnetostatics apply. Even for current (households) that varies 50 times a second.

When a steady current flows in a wire, its magnitude remains same all throughout, otherwise charge would be piling up somewhere, and it won't be a steady current. By the same token,  $(\frac{\partial \rho}{\partial t}) = 0$  in magnetostatics, and hence the continuity equation becomes

$$\boxed{\nabla \cdot \vec{J} = 0} \text{ for Magnetostatics.}$$

Note that a moving point charge cannot possibly constitute a steady current

$$\therefore B(\vec{r}) = \frac{\mu_0}{4\pi} q \left( \frac{\vec{v} \times \vec{r}}{r^3} \right) \text{ is WRONG.}$$

as Biot Savart's law holds only for steady currents.

Like Electrostatics, superposition principle applies to Magnetostatics also. If we have a collection of source currents, the net magnetic field is the vector sum of the field due to each of them taken separately.

$$\textcircled{*} \quad I d\vec{l} \Leftrightarrow \vec{k} \cdot d\vec{a} \Leftrightarrow \vec{J} d\vec{a} \Leftrightarrow q\vec{v}$$

### Divergence & Curl of B

By some mathematics upon Biot Savart's law, we can get

$$\boxed{\nabla \cdot \vec{B} = 0}$$

$$\boxed{\nabla \times \vec{B} = \mu_0 \vec{J}}$$

Other form of curl of  $\vec{B}$ , is obtained by using Stokes theorem. This is called AMPERE'S LAW.

$$\int (\nabla \times \vec{B}) \cdot d\vec{a} = \mu_0 \int \vec{J} \cdot d\vec{a} = \mu_0 I_{enc} \Rightarrow \boxed{\oint \vec{B} \cdot d\vec{l} = \mu_0 I_{enc}}$$

The direction of current taken as  $\oplus$ ve should be concomitant with the direction of coverage of loop according to right hand screw rule.

Gauss law in Electrostatics  $\Leftrightarrow$  Ampere's law in Magnetostatics

✓ Maxwell Equations for electrostatics :  $\nabla \times \vec{E} = 0$   
 (Gauss law)  $\nabla \cdot \vec{E} = \left( \frac{\rho}{\epsilon_0} \right)$

Maxwell Equations for magnetostatics :  $\nabla \times \vec{B} = \mu_0 \vec{J}$   
 (Ampere's law)  $\nabla \cdot \vec{B} = 0$

✓  $\nabla \cdot \vec{B} = 0$  : It implies that there are no divergence or convergence of magnetic fields; they simply form closed loops or extend out to infinity. There are no point sources for  $\vec{B}$  i.e. No Magnetic Monopoles exist.

★ Typically electric forces are enormously larger than magnetic forces.

### Magnetic Vector Potential

Since  $\nabla \cdot \vec{B} = 0$ , we can define a vector potential  $\vec{A}$  such that  $\boxed{\vec{B} = \nabla \times \vec{A}}$  and  $\boxed{\nabla \cdot \vec{A} = 0}$

We can state Ampere's law, in terms of  $\vec{A}$  as

$$\begin{aligned} \nabla \times \vec{B} &= \nabla \times (\nabla \times \vec{A}) = \nabla (\nabla \cdot \vec{A}) - \vec{A} (\nabla \cdot \nabla) \\ &= -\nabla^2 \vec{A} = \mu_0 \vec{J} \end{aligned}$$

$\nabla^2$ : scalar on scalar  
vector on vector

$$\boxed{\nabla^2 \vec{A} = -\mu_0 \vec{J}}$$

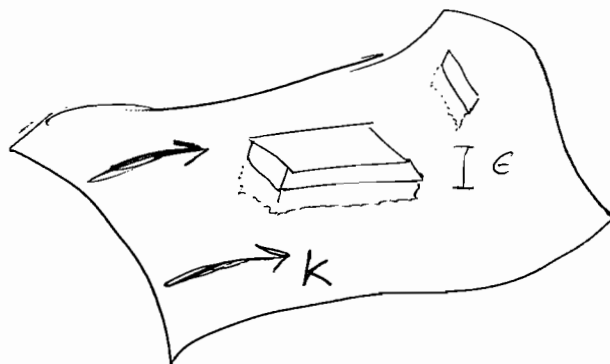
Ordinarily the direction of  $\vec{A}$  is the direction of current

## Boundary Conditions for $\vec{B}$

5.4

$$\oint \vec{B} \cdot d\vec{a} = 0$$

$$B_{\text{above}}^{\perp} = B_{\text{below}}^{\perp}$$



$$\oint \vec{B} \cdot d\vec{l} = (B_{\text{above}}^{\parallel} - B_{\text{below}}^{\parallel})l = \mu_0 I_{\text{enc}} = \mu_0 k l$$

$$\Rightarrow B_{\text{above}}^{\parallel} - B_{\text{below}}^{\parallel} = \mu_0 k$$

Thus, the component of  $\vec{B}$  that is parallel to the surface but perpendicular to the current is discontinuous in the amount  $\mu_0 k$ .

$$\Rightarrow \vec{B}_{\text{above}} - \vec{B}_{\text{below}} = \mu_0 (\vec{k} \times \hat{n})$$

where  $\hat{n}$  is a unit vector perpendicular to the surface, pointing upwards.

Just as the electric field is discontinuous at a surface charge, so the magnetic field is discontinuous at a surface current. Only this time, it's the tangential component that changes.

# MAGNETIC FIELDS IN MATTER

All magnetic phenomenon are due to electric charges in motion.

In magnetic materials, on an atomic scale, there are tiny currents, electrons orbiting around nuclei and electrons spinning.

For macroscopic purposes, these current loops are so small that we may treat them as magnetic dipoles. Ordinarily they cancel out each other due to random orientations of these dipoles, but when magnetic field is applied, a net alignment of these dipoles occur, and the material becomes magnetically polarized or "magnetized".

Unlike electric polarization, which is almost always in the direction of  $\vec{E}$ , some materials acquire magnetization parallel to  $\vec{B}$  (paramagnetic) and some opposite to  $\vec{B}$ . (diamagnetic). A few substances (ferromagnetic) retain their magnetization even after external field has been removed. These magnetizations are not determined by the present field but by the whole magnetic "history" of the object.

## Torque on a magnetic dipole

$$\vec{\tau} = \vec{m} \times \vec{B}$$

$$\text{where } \vec{m} = I \vec{a}$$

These torque try to align magnetic dipoles parallel to the magnetic field and therefore account for paramagnetism. But due to Pauli Exclusion Principle, only unpaired electrons show paramagnetism.

In a uniform field, force on a current loop is zero. but in a non-uniform field, this is no longer the case.

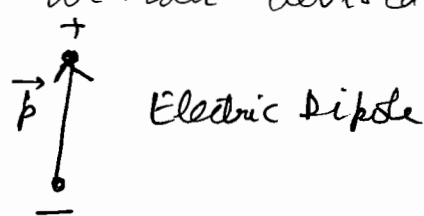
## Force on a dipole in Non-Uniform Field

## Electric Dipole

5.5

Remember that for an electric dipole, we had derived,

$$F = (\vec{p} \cdot \vec{\nabla}) \vec{E} \quad \text{--- (1)}$$



Now consider the argument,

$$U_{\text{dipole}} = -\vec{p} \cdot \vec{E}$$

$$\vec{F} = -\vec{\nabla} U = -\vec{\nabla}(-\vec{p} \cdot \vec{E}) = \vec{\nabla}(\vec{p} \cdot \vec{E}) \quad \text{--- (2)}$$

Now for electric field,

$$\begin{aligned} \vec{\nabla} \cdot (\vec{p} \cdot \vec{E}) &= \vec{p} \times (\vec{\nabla} \times \vec{E}) + \vec{E} \times (\vec{\nabla} \times \vec{p}) + (\vec{p} \cdot \vec{\nabla}) \vec{E} + (\vec{E} \cdot \vec{\nabla}) \vec{p} \\ &= 0 + 0 + (\vec{p} \cdot \vec{\nabla}) \vec{E} + 0 \\ &= (\vec{p} \cdot \vec{\nabla}) \vec{E} \end{aligned}$$

Hence 1<sup>st</sup> and 2<sup>nd</sup> expressions are same.

Also, Mathematically,

$$(\vec{p} \cdot \vec{\nabla}) \vec{E} = \sum \left[ p_x \left( \frac{\partial E_x}{\partial x} \right) + p_y \left( \frac{\partial E_x}{\partial y} \right) + p_z \left( \frac{\partial E_x}{\partial z} \right) \right] \hat{i}$$

$$\vec{\nabla} \cdot (\vec{p} \cdot \vec{E}) = \sum \left[ p_x \left( \frac{\partial E_x}{\partial x} \right) + p_y \left( \frac{\partial E_y}{\partial x} \right) + p_z \left( \frac{\partial E_z}{\partial x} \right) \right] \hat{i}$$

From  $\vec{\nabla} \times \vec{E} = 0$ , we know

$$\frac{\partial E_z}{\partial y} = \frac{\partial E_y}{\partial z}$$

$$\frac{\partial E_x}{\partial y} = \frac{\partial E_y}{\partial x}$$

$$\frac{\partial E_x}{\partial z} = \frac{\partial E_z}{\partial x}$$

$\therefore$  the two are same.

Preferable way of writing is  $(\vec{p} \cdot \vec{\nabla}) \vec{E}$



# Magnetic Dipole

since  $\nabla \times \vec{B} \neq 0$ , we cannot write in the preferable form

$$\therefore \boxed{F = \nabla (\vec{m} \cdot \vec{B})}$$



As we can see, the gilbert model, though incorrect ( $\because$  there are no magnetic monopoles), gives approximation to the true Ampere Model and can be applied in many situations. Application of gilbert Model gives the advantage of using results of electric dipole by replacing of

$$\left. \begin{array}{l} \vec{p} \text{ by } \vec{m} \\ (1/\epsilon_0) \text{ by } \mu_0 \end{array} \right\} \begin{array}{l} \text{H.C. Verma} \\ \text{ch1 chapter} \\ \text{part 12 based } \epsilon^2 !! \end{array}$$

Note again that there is no such thing as single magnetic North Pole or South Pole. If you break a bar magnet in half, you don't get a North Pole in one hand and south pole in ~~other~~; you get two complete magnets. Magnetism is not due to magnetic monopoles but due to moving electric charges. Magnetic dipoles are tiny current loops and it's an extraordinary thing, really, that the formulas involving  $\vec{m}$  bear any resemblance at all to the corresponding formula for  $\vec{p}$ .

## ⊛ Magnetization

Whatever be the cause of magnetization, we describe the state of magnetic polarization by vector quantity  $\vec{M}$  called Magnetization.

$\vec{M}$ : magnetic dipole moment per unit volume.

In general when a sample is placed in a region of non-uniform field, a paramagnet is attracted into the field while the diamagnet is repelled away.

# Ampere's law for magnetized materials

$$\vec{\nabla} \times \vec{B} = \mu_0 \vec{J}_{\text{total}} \quad (\text{Ampere's law})$$

$$\text{Now } \vec{J}_{\text{total}} = \vec{J}_{\text{free}} + \vec{J}_{\text{bound}}$$

$$\text{and } \vec{J}_{\text{bound}} = \vec{\nabla} \times \vec{M}$$

[Note that similar derivation can be done for Vector Magnetic Potential to get

$$\vec{J}_{\text{bound}} = \vec{\nabla} \times \vec{M}$$

$$\vec{K}_{\text{bound}} = \vec{M} \times \hat{n}$$

L

[ Replace  $\vec{J}$  by  $\vec{J}$   
 $\sigma$  by  $\vec{K}$   
( $\cdot$ ) by ( $\times$ )  
and make all signs  $\oplus$ ve  
 $\vec{p}$  by  $\vec{M}$  ]

$$\Rightarrow \vec{\nabla} \times \left[ \frac{\vec{B}}{\mu_0} - \vec{M} \right] = \vec{J}_{\text{free}}$$

$$\text{Define } \frac{\vec{B}}{\mu_0} - \vec{M} = \vec{H}$$

$$\Rightarrow \boxed{\vec{\nabla} \times \vec{H} = \vec{J}_{\text{free}}}$$

$$\boxed{\oint \vec{H} \cdot d\vec{l} = I_{\text{enc}}}$$

(\*) Many authors call H as Magnetic Field  
B as Flux Density

## Boundary Conditions

$$H_{\text{above}}^{\perp} - H_{\text{below}}^{\perp} = - (M_{\text{above}}^{\perp} - M_{\text{below}}^{\perp}) \quad (\text{from def'n})$$

$$H_{\text{above}}^{\parallel} - H_{\text{below}}^{\parallel} = \vec{K}_f \times \hat{n} \quad (\text{from Ampere's law})$$

# Linear Media

$$\vec{M} = \chi_m \vec{H}$$

$\chi_m$ : magnetic susceptibility

⊕ve for paramagnets  
⊖ve for diamagnets

Now  $\vec{B} = \mu_0 (\vec{H} + \vec{M})$

⇒ for linear media  $\vec{B} = \mu_0 [\vec{H} + \chi_m \vec{H}]$

$$\Rightarrow \vec{B} = \mu_0 (1 + \chi_m) \vec{H}$$

$$\mu = \mu_0 (1 + \chi_m)$$

$$\mu = \mu_0 \mu_r$$

: permeability of the material

:  $\mu_r$ : relative permeability

For space  $\vec{B} = \mu_0 \vec{H}$

Also for linear ~~media~~ magnetic media,

$$\vec{J}_{\text{bound}} = \nabla \times \vec{M} = \nabla \times \chi_m \vec{H} = \chi_m \vec{J}_{\text{free}}$$

$$\vec{J}_{\text{bound}} = \chi_m \vec{J}_{\text{free}}$$

i.e. unless a free current flows through the material, all bound current will be at the surface.

# Magnetic Scalar Potential

The differential form of Ampere's law ( $\nabla \times \vec{H} = \vec{J}_f$ ) shows that  $\vec{H}$  is not a conservative field and thus cannot be written in terms of a gradient of a scalar potential. But there are certain regions of space called "Simply Connected" (regions outside of the current sources in which there is no possibility of any closed path being linked with a current), where magnetic field  $\vec{H}$  could be described as a conservative field and can be expressed as the gradient of a scalar potential function

$$\vec{H} = -\nabla\phi_m \quad \& \quad \nabla \times \vec{H} = 0$$

s.t.

We know  $\nabla \cdot \vec{B} = 0 \Rightarrow \nabla \cdot \mu_0(\vec{H} + \vec{M}) = 0 \Rightarrow -\nabla \cdot \vec{H} = \nabla \cdot \vec{M} \Rightarrow \boxed{\nabla^2 \phi_m = -\nabla \cdot \vec{M}}$

From some mathematics, it can be shown that

$$\phi_m = \frac{\mu_0 I}{4\pi} \Omega$$

$\Omega$ : solid angle subtended at the point P by the current loop.

Similar to electric dipole, magnetic dipole also has potential energy

$$U = -\vec{m} \cdot \vec{B}$$

$$\vec{A} = \frac{\mu I}{4\pi} \int \frac{d\vec{\ell}}{r}$$

Correct  
formula

# MAGNETOSTATICS

## Magnetic Property

$\mu$  Permeability

$\mu_r$  Relative Permeability

$$\mu = \mu_r \mu_0$$

Non Magnetic Material

$$\mu_r = 1$$

It gets Magnetized

$$\vec{B} = \mu \vec{H}$$

Applied Magnetic Field

$$\vec{H}$$

$$\begin{aligned} \vec{B} &= \mu \vec{H} \\ &= \mu_0 (\vec{H} + \vec{M}) \end{aligned}$$

$$\Rightarrow (\mu - \mu_0) \vec{H} = \mu_0 \vec{M}$$

$$\Rightarrow \mu_r = 1 + \frac{\vec{M}}{\vec{H}}$$

$$= 1 + \chi$$

$$\chi_m = \frac{M}{H} \checkmark$$

Alignment of dipoles

## Equivalent Electrical Property

$\epsilon$  Permittivity

$\epsilon_r$  Relative Permittivity

$$\epsilon = \epsilon_r \epsilon_0$$

$$\text{Air } \epsilon_r = 1$$

It gets Polarized

$$\vec{D} = \epsilon \vec{E}$$

Applied Electrical Field

$$\vec{E}$$

$$\vec{D} = \epsilon_0 \vec{E} + \vec{P}$$

$$\vec{P} = (\epsilon - \epsilon_0) \vec{E}$$

$$= (k - 1) \epsilon_0 \vec{E}$$

$$(k = \epsilon_r)$$

$$\chi_e = \frac{\vec{P}}{\epsilon_0 \vec{E}} \checkmark$$

Alignment of charges

✓  $\vec{H}$  applied, then  $\vec{M}$  occurs due to alignment of dipoles, then  $\vec{B}$ .

⇒  $\vec{M}$  and  $\vec{B}$  lag behind  $\vec{H}$

⇒ this lagging is called HYSTeresis.

$$\vec{M} = \chi \vec{H}$$

✓ For ferromagnetic materials,

$$\mu \gg \mu_0 \quad \text{i.e. } \chi \gg 1$$

✓ For paramagnetic materials,

$$\mu > \mu_0 \quad \text{i.e. } \chi > 0$$

In ferromagnetic Materials,  $\vec{M}$  and  $\vec{B}$  lag behind  $\vec{H}$ , which leads to hysteresis !!

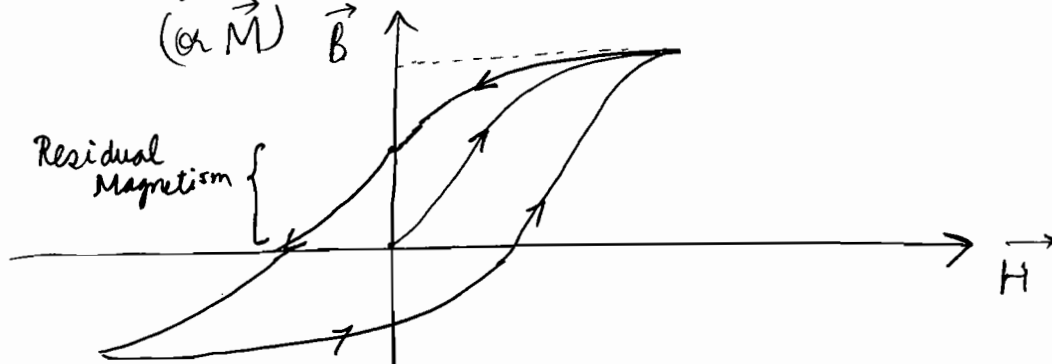
When ferromagnetic material is undergoing cycles of magnetization & demagnetization, there is dissipation of energy due to hysteresis.

$$\text{Loss of Energy } \propto \int H \, d\vec{m}$$

( $\propto \vec{M}$ )  $\vec{B}$

Property is called Retentivity

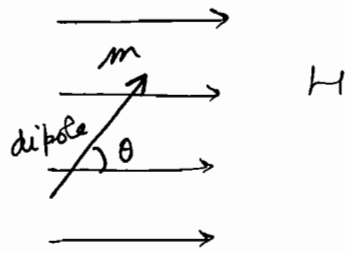
Residual Magnetism



Coercive Force

Property is called Coercivity

Electromagnetics  
[S. Singh]



$$\tau = \vec{m} \times \mu \vec{H}$$

$$= \mu m H \sin \theta$$

$$dW = \tau d\theta$$

$$= \mu m H \sin \theta d\theta$$

$$W = \mu H \int m \sin \theta d\theta$$

$$= \mu H \int -m d(\cos \theta)$$

$$W = \mu H -d \sum m \cos \theta$$

Hysteresis loss per unit volume <sup>per cycle</sup>

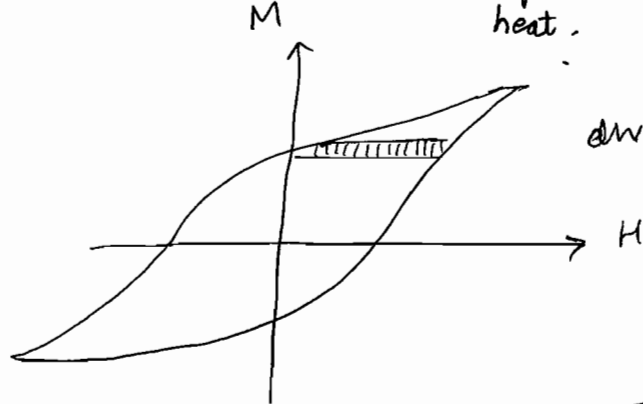
= Area of B-H Hysteresis loop

=  $\oint \vec{H} \cdot d\vec{B}$  =  $\oint H \cdot \mu_0 (dH + dM)$

=  $\mu_0 \oint H dM$  (WARBURG'S LAW)

Its dissipated in the form of heat.

$$\Rightarrow \left(\frac{W}{V}\right) = -\mu \vec{H} \cdot d\vec{M}$$



$$W = \oint \vec{H} \cdot d\vec{M}$$

= Area of loop !!

$\Rightarrow$  Hysteresis loss = [Area of MH loop] \*  $\mu_0$

$$W = \int \vec{H} \cdot \left(\frac{d\vec{B}}{\mu_0} - d\vec{H}\right)$$

$$= \frac{1}{\mu_0} \int \vec{H} \cdot d\vec{B}$$

$\uparrow$   $\left(\oint \vec{H} \cdot d\vec{H} = 0\right)$

(यह क्यों होता है, दिखाता नहीं)

- ✓ Soft Iron : low Area of Hysteresis loop  $\rightarrow$  Transformer
- ✓ Hard Iron : high Area of Hysteresis loop  $\rightarrow$  Permanent Magnet eg. Steel

Also since Work Done = Some constant times  $\mu_0 \oint H dM$ , equivalence used in Adiabatic Demagnetization was valid. Also in CGS units,  $\mu_0 = 1$  and its dimensionless.



## Numerical

$$\text{Mass core} = 10 \text{ kg}$$

$$f = 50 \text{ Hz}$$

$$\text{Time} = 1 \text{ hour}$$

$$\text{Density} = 1.5 \times 10^3 \text{ kg/m}^3$$

$$\text{Area of Hysteresis loop} = 6 \times 10^3 \text{ erg/cc}$$

Find out hourly loss of energy.

Assuming loss is only due to hysteresis:

$$\text{No. of cycles} = 50 \times 60 \times 60$$

$$\text{Volume} = \frac{10}{1.5 \times 10^3}$$

$$\text{Loss} = \text{Area} \times \text{Volume} \times \text{cycles}$$

$$= 6 \times 10^3 \times \frac{10}{1.5 \times 10^3} \times 50 \times 60 \times 60 \times 10^6$$

$$= 4 \times 5 \times 36 \times 10^{10} \text{ ergs}$$

$$= \underline{7.2 \times 10^{12} \text{ ergs}}$$

## Current Electricity

Preliminaries : Kirchhoff's laws :  $\sum I = 0$   
 $\sum V = 0$

Biot's Savart's law

Ampere's law

Faraday laws of

Lenz's law

EM Induction  $\left\{ \begin{array}{l} \text{Self} \\ \text{Induction} \\ \text{Mutual} \\ \text{Induction} \end{array} \right.$

Transient Circuits : LC, RC, LCR

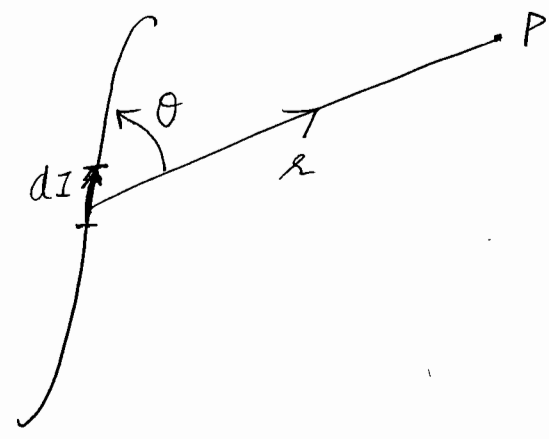
AC Circuits

# Biot Savart's law

⊙ I is scalar as it is not added by vector law of addition  
 But dI (current element) is considered as vector quantity.

Current element has ability to attract.

Larger the distance, lesser the attractive power.



It is noted

$$dB \propto I dl \sin \theta$$

$$dB \propto \frac{I}{r^2}$$

$$dB = \frac{\mu_0}{4\pi} \frac{I dl \sin \theta}{r^2} = \frac{\mu_0}{4\pi} \frac{I dl r \sin \theta}{r^3}$$

$$\boxed{\vec{dB} = \frac{\mu_0}{4\pi} I \left[ \frac{d\vec{l} \times \vec{r}}{r^3} \right]}$$

(empirical law)  
Biot Savart's law

✓ direction of  $\vec{B}$  given by Right Hand Rule for cross products

Let us write  $\boxed{\vec{H} = -\vec{\nabla} \phi_m}$

We can also write

$$\boxed{\vec{B} = \nabla \times \vec{A}}$$

Magnetostatic Vector Potential

Magnetostatic scalar Potential

[I does not vary w.r.t time]

⊙ Remember that  $\phi$  (Scalar Potential) and  $\vec{A}$  (Vector Potential) are just Mathematical Constructs

Take 
$$\phi_m = \frac{\mu_0 I \Omega}{4\pi}$$

Solid Angle subtended by current carrying loop.

✓ Note the every current carrying loop is a dipole.



Here  $\vec{A}$  is the area vector

$$\Omega = \frac{\vec{A} \cdot \vec{r}}{r^3}$$

$$\phi_m = \frac{\mu_0 I}{4\pi} \frac{\vec{A} \cdot \vec{r}}{r^3}$$

$$\vec{H} = -\vec{\nabla} \left\{ \frac{\mu_0 I}{4\pi} \frac{\vec{A} \cdot \vec{r}}{r^3} \right\}$$

We know  $d\Omega = (\nabla \Omega) \cdot d\vec{l}$

$$\vec{H} = - \frac{\mu_0 I}{4\pi} \int \frac{d\vec{l} \times \vec{r}}{r^3} = - \frac{\mu_0 I}{4\pi} \int d\vec{l} \times \vec{\nabla} \left( \frac{1}{r} \right)$$

$$\phi_m = \frac{\mu_0 I}{4\pi} \frac{\vec{A} \cdot \vec{r}}{r^3}$$

$$\phi_m = \frac{\mu_0 (\vec{I} \cdot \vec{A}) \cdot \vec{r}}{4\pi r^3} = \left( \frac{\mu_0}{4\pi} \right) \frac{\vec{m} \cdot \vec{r}}{r^3}$$

Note that  $\phi_m \Leftrightarrow V_{\text{dipole}}$  ✓ (Just  $\mu_0$  is missing : rest is same)

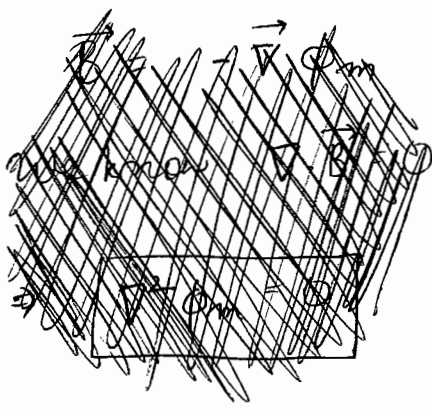
★ Hence all the results derived for dipole electrical are valid for  $\phi_m$ , corresponding to that of  $V$ .

$$\begin{aligned} \vec{H} &= - \vec{\nabla} \phi_m \\ &= - \frac{\mu_0}{4\pi} \vec{\nabla} \left( \frac{\vec{m} \cdot \vec{r}}{r^3} \right) \quad \left( \begin{array}{l} 2 \text{ scalars को} \\ \text{अलग अलग करो} \end{array} \right) \\ &= - \frac{\mu_0}{4\pi} \left[ \frac{\nabla \cdot (\vec{m} \cdot \vec{r})}{r^3} + (\vec{m} \cdot \vec{r}) \cdot \nabla \left( \frac{1}{r^3} \right) \right] \end{aligned}$$

$$\Rightarrow \vec{H} = + \frac{\mu_0}{4\pi} \left[ \frac{3(\vec{m} \cdot \vec{r}) \vec{r}}{r^5} - \frac{\vec{m}}{r^3} \right] \checkmark$$

$$U = - \vec{m} \cdot \vec{B} \quad \& \quad \vec{B} = \mu_0 \vec{H} \quad (\text{for vacuum})$$

$$\Rightarrow U = \frac{\mu_0}{4\pi} \left[ \frac{\vec{m}_1 \cdot \vec{m}_2}{r^3} - \frac{3(\vec{m}_1 \cdot \hat{r})(\vec{m}_2 \cdot \hat{r})}{r^3} \right]$$



i.e. Monopoles do not exist

experimentally proven

$$\oint_s \vec{B} \cdot d\vec{s} = 0$$

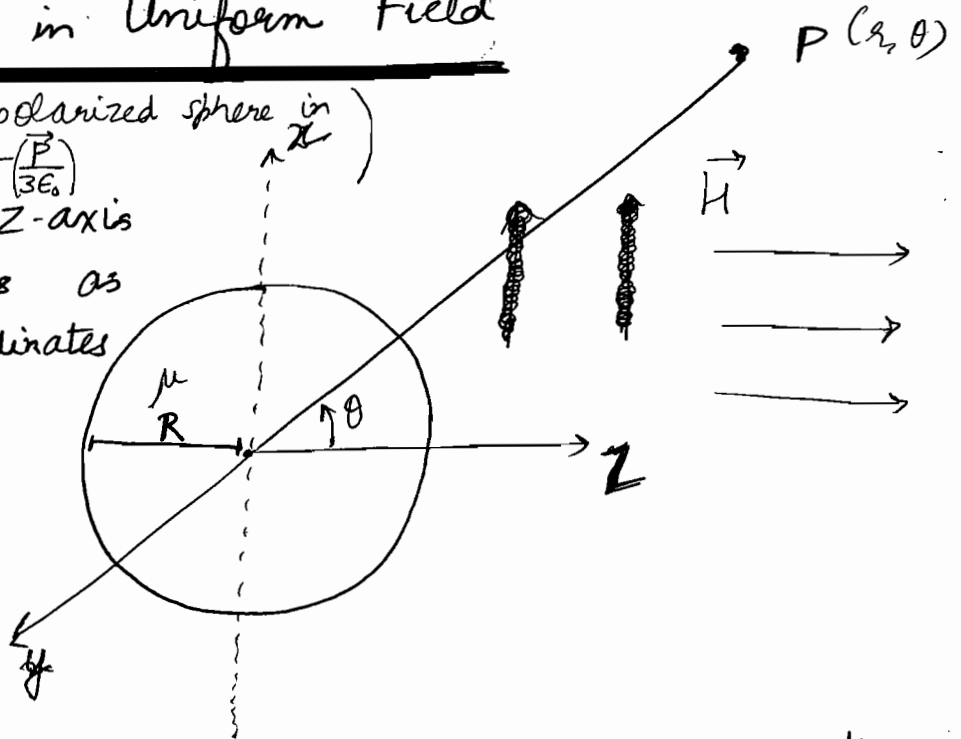
$$\Rightarrow \int_v \vec{\nabla} \cdot \vec{B} \, dv = 0$$

$$\Rightarrow \boxed{\vec{\nabla} \cdot \vec{B} = 0}$$

## Magnetized Sphere in Uniform Field

(Direct equivalence to polarized sphere in electric field  $\vec{E}_p = -\frac{\vec{P}}{3\epsilon_0}$ )

→ take  $\vec{H}$  along z-axis  
and  $\theta$  from z-axis as  
in spherical coordinates



$\vec{H}$  is uniform

$$\vec{M} = I \vec{A}$$

$$\vec{B} = \mu_0 \vec{H}$$

$$\vec{B} = \mu \vec{H}$$

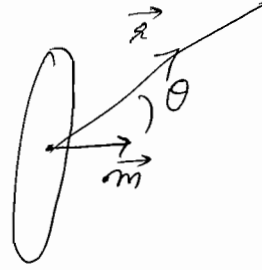
(at P, the free space)

(inside the sphere)

- ① there is current on the surface of the sphere
- ② current carrying loop is behaving as magnetic dipole

$R > r$

$$\phi_m = \frac{1}{4\pi} \frac{m \cdot \vec{r}}{r^3} = \frac{m \cos \theta}{4\pi r^2}$$



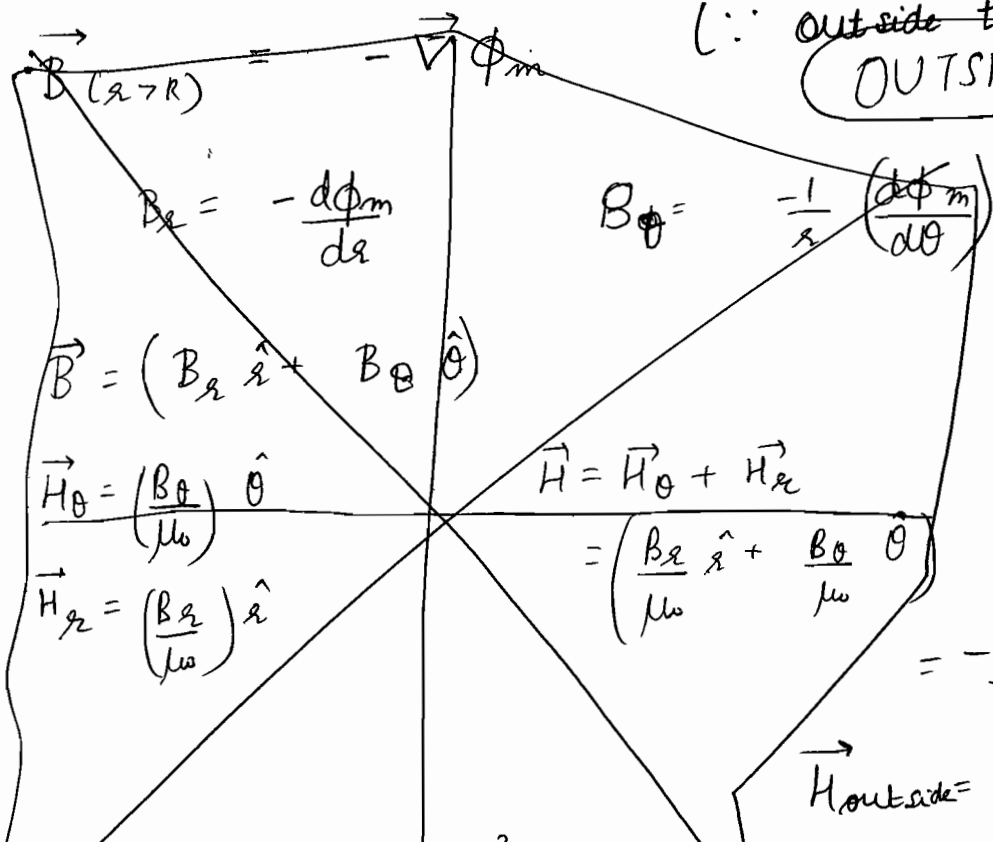
We know  $M = \frac{m}{\left(\frac{4}{3}\pi R^3\right)}$

Part A No external field

$$\Rightarrow \phi_m = \frac{1}{4\pi} \cdot \frac{m \cos \theta}{r^2} = \frac{1}{4\pi} \cdot \frac{4\pi M R^3}{3} \frac{\cos \theta}{r^2}$$

$$\phi_m = \frac{M R^3}{3} \left( \frac{\cos \theta}{r^2} \right)$$

$\therefore$  outside there is no free currents  
**OUTSIDE**



$$\vec{H}_{outside} = -\vec{\nabla} \phi_m$$

$$\vec{B}_{outside} = \mu_0 \vec{H}$$

$$\vec{H}_{out} = -\vec{\nabla} \left( \frac{M R^3 \cos \theta}{3 r^2} \right)$$

$$= -\frac{M R^3}{3} \vec{\nabla} \left( \frac{\cos \theta}{r^2} \right)$$

$$= -\frac{M R^3}{3} \left[ \frac{-2 \cos \theta}{r^3} \hat{r} - \frac{\sin \theta}{r^3} \hat{\theta} \right]$$

$$\vec{H}_{outside} = \frac{M R^3}{3 r^3} \left[ 2 \cos \theta \hat{r} + \sin \theta \hat{\theta} \right]$$

We know  $\nabla^2 \phi_m = 0$

$$\& \phi_{in} = \phi_{out} |_{r=R} = \frac{\mu_0 M R \cos \theta}{3}$$

$$\text{let } \phi_{in} = \frac{\mu_0 M}{3} z$$

$$\vec{B} = -\vec{\nabla} \phi_{in}$$

$$= -\frac{d\phi_m}{dz}$$

$$= -\left(\frac{\mu_0 M}{3}\right)$$

$$\vec{H}_{total} = \frac{\vec{B}}{\mu} = -\left(\frac{\vec{M}}{3}\right)$$

$$\vec{H} = \vec{H}_0 - \frac{\vec{M}}{3}$$

INSIDE

$$\phi_m (r=R) = \frac{\mu_0 M R \cos\theta}{3}$$

$$\phi_m (inside) = \frac{\mu_0 M z}{3}$$

$$\Rightarrow \vec{H}_{inside} = -\vec{\nabla} \phi_m$$

$$= -\frac{\mu_0 \vec{M}}{3} \hat{z}$$

This method is justified by the Uniqueness theorem

$$\vec{B}_{in} = \mu_0 (\vec{H} + \vec{M})$$

$$= \frac{2}{3} \mu_0 \vec{M}$$

Now let us also apply an external field  $H_0$

$$H_{total} = H_{applied} + H_{magnetization}$$

$$\Rightarrow \vec{H}_{total\ inside} = \vec{H}_0 - \frac{\vec{M}}{3}$$

Part B  
External Field  $H_0$

if linear medium, we have

$$\vec{M} = (\mu_r - 1) \vec{H}_{total}$$

$$= (\mu_r - 1) \left( \vec{H}_0 - \frac{\vec{M}}{3} \right)$$

$$\Rightarrow \vec{M} \left[ 1 + \frac{\mu_r - 1}{3} \right] = (\mu_r - 1) \vec{H}_0$$

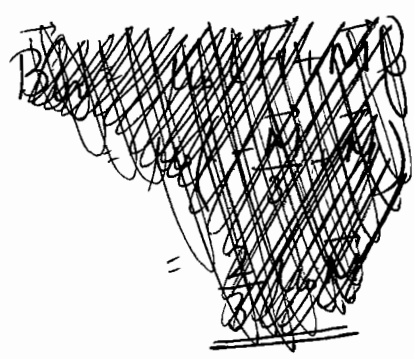
$$\Rightarrow \vec{M} = \frac{3(\mu_r - 1) \vec{H}_0}{(\mu_r + 2)}$$

$$\Rightarrow \vec{H}_{total} = \vec{H}_0 - \frac{(\mu_r - 1) \vec{H}_0}{(\mu_r + 2)} = \frac{3 \vec{H}_0}{(\mu_r + 2)}$$

$$B = \mu_0 (\vec{H}_{total} + \vec{M})$$

$$\vec{D} = \epsilon_0 \vec{E} + \vec{P}$$

$$\vec{B} = \mu_0 (\vec{H}_{total} + \vec{M})$$



# Electrodynamics

## Ohm's law

1st form  $\vec{J} = \sigma \vec{E}$

2nd form  $V = IR$

$\sigma$ : conductivity of the medium (No Proof)

$R$ : resistance of the medium (No Proof shown from cylindrical example)

Now we know.  $\vec{\nabla} \cdot \vec{J} = -\left(\frac{\partial \rho}{\partial t}\right)$  Continuity equation

$\vec{\nabla} \cdot \vec{E} = \left(\frac{\rho}{\epsilon_0}\right)$  Gauss law

$\Rightarrow$  For steady currents and uniform conductivity,

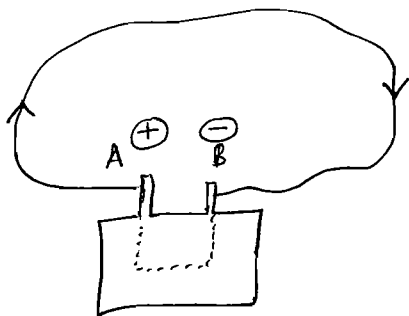
$\vec{\nabla} \cdot \vec{E} = \frac{1}{\sigma} \vec{\nabla} \cdot \vec{J} = 0$  (since  $\vec{\nabla} \cdot \vec{J} = 0$  for steady currents)

Now, from Gauss law,  $\rho = 0$

Therefore the charge density is zero; only unbalanced charge resides on the surface. Therefore, this is true not only for stationary charges but also for steady currents.

$\Rightarrow$  Laplace equation holds good within homogeneous Ohmic material.

## Electromotive force (E)



$$E = V_A - V_B = - \int_b^a \vec{E} \cdot d\vec{\ell}$$

$$= \int_a^b \vec{E} \cdot d\vec{\ell}$$

$$= \int_b^a \vec{f}_s \cdot d\vec{\ell}$$

$f_s$ : external force per unit charge provided by battery.

There are two forces that are involved in driving current around the loop:

1) the source  $f_s$ , which is ordinarily

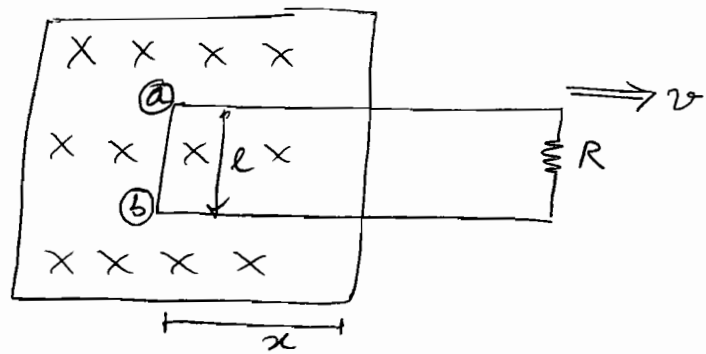


Confined to one portion of the loop (eg. battery)

- ② Electrostatic force, which serves to smooth out the flow and the communicate the influence of the source to distant parts of the circuit.

## Motional EMF

- ① loop is moving towards right with velocity  $v$



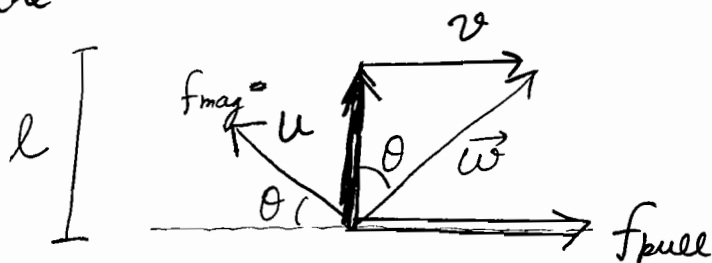
$$f_{\text{mag}} = \frac{q \cdot v \cdot B}{r} = vB$$

$$\mathcal{E} = \int_b^a f_{\text{mag}} \cdot dl = \underline{\underline{vB \cdot l}}$$

this is called Motional EMF.

## Work Done in Motional EMF

Although magnetic force is responsible for establishing the emf, it certainly does ~~not~~ work: magnetic forces never do work. The person who pulls the loop does the work.



$u$ : vertical component of velocity of charges

$$f_{\text{pull}} = f_{\text{mag}} \cos \theta = uB$$

Work done per unit charge,

$$\int f_{\text{pull}} \cdot dl = uB \cdot \frac{l}{\cos \theta} \sin \theta = \underline{\underline{Bl \cdot v}} = \mathcal{E}$$

Note that  $f_{\text{pull}}$  contributes nothing to EMF  $\therefore$  it is  $\perp$  to wire while  $f_{\text{mag}}$  contributes nothing to work  $\therefore$

its perpendicular to motion of charge. This is because, to calculate EMF, we integrate around the loop at one instant, but to calculate work done, we need to follow a charge in its path of motion.

The motional EMF in the general case is written as

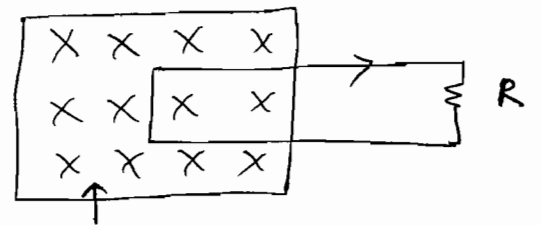
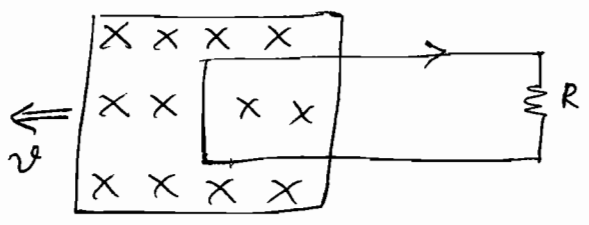
$$\mathcal{E} = -\left(\frac{d\phi}{dt}\right)$$

< Proof on P-315 >  
D. Griffiths

where  $\phi$  is the flux cutting through the loop.

### Electromagnetic Induction

Now consider the other two experiments of Faraday (apart from Motional EMF)



We have induced EMF in both these cases, even though, no charge is moving. So what applies for ??

In both these field is changing (In 1, at a fixed location, field is changing with time as the magnet passes by)

Faraday suggested

A changing magnetic field induces an electric field

$$\mathcal{E} = \oint \vec{E} \cdot d\vec{l} = -\left(\frac{d\phi}{dt}\right)$$

[Motional emf formula is always valid]

$$\Rightarrow \oint \vec{E} \cdot d\vec{l} = -\int \left(\frac{\partial \vec{B}}{\partial t}\right) \cdot d\vec{a}$$

$$\mathcal{E} = -\left(\frac{d\phi}{dt}\right)$$

$$\vec{\nabla} \times \vec{E} = -\left(\frac{\partial \vec{B}}{\partial t}\right)$$

ज्यादा दिमाग लगाने की जरूरत नहीं है !!

Direction of current flow is given by Lenz's law :

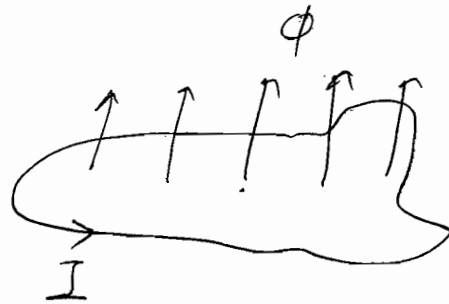
Nature abhors a change in flux

## Inductance

Flux is proportional to the current that produces the flux

For a single loop,

current  $I$  will produce some flux in the loop

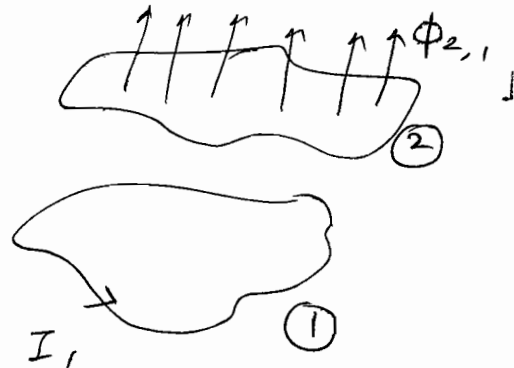


$$\phi \propto I$$

$$\phi = LI$$

$L$ : self inductance (depends upon geometry of the loop)

For a double loop, due to current flowing in loop ①, a flux will pierce through loop ②



$$\phi_{2,1} = M_{2,1} I_1$$

$M_{21}$ : Mutual inductance of two loops

Whatever the shapes and positions of the loops, the flux

depend on geometry & relative position of two loops

$$M_{21} = M_{12}$$

through 2 when we run a current  $I$  through 1 is identical to flux through 1 when we send the same current  $I$  around 2.

Now when the currents are varied, EMF is induced in both the situations,

$$\mathcal{E}_2 = -\frac{d\Phi_2}{dt} = -M \left( \frac{dI_1}{dt} \right)$$

$$\mathcal{E} = -L \left( \frac{dI}{dt} \right)$$

Inductance is measured in Henrys (H).  $1 \text{ H} = 1 \text{ V}\cdot\text{sec}/\text{Am}$

\* Note that in inductance, flux is always taken as  $N\Phi$  where  $N$  are no. of loops and  $\Phi$  is flux through single loop. (flux linkage)

\*  $\ominus$ ve sign in induced EMF means Back EMF. It acts opposite to the EMF that tries to alter the current in the wire.

### Energy in Magnetic Fields

$$W = \frac{1}{2} L I^2$$

$W$ : Work done to build up current  $I$ .

$$\text{Also, } W = \frac{1}{2\mu_0} \int_{\text{all space}} B^2 dV$$

This is the energy stored in magnetic field.

$W$ : Work done to set up magnetic field  $B$ .

\* It feels strange that though magnetic fields themselves do no work, energy is required to set up magnetic fields.

Point is that producing a magnetic field, where previously there was none, requires changing the field, and a changing magnetic field induces Electric field which of course does work.

In the beginning there is no  $\vec{E}$  and at the end there is no  $\vec{E}$ ; but in between while  $\vec{B}$  is building up, there is an  $\vec{E}$  and against this that the work is done.

## Modification in Ampere's law

$$\oint \vec{B} \cdot d\vec{\ell} = \mu_0 I_{enc}$$

Consider a charging capacitor ckt. Let's construct an Amperian loop

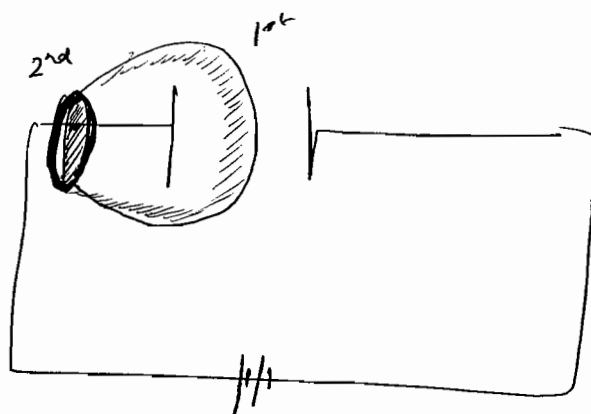
From 1<sup>st</sup> surface,

$$\oint \vec{B} \cdot d\vec{\ell} = 0$$

From 2<sup>nd</sup> surface

$$\oint \vec{B} \cdot d\vec{\ell} = \mu_0 I_{enc}$$

This is a contradiction.



Also from Ampere's law,

$$\vec{\nabla} \times \vec{B} = \mu_0 \vec{J}$$

$$\Rightarrow \vec{\nabla} \cdot (\vec{\nabla} \times \vec{B}) = 0 \quad (\text{divergence of curl} = 0)$$

$$\Rightarrow \mu_0 (\vec{\nabla} \cdot \vec{J}) = 0$$

$$\text{Now } \vec{\nabla} \cdot \vec{J} = -\frac{\partial \rho}{\partial t} = -\frac{\partial}{\partial t} (\epsilon_0 \vec{\nabla} \cdot \vec{E}) = -\vec{\nabla} \cdot \left( \epsilon_0 \frac{\partial \vec{E}}{\partial t} \right)$$

Therefore the contradiction is removed by introducing another electric field, st.

$$\vec{\nabla} \times \vec{B} = \mu_0 \vec{J} + \mu_0 \epsilon_0 \left( \frac{\partial \vec{E}}{\partial t} \right)$$

Therefore,

A changing electric field induces a magnetic field

This theoretical argument was confirmed by 6.4  
HERTZ EXPERIMENTS.

Now this new current density was called  
Displacement Current Density,

$$\vec{J}_d = \epsilon_0 \left( \frac{\partial \vec{E}}{\partial t} \right)$$

## Maxwell Equations

### General form

$$\vec{\nabla} \cdot \vec{E} = \frac{\rho}{\epsilon_0}$$

$$\vec{\nabla} \cdot \vec{B} = 0$$

$$\vec{\nabla} \times \vec{E} = - \left( \frac{\partial \vec{B}}{\partial t} \right)$$

$$\vec{\nabla} \times \vec{B} = \mu_0 \vec{J} + \mu_0 \epsilon_0 \left( \frac{\partial \vec{E}}{\partial t} \right)$$

### Maxwell Equations in Matter

$$\checkmark \vec{\nabla} \cdot \vec{D} = \rho_f$$

$$\checkmark \vec{\nabla} \cdot \vec{B} = 0$$

$$\checkmark \vec{\nabla} \times \vec{E} = - \left( \frac{\partial \vec{B}}{\partial t} \right)$$

$$\checkmark \vec{\nabla} \times \vec{H} = \vec{J}_f + \left( \frac{\partial \vec{D}}{\partial t} \right)$$

$$\begin{aligned} \text{Now } \oint \vec{B} \cdot d\vec{l} &= \int \mu_0 \cdot \epsilon_0 \cdot \left( \frac{\partial \vec{E}}{\partial t} \right) \cdot d\vec{a} \\ &= \mu_0 \epsilon_0 \frac{\partial}{\partial t} \int \vec{E} \cdot d\vec{a} \\ &= \mu_0 \frac{\partial}{\partial t} q_{enc} = \mu_0 \vec{J}_d \end{aligned}$$

$$\vec{F} = q (\vec{E} + \vec{v} \times \vec{B})$$

Maxwell's equations tell  
us how charges and  
currents produce fields.

Reciprocally, the force  
law tell us how  
fields affect charges

(particular case  
of linear materials)

We write them because  
inside polarized or magnetized  
matter there will be  
accumulations of "bound" charge  
and current over which we have  
no direct control. Its nice  
to formulate them in such  
a way that they refer to  
those sources that we control  
ie. "free" charge and currents.

here  $\vec{D} = \epsilon_0 \vec{E} + \vec{P}$   
 $\vec{H} = \frac{\vec{B}}{\mu_0} - \vec{M}$

For linear media,  $\vec{D} = \epsilon \vec{E}$   $\epsilon = \epsilon_0 (1 + \chi_e)$   
 $\vec{H} = \frac{\vec{B}}{\mu}$   $\mu = \mu_0 (1 + \chi_m)$

### Boundary Conditions

①  $\epsilon_1 E_{1\perp} - \epsilon_2 E_{2\perp} = \sigma_f$

②  $\vec{E}_1^{\parallel} - \vec{E}_2^{\parallel} = 0$  ( $\vec{\nabla} \times \vec{D} \neq 0$  but  $\vec{\nabla} \times \vec{E} = 0$ )

③  $B_{1\perp} - B_{2\perp} = 0$

④  $\frac{1}{\mu_1} \vec{B}_1^{\parallel} - \frac{1}{\mu_2} \vec{B}_2^{\parallel} = \vec{k}_f \times \hat{n}$  ( $\vec{\nabla} \times \vec{H} = \vec{J}_{free}$ )

जिन surfaces के across field component continuous है, उनमें तो normal field ( $E_{\parallel}$ ,  $B_{\perp}$ ) आरेंगी !!

जिन components में discontinuity है, उनमें modified forms आरेंगी ( $D_{\perp}$ ,  $H_{\parallel}$ )

In electrostatics & magnetostatic,  $\rho_b = -\vec{\nabla} \cdot \vec{P}$ ;  $\vec{J}_b = \vec{\nabla} \times \vec{M}$

In electro-dynamics cases, there is 1 more thing:  $\vec{J}_p$  Polarization current density

Any change in Polarization involves flow of bound charges suppose Polarization introduces a charge density  $\sigma_b = P$  at one end and  $-\sigma_b$  at other. If  $P$  now increases a bit, charge on each end increases  $\Rightarrow$  new current  $\vec{J}_p = \left( \frac{\partial \vec{P}}{\partial t} \right)$

$\Rightarrow \vec{J} = \vec{J}_f + \vec{J}_b + \vec{J}_p = \vec{J}_f + \vec{\nabla} \times \vec{M} + \left( \frac{\partial \vec{P}}{\partial t} \right)$

$\Rightarrow \vec{\nabla} \times \vec{B} = \mu_0 \vec{J}_f + \mu_0 (\vec{\nabla} \times \vec{M}) + \mu_0 \left( \frac{\partial \vec{P}}{\partial t} \right) + \mu_0 \frac{\partial \vec{D}}{\partial t} - \mu_0 \left( \frac{\partial \vec{P}}{\partial t} \right) \Rightarrow \vec{\nabla} \times \vec{H} = \vec{J}_f + \left( \frac{\partial \vec{D}}{\partial t} \right)$

⊙ Biot Savart's law : 
$$d\vec{B} = \frac{\mu_0 I}{4\pi} \frac{d\vec{l} \times \vec{r}}{r^3}$$

- application :
- ① long wire
  - ② circular coil
  - ③ solenoid

### Maxwell Modification of Ampere's law

⊙ Ampere's law 
$$\oint_S \vec{B} \cdot d\vec{l} = \mu_0 I$$

⊙ Conservation of charge 
$$I = \int_S \vec{J} \cdot d\vec{s}$$

$$\vec{I} \cdot d\vec{l} = \int_V \vec{J} \cdot d\vec{\tau}$$

$$I = - \frac{dq}{dt}$$

$$\int_S \vec{J} \cdot d\vec{s} = - \frac{d}{dt} \int_V \rho \, dv$$

$$\int \nabla \cdot \vec{J} \, dv = - \int \frac{d\rho}{dt} \, dv$$

$$\Rightarrow \int (\nabla \cdot \vec{J} + \frac{\partial \rho}{\partial t}) \, d\tau = 0$$

⇒ 
$$\boxed{\nabla \cdot \vec{J} + \left(\frac{\partial \rho}{\partial t}\right) = 0}$$

always valid  
"Law of Conservation of Charge"



We know by Ampere's law,

$$\oint \vec{B} \cdot d\vec{\ell} = \mu_0 \int_S \vec{J} \cdot d\vec{s}$$

$$\int (\vec{\nabla} \times \vec{B}) \cdot d\vec{s} = \int \mu_0 \vec{J} \cdot d\vec{s} = 0$$

$$\Rightarrow \int (\vec{\nabla} \times \vec{B} - \mu_0 \vec{J}) \cdot d\vec{s} = 0$$

$$\Rightarrow \boxed{\frac{1}{\mu_0} (\vec{\nabla} \times \vec{B}) = \vec{J}}$$

But divergence of a curl = 0

$$\Rightarrow \nabla \cdot \vec{J} = \frac{1}{\mu_0} \vec{\nabla} \cdot (\vec{\nabla} \times \vec{B}) = 0$$

But by continuity of charge,  $\nabla \cdot \vec{J} = -\left(\frac{\partial \rho}{\partial t}\right)$

$$\text{We know } \rho = \epsilon_0 \vec{\nabla} \cdot \vec{E}$$

$$\frac{d\rho}{dt} = \vec{\nabla} \cdot \left( \epsilon_0 \frac{\partial \vec{E}}{\partial t} \right)$$

Conservation of charge cannot be violated.

Hence Ampere's law is valid in form of  $\left[ \oint \vec{B} \cdot d\vec{\ell} = \mu_0 I \right]$

only when  $\left(\frac{\partial \vec{E}}{\partial t}\right) = 0$  i.e. Electric Field is not varying

Applying the correction for time varying fields,

$$\boxed{\vec{\nabla} \times \vec{B} = \mu_0 \left[ \vec{J} + \epsilon_0 \frac{\partial \vec{E}}{\partial t} \right]}$$

$$\boxed{\vec{J} = \sigma \vec{E}}$$

This correction to Ampere's law was done by Maxwell for time varying Electric Fields.

We know for.

$$\vec{P} + \epsilon_0 \vec{E} = \vec{D} \quad ; \quad \frac{\vec{B}}{\mu_0} - \vec{M} = \vec{H} \quad ; \quad \begin{aligned} J_b &= \nabla \times \vec{M} \\ J_p &= (\frac{\partial \vec{P}}{\partial t}) \end{aligned}$$

$$\Rightarrow \nabla \times \vec{H} = \left[ \vec{J}_f + \frac{\partial \vec{D}}{\partial t} \right]$$

Define  $\frac{\partial \vec{D}}{\partial t}$  as Surface Displacement Current density,  $J_D$ .

We know,  $\vec{J}$  is called Surface Conduction Current density,  $J$ .

Displacement Current arises due to variation of charge density.

$$J_D = (\frac{\partial D}{\partial t})$$

$$I_D = J_D \cdot A$$

$$I_D = \epsilon A \left( \frac{\partial E}{\partial t} \right)$$

Experimental foundation of Ampere's law is in Biot-Savart's law

### o Faraday's Law of Electromagnetic Induction

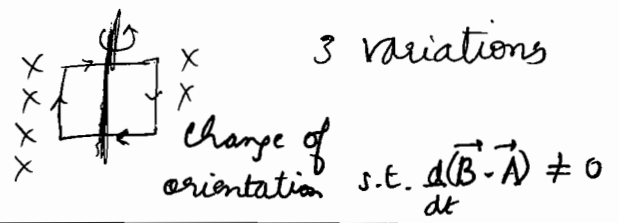
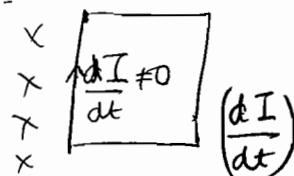
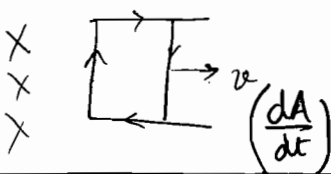
$$\text{Flux } \Phi_B = \int \vec{B} \cdot d\vec{s} = \vec{B} \cdot \vec{A}$$

\* Note that  $\int \vec{B} \cdot d\vec{a} \neq 0$   
but rather  $\oint \vec{B} \cdot d\vec{a} = 0$

It can be hypothetically represented as lines of force.

$$\frac{d\Phi}{dt} = \vec{B} \cdot \left( \frac{d\vec{A}}{dt} \right) + \left( \frac{d\vec{B}}{dt} \right) \cdot \vec{A}$$

$$\text{We know } B \propto I \Rightarrow \left( \frac{dB}{dt} \right) \propto \left( \frac{dI}{dt} \right)$$



3 variations

Change of orientation s.t.  $\frac{d(\vec{B} \cdot \vec{A})}{dt} \neq 0$

$$\phi_B = BA \cos \omega t$$

Rotation of current loop.

$$\left[ \begin{array}{l} \text{For } n \text{ turns of coil,} \\ \phi_B = nBA \cos \omega t \end{array} \right]$$

$$\frac{d\phi_B}{dt} = -\omega BA \sin(\omega t)$$

this was observed!!

Due to any of these 3 changes, whenever  $\phi$  changes, there is an induced emf. The direction of flow of current can be understood from lenz's law.

$$\boxed{\text{Emf} = \int_C \vec{E} \cdot d\vec{l} = - \left( \frac{d\phi_B}{dt} \right)} \quad \text{Faraday's law}$$

Now we need to write the law in integral & differential forms

$$\oint_C \vec{E} \cdot d\vec{l} = - \frac{d}{dt} \int_S \vec{B} \cdot d\vec{s}$$

$$\boxed{\oint_C \vec{E} \cdot d\vec{l} = - \int_S \left( \frac{\partial \vec{B}}{\partial t} \right) \cdot d\vec{s}}$$

Note that here we are interested in a closed loop & not closed surface  $\Rightarrow$  we will apply Stokes Law

Faraday's law in Integral form

$$\begin{aligned} \int (\vec{\nabla} \times \vec{E}) \cdot d\vec{s} &= - \int \left( \frac{\partial \vec{B}}{\partial t} \right) \cdot d\vec{s} \\ \Rightarrow \int \left[ \vec{\nabla} \times \vec{E} + \left( \frac{\partial \vec{B}}{\partial t} \right) \right] \cdot d\vec{s} &= 0 \end{aligned}$$

$$\Rightarrow \boxed{\vec{\nabla} \times \vec{E} + \left( \frac{\partial \vec{B}}{\partial t} \right) = 0}$$

Faraday's law in differential form

For electrostatic field  $\vec{\nabla} \times \vec{E} = 0$

but for time varying fields,  $\vec{\nabla} \times \vec{E} = -\left(\frac{\partial \vec{B}}{\partial t}\right)$

It is an experimental law.

① Symmetry in Nature and Faraday's law motivated Maxwell to introduce correction into Ampere's law.

### Maxwell Equations

From the theory studied so far,

$$\left[ \oint_s \vec{E} \cdot d\vec{s} = \frac{q_{\text{ext}}}{\epsilon_0} = \frac{1}{\epsilon_0} \int_V \rho_{\text{ext}} dV \right] \begin{array}{l} \text{Gauss Law in Integral} \\ \text{Form} \end{array}$$

$$\left[ \begin{array}{l} \nabla \cdot \vec{E} = \frac{\rho}{\epsilon_0} \quad : \text{Free Space} \\ \nabla \cdot \vec{E} = \frac{\rho_f}{\epsilon} \quad : \text{linear dielectric} \\ \quad \quad \quad \quad \quad \quad : \text{Material} \end{array} \right] \begin{array}{l} \text{Differential Form} \\ \text{of Gauss law} \end{array}$$

$$\left[ \begin{array}{l} \vec{\nabla} \cdot \vec{B} = 0 \\ \oint_s \vec{B} \cdot d\vec{s} = 0 \end{array} \right] \quad \text{Monopoles do not exist}$$

$$\left[ \begin{array}{l} \int_c \vec{E} \cdot d\vec{l} = - \int_s \left(\frac{\partial \vec{B}}{\partial t}\right) \cdot d\vec{s} \\ \vec{\nabla} \times \vec{E} = - \left(\frac{\partial \vec{B}}{\partial t}\right) \end{array} \right] \quad \begin{array}{l} \text{Faraday's law of} \\ \text{Electromagnetic Induction} \end{array}$$

$$\left[ \begin{array}{l} \int_c \vec{B} \cdot d\vec{l} = \mu_0 \int (\vec{J} + \vec{J}_d) d\vec{s} \\ \quad \quad \quad = \mu_0 (I + I_d) \\ (\vec{\nabla} \times \vec{B}) = \mu_0 [\vec{J} + \vec{J}_d] \end{array} \right] \quad \begin{array}{l} \text{Ampere's law as corrected} \\ \text{by Maxwell} \quad \vec{J}_d = \epsilon \left(\frac{\partial \vec{E}}{\partial t}\right) \end{array}$$

# Maxwell's Equations in Electromagnetics

$$\vec{\nabla} \cdot \vec{E} = \left( \frac{\rho}{\epsilon_0} \right) \quad \text{--- (1) } \checkmark$$

$$\vec{\nabla} \cdot \vec{B} = 0 \quad \text{--- (2) } \checkmark$$

$$\vec{\nabla} \times \vec{E} = - \left( \frac{\partial \vec{B}}{\partial t} \right) \quad \text{--- (3) } \checkmark$$

$$\vec{\nabla} \times \vec{B} = \mu_0 (\vec{J} + \vec{J}_D) \quad \text{--- (4) } \checkmark$$

Note that differential form of Maxwell Equations is always preferred.

There are 3 cases of interest :-

1) Free Space (Vacuum)

2) Dielectrics : Isotropic  
or  
Anisotropic

dielectric are classified according to Permittivity  $\epsilon$

If  $\epsilon$  is same everywhere: ISOTROPIC

If  $\epsilon$  is changing: ANISOTROPIC

$$\checkmark \epsilon = \begin{bmatrix} \epsilon_{xx} & \epsilon_{xy} & \epsilon_{xz} \\ \epsilon_{yx} & \epsilon_{yy} & \epsilon_{yz} \\ \epsilon_{zx} & \epsilon_{zy} & \epsilon_{zz} \end{bmatrix}$$

3) Conducting Medium

Force free space,

$$\vec{J}_e = \sigma \vec{E}$$

$$\vec{J}_D = \left( \frac{\partial \vec{D}}{\partial t} \right)$$

Note that variable in Maxwell Equations are

$$\rho, \sigma, \epsilon, \mu$$

↑  
Conductivity

For free space,  $\rho = 0 \checkmark$   
 $\sigma = 0 \checkmark$   
 $\epsilon = \epsilon_0 \checkmark$   
 $\mu = \mu_0 \checkmark$

$$\Rightarrow \vec{\nabla} \cdot \vec{E} = 0 \checkmark$$

$$\vec{\nabla} \cdot \vec{B} = 0 \checkmark$$

$$\vec{\nabla} \times \vec{E} = - \left( \frac{\partial \vec{B}}{\partial t} \right) \checkmark$$

$$\vec{\nabla} \times \vec{B} = +\mu_0 \vec{J}_D = +\mu_0 \epsilon_0 \left( \frac{\partial \vec{E}}{\partial t} \right) \checkmark$$

For isotropic dielectric,

$$\epsilon = \epsilon$$

$$\underline{\mu = \mu_0} \text{ [Non Magnetic dielectric]}$$

$$\rho = 0 \text{ [No charge in glass]}$$

$$\sigma = 0 \text{ [Dielectric is non conducting medium]}$$

$$\Rightarrow \vec{\nabla} \cdot \vec{E} = 0, \vec{\nabla} \cdot \vec{B} = 0, \vec{\nabla} \times \vec{E} = - \left( \frac{\partial \vec{B}}{\partial t} \right), \vec{\nabla} \times \vec{B} = \mu_0 \epsilon \left( \frac{\partial \vec{E}}{\partial t} \right)$$

Note that its a flawed reasoning that since vacuum के लिए  $\epsilon \rightarrow$  material के लिए  $\epsilon$ .

Actually for material, we have addition polarization current

$$\Rightarrow \frac{\partial}{\partial t} (\epsilon \vec{E} + \vec{P}) = \frac{\partial \vec{D}}{\partial t} = \frac{\partial \epsilon \vec{E}}{\partial t}$$

$$= \epsilon \left( \frac{\partial \vec{E}}{\partial t} \right)$$



For Anisotropic dielectric,

(eg. Crystals are anisotropic different behaviour in preferred directions.)

$$\rho = 0, \quad \nabla = 0$$

$$\vec{D} = \epsilon \vec{E} \Rightarrow \begin{bmatrix} D_x \\ D_y \\ D_z \end{bmatrix} = \begin{bmatrix} \epsilon_{xx} & \epsilon_{xy} & \epsilon_{xz} \\ \epsilon_{yx} & \epsilon_{yy} & \epsilon_{yz} \\ \epsilon_{zx} & \epsilon_{zy} & \epsilon_{zz} \end{bmatrix} \begin{bmatrix} E_x \\ E_y \\ E_z \end{bmatrix}$$

for  $\nabla \cdot \vec{D} = 0$  ✓ ⊗

i.e.  $\frac{\partial D_x}{\partial x} + \frac{\partial D_y}{\partial y} + \frac{\partial D_z}{\partial z} = 0$

$$\textcircled{1} \Rightarrow \frac{\partial}{\partial x} [\epsilon_{xx} E_x + \epsilon_{xy} E_y + \epsilon_{xz} E_z] + \frac{\partial}{\partial y} [ \quad ] + \frac{\partial}{\partial z} [ \quad ] = 0$$

It can be seen that for vector form, equation is simple.  
But for component wise equation is lengthy.

$$\textcircled{2} \quad \nabla \cdot \vec{B} = 0 \quad \checkmark$$

$$\textcircled{3} \quad \nabla \times \vec{E} = -\left(\frac{\partial \vec{B}}{\partial t}\right) \quad \checkmark$$

$$\textcircled{4} \quad \nabla \times \vec{B} = \mu_0 \left(\frac{\partial \vec{D}}{\partial t}\right) \quad \checkmark$$

For Conducting Media,

$$\underline{\rho} = 0, \quad \nabla = \nabla, \quad \epsilon = \epsilon, \quad \mu = \mu_0 \text{ [Non Magnetic]}$$

$$\nabla \cdot \vec{E} = 0 \quad \checkmark$$

$$\nabla \cdot \vec{B} = 0 \quad \checkmark$$

$$\nabla \times \vec{E} = -\frac{\partial \vec{B}}{\partial t} \quad \checkmark$$

$$\nabla \times \vec{B} = \underbrace{\mu_0 \sigma \vec{E}} + \mu_0 \epsilon \left(\frac{\partial \vec{E}}{\partial t}\right) \quad \checkmark$$

↑  
[const.]

if  
isotropic Conducting Media

The next step in all Maxwell Equations is to find out  $\vec{E}$  and  $\vec{B}$ . The solution is wave equations  $\nabla^2 \psi = \frac{1}{v_p^2} \left( \frac{\partial^2 \psi}{\partial t^2} \right)$

where  $\psi$  comes out to be  $\psi = \psi_0 e^{i(\vec{k} \cdot \vec{r} - \omega t)}$

$$\text{w.t. } v_p = \left( \frac{\omega}{k} \right)$$

TAKE THE CURL OF 3<sup>rd</sup> & 4<sup>th</sup>

For Free space,

$$\vec{\nabla} \times (\vec{\nabla} \times \vec{E}) = -\vec{\nabla} \times \left( \frac{\partial \vec{B}}{\partial t} \right)$$

$$\vec{\nabla} (\vec{\nabla} \cdot \vec{E}) - \vec{E} (\vec{\nabla} \cdot \vec{\nabla}) = -\frac{\partial}{\partial t} (\vec{\nabla} \times \vec{B})$$

Invoking ① and ④ equations,

$$\vec{\nabla} (0) - \nabla^2 \vec{E} = -\frac{\partial}{\partial t} \left( \mu_0 \epsilon_0 \frac{\partial \vec{E}}{\partial t} \right)$$

$$\nabla^2 \vec{E} = \mu_0 \epsilon_0 \frac{\partial^2 \vec{E}}{\partial t^2}$$

For dielectric isotropic

(Note that  $\rho = 0$  for dielectric)

$$\nabla^2 \vec{E} = \mu \epsilon \left( \frac{\partial^2 \vec{E}}{\partial t^2} \right)$$

For conducting isotropic

$$\nabla^2 \vec{E} = \mu \sigma \left( \frac{\partial \vec{E}}{\partial t} \right) + \mu \epsilon \left( \frac{\partial^2 \vec{E}}{\partial t^2} \right)$$

For Free space,

$$\vec{\nabla} \times (\vec{\nabla} \times \vec{B}) = \vec{\nabla} \times \left[ \mu_0 \epsilon_0 \frac{\partial \vec{E}}{\partial t} \right]$$

$$\nabla^2 \vec{B} = -\mu_0 \epsilon_0 \frac{\partial}{\partial t} (\vec{\nabla} \times \vec{E}) = -\mu_0 \epsilon_0 \frac{\partial}{\partial t} \left( -\frac{\partial \vec{B}}{\partial t} \right) = \mu_0 \epsilon_0 \left( \frac{\partial^2 \vec{B}}{\partial t^2} \right)$$

$$\Rightarrow \nabla^2 \vec{B} = \mu_0 \epsilon_0 \left( \frac{\partial^2 \vec{B}}{\partial t^2} \right)$$

Curl of  $\vec{\nabla} \times \vec{B}$



# Conducting Media

$$\nabla^2 \vec{B} = \mu_0 \sigma \left( \frac{\partial \vec{B}}{\partial t} \right) + \mu_0 \epsilon \left( \frac{\partial^2 \vec{B}}{\partial t^2} \right)$$

✓ Since in all media,  $\vec{E}$  and  $\vec{B}$  follow same equations,  
hence both are always in same phase.

Define Mathematical Operator

$$\square^2 = \left[ \nabla^2 - \mu \epsilon \frac{\partial^2}{\partial t^2} \right] \quad : \quad \text{de Alembertian Operator}$$

for dielectric

$$\left[ \nabla^2 - \mu \epsilon \frac{\partial^2}{\partial t^2} \right] \{ \vec{E} \} = 0 \quad \& \quad \left[ \nabla^2 - \mu \epsilon \frac{\partial^2}{\partial t^2} \right] \{ \vec{B} \} = 0$$

$$\square^2 = \nabla^2 - \frac{1}{v^2} \frac{\partial^2}{\partial t^2}$$

where  $v = \frac{1}{\sqrt{\mu \epsilon}}$

For free space,  $\nabla^2 \vec{E} = \mu_0 \epsilon_0 \left( \frac{\partial^2 \vec{E}}{\partial t^2} \right)$   
 $\nabla^2 \vec{B} = \mu_0 \epsilon_0 \left( \frac{\partial^2 \vec{B}}{\partial t^2} \right)$  } these are wave equations

$$\nabla^2 \psi = \frac{1}{v^2} \frac{\partial^2 \psi}{\partial t^2}$$

$$v = \frac{1}{\sqrt{\mu_0 \epsilon_0}} = \frac{1}{\sqrt{4\pi \times 10^{-7} \times \frac{1}{9 \times 4\pi \times 10^9}}} = \underline{\underline{3 \times 10^8 \text{ m/s}}}$$

For any dielectric,  $v = \frac{1}{\sqrt{\mu_0 \epsilon_0 \epsilon_r}} = \frac{c}{\sqrt{\epsilon_r}} = \left( \frac{c}{n} \right)$

$$\Rightarrow \boxed{n = \sqrt{\epsilon_r}} \quad \checkmark$$

[although not explained empirically]

For most of the materials  $\mu \approx \mu_0$   
 $\therefore$  not much contribution to  $n$   
refractive index

We know,

$$\vec{P} = (\epsilon_r - 1) \epsilon_0 \vec{E} \quad (\text{for linear dielectric})$$

$$\epsilon_r = \frac{\vec{P}}{\epsilon_0 \vec{E}} + 1$$

$$\Rightarrow \boxed{n^2 = \frac{|\vec{P}|}{|\epsilon_0 \vec{E}|} + 1}$$

$$\psi = \psi_0 e^{i(\vec{k} \cdot \vec{r} - \omega t)}$$

$$\boxed{\begin{aligned} \vec{E}(r, t) &= \vec{E}_0 e^{i(\vec{k} \cdot \vec{r} - \omega t)} \\ \vec{B}(r, t) &= \vec{B}_0 e^{i(\vec{k} \cdot \vec{r} - \omega t)} \end{aligned}}$$

✓ PHASE PART OF BOTH  $\vec{E}$  and  $\vec{B}$  are same.....

$$\boxed{v_p = \left(\frac{\omega}{k}\right) = \frac{1}{\sqrt{\mu \epsilon}}}$$

✓ NOW EM waves are produced only when  $\vec{E}$  and  $\vec{B}$  change wrt. time. It happens when charges are accelerated or decelerated. It leads to propagation of  $\vec{E}$  and  $\vec{B}$  in space  
Acceleration / Deceleration usually occurs whenever charge particles hit some surface & collision occurs.

⊙ Note that the 2  $\vec{E}$  &  $\vec{B}$  are in same phase as long as  $\vec{k}$  is real.

$$(\nabla^2 - \mu \epsilon \frac{\partial^2}{\partial t^2}) \vec{\Psi} = \mu \epsilon \frac{\partial \vec{\Psi}}{\partial t}$$

$\vec{k}$  is imaginary for conductors.

$$\Psi = \Psi_0 e^{i(\vec{k} \cdot \vec{r} - \omega t)}$$

For conductors, we write  $\vec{k}^*$

$$\vec{k}^* = \vec{k}_+ + i \vec{k}_-$$

$$\Rightarrow \Psi = \Psi_0 e^{-i\omega t} [e^{i\vec{k}_+ \cdot \vec{r}} e^{i\vec{k}_- \cdot \vec{r}}]$$

$$= \Psi_0 e^{-\vec{k}_- \cdot \vec{r}} [e^{i(\vec{k}_+ \cdot \vec{r} - \omega t)}]$$

Conductor

- attenuation
- $\Delta \phi$  between E and B
- k imaginary

Conductor

It is attenuated wave equation.

If  $\sigma = 0$ , then it is dielectric  $\Rightarrow k$  is real

$\Rightarrow$  No attenuation of Amplitude

$$\Psi = \Psi_0 [e^{i(\vec{k} \cdot \vec{r} - \omega t)}]$$

Dielectric

Note that, for conductors, we can also write

$$\vec{k} = \vec{\alpha} + i\vec{\beta}$$

Phase Factor      Attenuation Factor

$$\Rightarrow \Psi = \Psi_0 e^{-\beta r} e^{-i(\vec{\alpha} \cdot \vec{r} - \omega t)}$$

$$v_p = \left( \frac{\omega}{\alpha} \right)$$

Proof of Transverse Nature of EM waves :-

$\nabla^2 \psi = \mu E \left( \frac{\partial^2 \psi}{\partial t^2} \right)$   $\star$  In order to prove transverse nature, we will need to define ~~the~~ operators corresponding to  $\vec{\nabla}$  and  $\frac{\partial}{\partial t}$

Solution for this equation :

$$\psi = \psi_0 e^{-i(\vec{k} \cdot \vec{r} - \omega t)}$$

Differentiating,

$$\left( \frac{\partial \psi}{\partial t} \right) = \psi_0 e^{-i(\vec{k} \cdot \vec{r} - \omega t)} i\omega$$

$$\Rightarrow \frac{\partial(\psi)}{\partial t} = i\omega(\psi)$$

$\langle$  Using it as Operator  $\rangle$

$$\Rightarrow \boxed{\frac{\partial}{\partial t} \Leftrightarrow i\omega} \quad \&$$

$$\boxed{\omega \Leftrightarrow -i \frac{\partial}{\partial t}}$$

$$\vec{\nabla} \psi = \sum i \frac{\partial}{\partial x} \left[ \psi_0 e^{i\omega t} e^{-i(k_x x + k_y y + k_z z)} \right]$$

$$= \psi_0 e^{i\omega t} \left[ \sum \hat{i} e^{-i(\vec{k} \cdot \vec{r})} \cdot -ik_x \right]$$

$$= -i\vec{k} \psi$$

$$\Rightarrow \vec{\nabla}(\psi) = -i\vec{k}(\psi)$$

$$\Rightarrow \boxed{\vec{\nabla} \Leftrightarrow -i\vec{k}} \quad \&$$

$$\boxed{\vec{k} \Leftrightarrow i\vec{\nabla}}$$

Using the operators, recasting the Maxwell equations,

$$\vec{\nabla} \cdot \vec{E} = 0 \Rightarrow \vec{k} \cdot \vec{E} = 0 \quad \checkmark \text{ transverse}$$

$$\vec{\nabla} \cdot \vec{B} = 0 \Rightarrow \vec{k} \cdot \vec{B} = 0 \quad \checkmark$$

$$\vec{\nabla} \times \vec{E} = -i\omega \vec{B} \Rightarrow -i\vec{k} \times \vec{E} = -i\omega \vec{B}$$

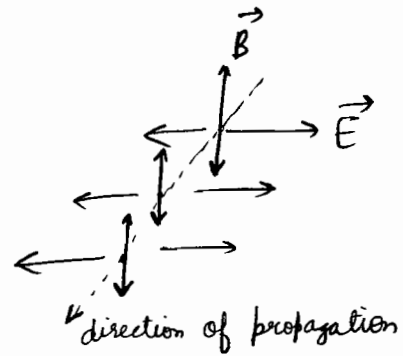
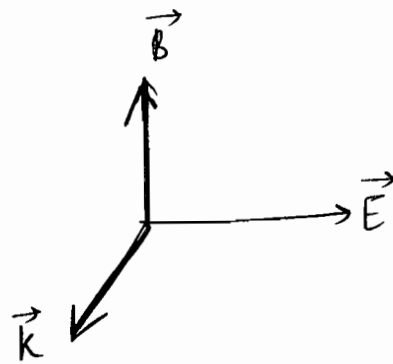
$$\Rightarrow \boxed{\vec{k} \times \vec{E} = \omega \vec{B}} \quad \checkmark$$

$$\vec{\nabla} \times \vec{B} = \mu \epsilon \frac{\partial \vec{E}}{\partial t} \Rightarrow -i\vec{k} \times \vec{B} = \mu \epsilon i\omega \vec{E}$$

$$\Rightarrow \boxed{\vec{k} \times \vec{B} = -\mu \epsilon \omega \vec{E}}$$

} Righthand triad of  $[\vec{E}, \vec{B}, \vec{k}]$

All the equations prove that all 3 are perpendicular and  $[\vec{E}, \vec{B}, \vec{k}]$  form a right handed triad



★ Till now we have seen 4 properties :

✓  $v = \frac{1}{\sqrt{\mu_0 \epsilon_0}} = c$  in empty space

✓  $v = \frac{c}{n} = \frac{c}{\sqrt{k}}$  in medium

✓ waves are transverse in nature

✓  $[\vec{E}, \vec{B}, \vec{k}]$  form a right handed triad.

✓  $\vec{E} = v \vec{B}$

$$\vec{B} = \frac{(\vec{k} \times \vec{E})}{\omega}$$

From 3<sup>rd</sup> equation:

$$\checkmark \frac{B_0}{E_0} = \frac{k}{\omega}$$

k can be real or complex

Taking the Amplitudes,

$$B = \frac{kE}{\omega} = \frac{E}{v}$$

$$\frac{E}{B} = \frac{\omega}{k} = \frac{1}{\sqrt{\mu\epsilon}} \text{ m/s} \checkmark$$

$$\frac{E}{H} = \frac{\mu}{\sqrt{\mu\epsilon}} = \sqrt{\frac{\mu}{\epsilon}} \text{ ohms} \checkmark$$

$$\frac{E}{H} = Z = \frac{Z_0}{n}$$

$$\Rightarrow H = \frac{nE}{Z_0} \checkmark$$

For free space,  $\frac{E}{H} = \sqrt{\frac{4\pi \times 10^{-7} \times 4\pi}{9 \times 10^9}} \text{ Ohms} = 4\pi \times 3 \times 10 \text{ } \Omega$   
 $= 120\pi \Omega = 376.99 \Omega$

$\frac{E}{H}$  is the impedance of electromagnetic waves in the medium.  
 It is the 'characteristic impedance' of the medium.

$\sqrt{\frac{\mu_0}{\epsilon_0}}$  is the characteristic impedance of free space = 377  $\Omega$  ( $Z_0$ )

$$\underline{Z} = \frac{|\vec{E}|}{|\vec{H}|} = \sqrt{\frac{\mu}{\epsilon}} = \sqrt{\frac{\mu_0}{\epsilon_0 \epsilon_r}} = \frac{Z_0}{\sqrt{\epsilon_r}} = \left(\frac{Z_0}{n}\right) \checkmark$$

in a dielectric

⊗ It is the resistance offered by the medium for the propagation of EM wave.

→ 4<sup>th</sup> equation is used only for verification purpose.

We know

# Poynting's Theorem

$$\vec{B} = \frac{\vec{k} \times \vec{E}}{\omega}$$

\* 2 things to note in the proof:

$$\Rightarrow \vec{E} \times \vec{B} = \vec{E} \times \frac{(\vec{k} \times \vec{E})}{\omega}$$

$$\textcircled{1} \int_V \nabla \cdot (\vec{E} \times \vec{H}) d\tau = \int_V (\vec{E} \times \vec{H}) \cdot d\vec{a}$$

$$= \frac{1}{\omega} \left[ \vec{k} (\vec{E} \cdot \vec{E}) - \vec{E} (\vec{E} \cdot \vec{k}) \right]$$

$$\textcircled{2} \nabla \cdot (\vec{A} \times \vec{B}) = (\nabla \times \vec{A}) \cdot \vec{B} - (\nabla \times \vec{B}) \cdot \vec{A}$$

$$= \frac{\vec{k}}{\omega} E^2 = \left( \frac{E^2}{\omega} \right) \vec{k}$$

$$\Rightarrow \boxed{\vec{E} \times \vec{B} = \frac{E^2}{\omega} \vec{k}}$$

(note the negative sign)

$$\nabla \cdot (\vec{E} \times \vec{B}) = \vec{B} \cdot (\nabla \times \vec{E}) - \vec{E} \cdot (\nabla \times \vec{B})$$

Similarly, dividing by  $\mu$

Note that divergence of a curl = 0 but divergence of cross product may not be zero

$$\nabla \cdot (\vec{E} \times \vec{H}) = \vec{H} \cdot (\nabla \times \vec{E}) - \vec{E} \cdot (\nabla \times \vec{H})$$

$$\text{From 4th eqn, } \vec{E} \cdot \nabla \times \vec{H} = \left[ \vec{J} + \left( \frac{\partial \vec{D}}{\partial t} \right) \right] \cdot \vec{E} \quad \text{--- (1)}$$

$$\text{From 3rd eqn, } \vec{H} \cdot \nabla \times \vec{E} = -\mu \left( \frac{\partial \vec{H}}{\partial t} \right) \cdot \vec{H} \quad \text{--- (2)}$$

subtracting (1) from (2)

$$\nabla \cdot (\vec{E} \times \vec{H}) = -\mu \vec{H} \cdot \left( \frac{\partial \vec{H}}{\partial t} \right) - \vec{E} \cdot \left[ \vec{J} + \frac{\partial \vec{D}}{\partial t} \right]$$

$$= -\vec{E} \cdot \vec{J} - \frac{\partial}{\partial t} \left[ \frac{1}{2} \mu (\vec{H} \cdot \vec{H}) + \frac{1}{2} \epsilon (\vec{E} \cdot \vec{E}) \right]$$

$$\vec{\nabla} \cdot (\vec{E} \times \vec{H}) = -\vec{E} \cdot \vec{J} - \frac{\partial}{\partial t} \left[ \frac{1}{2} \mu H^2 + \frac{1}{2} \epsilon E^2 \right]$$

Taking the volume integral on both sides,

$$\int_V \vec{\nabla} \cdot (\vec{E} \times \vec{H}) dV = - \int_V \vec{E} \cdot \vec{J} dV - \frac{\partial}{\partial t} \int_V \left( \frac{1}{2} \mu H^2 + \frac{1}{2} \epsilon E^2 \right) dV$$

$$\boxed{- \int (\vec{E} \times \vec{H}) \cdot d\vec{A} = + \int_V \vec{E} \cdot \vec{J} dV + \frac{\partial}{\partial t} \int_V \left( \frac{1}{2} \mu H^2 + \frac{1}{2} \epsilon E^2 \right) dV}$$

Energy Continuity Equation i.e. Statement of Conservation of Energy in Electromagnetic Wave.

$$\Rightarrow \int_V \vec{E} \cdot \vec{J} dV + \frac{\partial}{\partial t} \int_V \left( \frac{1}{2} \mu H^2 + \frac{1}{2} \epsilon E^2 \right) dV + \int (\vec{E} \times \vec{H}) \cdot d\vec{A} = 0$$

↑  
Work done by EM wave on field charge

↑  
Total electrical & Magnetic Energy stored with EM field

↑  
Energy carried out by the wave

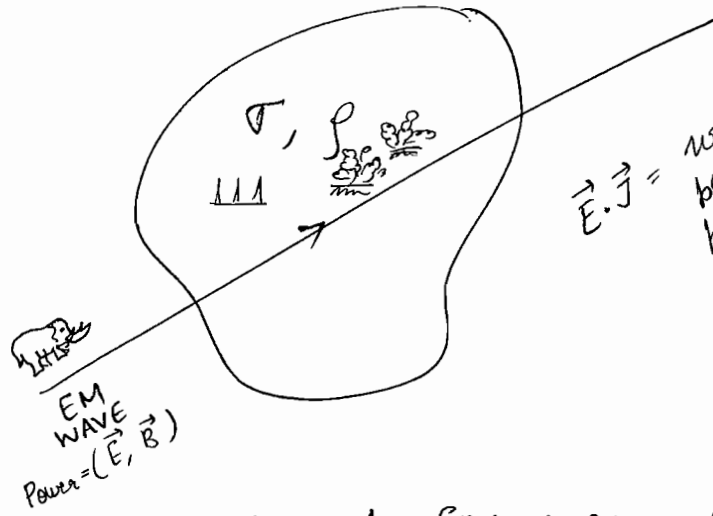
define:  $\vec{S} = \vec{E} \times \vec{H}$   
POYNTING VECTOR

: Energy carried out per second per unit area in the direction of propagation of wave!!

In wave theory, its called Intensity

✓ Note that the Poynting Equation is very general





○  $\int_V \vec{J} \cdot \vec{E} \, dV$  : Work done by EM wave on charges

If  $\sigma = 0 \Rightarrow \vec{J} = \sigma \vec{E} = 0 \Rightarrow$  NO work done.

$\vec{F} = [q\vec{E} + q(\vec{v} \times \vec{B})]$  (Force exerted by Elephant or EM wave)

~~Work~~, Work =  $\vec{F} \cdot \vec{v} \, dt \equiv \int q \vec{E} \cdot \vec{v} \, dt$

$$\Rightarrow \left( \frac{dW}{dt} \right) = \int_V \rho \, d\vec{E} \cdot \vec{v}$$

$$= \int_V \vec{E} \cdot \vec{J} \, dV$$

$$\boxed{q = \rho \, dV}$$

$$\boxed{\rho \vec{v} = \vec{J}}$$

$$\boxed{\sigma \vec{E} = \vec{J}}$$

(Ohmic materials)

This term becomes 0 when either  $\sigma = 0$  or  $\rho = 0$

○ Energy associated with EM wave having  $\vec{E}, \vec{B}$

$$= \int_V \left( \frac{1}{2} \mu H^2 + \frac{1}{2} \epsilon E^2 \right) dV$$

loss of energy due to resistance by  $\text{किर्च}$  and  $\text{सांघी}$  and  $\text{थकावट}$

$$= \frac{\partial}{\partial t} \left[ \int_V \left( \frac{1}{2} \mu H^2 + \frac{1}{2} \epsilon E^2 \right) dV \right]$$

o After doing work and encountering loss of energy, the elephant carries  $(\vec{E} \times \vec{H})$  energy per unit area per unit time.

o Electrical Energy : Magnetic Energy

$$\frac{1}{2} \epsilon E^2 : \frac{1}{2} \mu H^2 \quad (\text{per unit volume})$$

$$= \frac{1}{2} \frac{\epsilon}{\mu} : \frac{1}{2} \frac{\mu}{\epsilon}$$

$$= 1 : 1$$

$$\Rightarrow \boxed{\frac{1}{2} \epsilon E^2 = \frac{1}{2} \mu H^2} \quad \text{Electrical energy} = \text{Magnetic Energy}$$

$$\Rightarrow \frac{\partial}{\partial t} \int_V \left( \frac{1}{2} \mu H^2 + \frac{1}{2} \epsilon E^2 \right) dV$$

$$= \frac{\partial}{\partial t} \int_V \mu H^2 dV = \frac{\partial}{\partial t} \int_V \epsilon E^2 dV$$

If  $\sigma = 0$ ,

$$\int \vec{s} \cdot d\vec{A} = -\frac{\partial}{\partial t} \int \epsilon E^2 dV = -\frac{\partial}{\partial t} \int U dV$$

$$\Rightarrow \int_V \vec{\nabla} \cdot \vec{s} dV = \int -\frac{\partial U}{\partial t} dV$$

\* U: energy per unit volume  
=  $\epsilon E^2$

$$\Rightarrow \boxed{\vec{\nabla} \cdot \vec{s} + \frac{\partial U}{\partial t} = 0}$$

← that why called Energy Continuity Theorem

$$\vec{S} = \vec{E} \times \vec{H}$$

$$= \frac{1}{\mu} (\vec{E} \times \vec{B})$$

$$\boxed{\vec{S} = \frac{1}{\mu} \frac{E^2}{\omega} \vec{k}}$$

$$|\vec{S}| = \frac{E^2 |\vec{k}|}{\mu \omega} = \frac{E^2 \sqrt{\mu \epsilon}}{\mu} = \left[ \frac{E^2}{Z} \right]$$

$$\boxed{|\vec{S}| = \frac{E^2}{Z} = \frac{H^2 Z}{\bullet}} \quad \checkmark$$

For free space

$$|\vec{S}| = \left( \frac{E^2}{Z_0} \right) = E^2 \sqrt{\frac{\epsilon_0}{\mu_0}} = \epsilon_0 c E^2$$

$$\Rightarrow \begin{array}{l} |\vec{S}| = \epsilon_0 c E^2 \\ |\vec{S}| = c U \end{array}$$

instantaneous

$$\boxed{\begin{array}{l} \text{total energy per unit} \\ \text{volume} = \epsilon E^2 \end{array}}$$

We know,

$$E = E_0 e^{-i(\vec{k} \cdot \vec{r} - \omega t)}$$

We can write,  $E = E_0 \sin(\vec{k} \cdot \vec{r} - \omega t)$  or  $E = E_0 \cos(\vec{k} \cdot \vec{r} - \omega t)$

$$\Rightarrow |\vec{S}|_{\text{avg}} = \epsilon_0 c E_0^2 \langle \sin^2(\vec{k} \cdot \vec{r} - \omega t) \rangle$$

$$\boxed{|\vec{S}|_{\text{avg}} = \frac{1}{2} c \epsilon_0 E_0^2} \quad \text{average}$$

$$\Rightarrow \boxed{|\vec{S}|_{\text{avg}} = \bullet c U_{\text{avg}}}$$

# Continuity Equation

7.1

What precisely does conservation of charge means:

- ① Global Conservation: Total charge in the universe is Const.
- ② Local Conservation: If the total charge in some volume changes, then exactly that amount of charge must have passed in or passed out of the volume through the surface.

$$\text{i.e. } \frac{dQ}{dt} = - \oint_S \vec{J} \cdot d\vec{a}$$

$$\Rightarrow \int_V \left( \frac{\partial \rho}{\partial t} \right) d\tau = - \int_V (\vec{\nabla} \cdot \vec{J}) \cdot d\tau$$

$$\Rightarrow \boxed{\vec{\nabla} \cdot \vec{J} = - \left( \frac{\partial \rho}{\partial t} \right)}$$

Note that it is not an independent assumption but a consequence of the laws of electrodynamics.

$$\text{i.e. From 4th eqn: } \vec{\nabla} \times \vec{B} = \mu_0 \vec{J} + \mu_0 \epsilon_0 \left( \frac{\partial \vec{E}}{\partial t} \right)$$

Taking divergence,

$$0 = \mu_0 \cdot \vec{\nabla} \cdot \vec{J} + \mu_0 \epsilon_0 \cdot \frac{\partial}{\partial t} \cdot \frac{\rho}{\epsilon_0}$$

$$\Rightarrow \boxed{\vec{\nabla} \cdot \vec{J} = - \left( \frac{\partial \rho}{\partial t} \right)}$$

## Poynting's Theorem

It's the energy conservation theorem of electrodynamics. It states that work done on the charges by the electromagnetic force is equal to the decrease in energy stored in the fields, less the energy that flowed out through the surface.

$$\int \mathbf{E} \cdot \mathbf{J} \, d\tau = - \frac{\partial}{\partial t} \int \left( \frac{1}{2} \epsilon_0 E^2 + \frac{1}{2} \frac{B^2}{\mu_0} \right) d\tau - \oint \frac{\mathbf{E} \times \mathbf{B}}{\mu} \cdot d\vec{a}$$

For Understanding

similar to 1<sup>st</sup> law of Thermodynamics

$$Q = dU + P \, dV$$

$$\begin{array}{c} \uparrow \\ - \oint_S \mathbf{E} \times \mathbf{H} \cdot d\vec{a} \\ \text{(energy coming in)} \end{array} \quad \uparrow \quad + \frac{\partial}{\partial t} \int_V \left( \frac{1}{2} \epsilon_0 E^2 + \frac{1}{2} \frac{B^2}{\mu_0} \right) d\tau \quad \uparrow \quad \int_V \mathbf{E} \cdot \mathbf{J} \, d\tau \quad \text{(work done)}$$

(change in energy)

## Electromagnetic Waves

### General Wave

A wave is a disturbance of a continuous medium that propagates with a fixed shape at constant velocity.

Mathematically, the equation for a wave is

$$f(x, t) = f(x - vt, 0) = \psi(x - vt)$$

For a wave function that will satisfy the below relation can be represented as a wave, as it can be represented in a relation  $\psi(x - vt)$ .

$$\left( \frac{\partial^2 \psi}{\partial x^2} \right) = \frac{1}{v^2} \left( \frac{\partial^2 \psi}{\partial t^2} \right)$$

i.e.  $\psi$  should be a function that depends on  $(x, t)$  in the special combination  $(x - vt)$

But its nicer to write the waves in terms of  $\omega$  rather than  $v$ , in the form

$$\psi(kx - \omega t)$$

where  $k = \frac{2\pi}{\lambda}$  : wave number

$$\left(\frac{\omega}{k}\right) = v : \text{velocity of wave}$$

$\omega$ : angular frequency of the wave

### Complex Notation of a general wave

Now one possible solution of wave is

$$\psi(x, t) = A \cos(kx - \omega t + \delta)$$

Also from Fourier analysis we know that any wave can be linear as linear combination of sinusoidal waves, and therefore, it is sufficient for us to analyze just sinusoidal waves.

Now we know exponentials are much easier to manipulate than sines & cosines.

$\therefore$  introducing  $\tilde{\psi}(x, t) = \tilde{A} e^{i(kx - \omega t)}$  : Complex Wave Function

where  $\tilde{A} = A e^{i\delta}$  : complex Amplitude

$$\psi(x, t) = \text{Re}[\tilde{\psi}(x, t)]$$

The actual wave function is the real part of complex wave function.

Remember that complex exponentials are simply for mathematical ease.

Real wave is always real part of the complex wave.

## Some more points on field inside conductors

As we know waves (electric field) cannot reside inside a conductor, we will have an attenuated wave.

where  $\tilde{k}$  is ~~complex~~ complex

$$\tilde{k} = \alpha + i\beta$$

$$\alpha = \omega \sqrt{\frac{\epsilon\mu}{2}} \left[ \sqrt{1 + \left(\frac{\sigma}{\epsilon\omega}\right)^2} + 1 \right]^{1/2}$$

$$\beta = \omega \sqrt{\frac{\epsilon\mu}{2}} \left[ \sqrt{1 + \left(\frac{\sigma}{\epsilon\omega}\right)^2} - 1 \right]^{1/2}$$

$$\tilde{E} = \tilde{E}_0 e^{-\beta x} e^{i(\alpha x - \omega t)}$$

$$\tilde{B} = \tilde{B}_0 e^{-\beta x} e^{i(\alpha x - \omega t)}$$

From Maxwell equations

$$\tilde{B}_0 = \frac{|\tilde{k}|}{\omega} \tilde{E}_0$$

Now since  $k$  is complex,  $B$  and  $E$  are not longer in same phase.

write  $k = \sqrt{\alpha^2 + \beta^2} \angle \tan^{-1}\left(\frac{\beta}{\alpha}\right)$

$$\checkmark \frac{B_0}{E_0} = \frac{|\tilde{k}|}{\omega} = \sqrt{\epsilon\mu} \sqrt{1 + \left(\frac{\sigma}{\epsilon\omega}\right)^2}$$

Note that we have  $e^{i(\alpha x - \omega t)}$   
 $\therefore$  for additional  $e^{i\theta}$   
where  $\theta = \tan^{-1}(\beta/\alpha)$ , we have  
 $e^{i(\alpha x - \omega t + \theta)}$   
 $= e^{i(\alpha x - \omega(t - \frac{\theta}{\omega}))}$   
 $= \text{LAG by } t_0 = \frac{\theta}{\omega} \text{ i.e. } \phi = \theta$

$\checkmark$  Magnetic field lags electric field by  $\tan^{-1}\left(\frac{\beta}{\alpha}\right)$

eg. for good conductors ( $\alpha = \beta$ ), magnetic field lags the electric field by  $45^\circ$ . For any conductor, lag is something in between  $0^\circ$  to  $45^\circ$ .

Q16)

$$\begin{aligned}\vec{B} &= \frac{\vec{k} \times \vec{E}}{\omega} \\ &= \frac{k\hat{z}}{\omega} \times E_0 \cos(kz - \omega t + \delta) \hat{x} \\ &= \frac{kE_0}{\omega} \cos(kz - \omega t + \delta) \hat{y}\end{aligned}$$

$$\begin{aligned}U &= \frac{1}{2} \epsilon E^2 + \frac{1}{2} \mu H^2 \\ &= \epsilon E^2 \\ &= \epsilon^2 E_0^2 \cos^2(kz - \omega t + \delta)\end{aligned}$$

$$\begin{aligned}S &= vU \\ &= \frac{\omega}{k} \epsilon^2 E_0^2 \cos^2(kz - \omega t + \delta)\end{aligned}$$

$$\langle S \rangle = \frac{1}{2} \frac{\omega}{k} \epsilon^2 E_0^2$$

Q17) Assuming Area = Circle =  $\pi R^2$

$$S = \left( \frac{\text{Power}}{\text{Area}} \right)$$

$$\begin{aligned}\langle S \rangle &= \frac{1}{2} \frac{E_0^2}{Z_0} \Rightarrow E_0 = \sqrt{2Z_0 \langle S \rangle} \\ &= \sqrt{\frac{2Z_0 \text{ Power}}{\text{Area}}}\end{aligned}$$

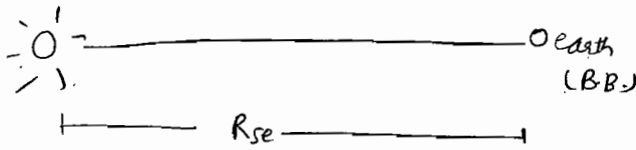
$$H_0 = \frac{E_0}{Z_0} = \sqrt{\frac{2 \text{ Power}}{Z_0 \text{ Area}}}$$



Pressure =  $U = \left(\frac{S}{c}\right)$  : Pressure  $\Leftrightarrow$  Energy per unit volume.

$= \frac{U}{3} = \left(\frac{S}{3c}\right)$

### Solar Const



$$R_{se} \approx 10^{11} \text{ m}$$

Solar Const. = Energy received by a B.B. kept at a distance of  $R_{\text{sun-earth}}$ , per second, with no intervening atmosphere, per unit area.  
(intensity)

$$\boxed{S} \cdot 4\pi R_{se}^2 = E = (\sigma T^4) (4\pi R_{\text{sun}}^2)$$

$\uparrow$  Energy per unit Time  
 $\uparrow$  Solar Const. (according to Thermodynamics)  
 $\uparrow$  Poynting Vector (according to EM theory)

T: Temperature of surface of sun

$$\Rightarrow \boxed{S = \sigma T^4 \left(\frac{R_{\text{sun}}}{R_{\text{sun-earth}}}\right)^2}$$

$$\boxed{1 \text{ cal} = 4.18 \text{ J}}$$

$$= \underline{1.36 \times 10^3 \text{ Watt/m}^2}$$

$$\approx 2 \text{ cal/cm}^2 \text{ minute}$$

easier to remember

→ Till now we have done analysis of dielectric medium and vacuum where non-attenuated propagation of electromagnetic wave takes place, as  $k$  is real. Now let's have a look at conducting media.

In conducting media,

$$\nabla^2 \psi - \mu\sigma \left(\frac{\partial \psi}{\partial t}\right) - \mu\epsilon \left(\frac{\partial^2 \psi}{\partial t^2}\right) = 0$$

Attenuated wave Equation

○ If  $\sigma = 0$ , We know, then  $\psi = \psi_0 e^{-i(\vec{k} \cdot \vec{r} - \omega t)}$ . Its easy

○ If  $\sigma \neq 0$ , then  $\vec{k}^*$  is complex &  $\psi = \psi_0 e^{-i(\vec{k}^* \cdot \vec{r} - \omega t)}$

→ We are assuming this to be the solution

Put  $\vec{k}^* = \vec{\alpha} + i\vec{\beta}$

$$\vec{\nabla} = i\vec{k}^*$$

$$\frac{\partial}{\partial t} = -i\omega$$

$$\Rightarrow \psi = \psi_0 e^{-\vec{\beta} \cdot \vec{r}} e^{-i(\vec{\alpha} \cdot \vec{r} - \omega t)}$$

★ Note that all characteristics of EM wave will be there in attenuated wave except that Amplitude will decrease with time.

Hence using these conversions, the attenuated wave eqn becomes,

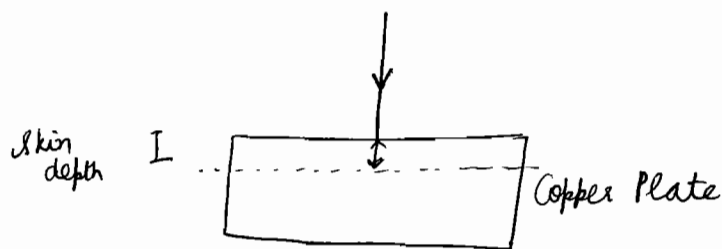
$$-k^{*2} + i\mu\sigma\omega + \mu\epsilon\omega^2 = 0$$

$$-(\alpha^2 - \beta^2 + 2i\alpha\beta) + i\mu\sigma\omega + \mu\epsilon\omega^2 = 0$$

$$\Rightarrow -(\alpha^2 - \beta^2 - \mu\epsilon\omega^2) + i(\mu\sigma\omega - 2\alpha\beta) = 0$$

$$\Rightarrow \alpha\beta = \left(\frac{\mu\sigma\omega}{2}\right) \quad \text{--- (1)}$$

$$\alpha^2 - \beta^2 = \mu\epsilon\omega^2 \quad \text{--- (2)}$$



✓ Skin Depth: distance after which field falls to  $\left(\frac{1}{e}\right)$  of its value in free space

$$\nabla^2 \vec{E} = \mu\epsilon \left(\frac{\partial^2 \vec{E}}{\partial t^2}\right) + \mu\sigma \left(\frac{\partial \vec{E}}{\partial t}\right)$$

(for conductors)

These equations still permit plane wave solutions but this time wave number is complex

$$E_0 e^{-\beta d} = \frac{1}{e} E_0$$

$$\Rightarrow \boxed{d = \frac{1}{\beta} = \text{Skin Depth}}$$

Solving ① and ② algebraically,

→ using

$$\alpha^2 - \left(\frac{\mu\sigma\omega}{2\alpha}\right)^2 = \mu\epsilon\omega^2$$

$$(\alpha^2 + \beta^2)^2 = (\alpha^2 - \beta^2)^2 + 4\alpha^2\beta^2$$

Calculation is drastically reduced

$$4\alpha^4 = \mu^2\sigma^2\omega^2 = 4\alpha^2\mu\epsilon\omega^2$$

$$\Rightarrow \alpha^4 - \alpha^2(\mu\epsilon\omega^2) - \frac{\mu^2\sigma^2\omega^2}{4} = 0$$

$$\Rightarrow \alpha^2 = \frac{+(\mu\epsilon\omega^2) \pm \sqrt{(\mu\epsilon\omega^2)^2 + \mu^2\sigma^2\omega^2}}{2}$$

$$= \frac{\mu\epsilon\omega^2 \pm \mu\omega\sqrt{(\epsilon\omega)^2 + \sigma^2}}{2}$$

$$= \frac{\mu\epsilon\omega^2 \pm \mu\epsilon\omega^2\sqrt{1 + \left(\frac{\sigma}{\epsilon\omega}\right)^2}}{2}$$

$$\Rightarrow \alpha = \left[\frac{\mu\epsilon\omega^2}{2}\right]^{\frac{1}{2}} \left[1 \pm \sqrt{1 + \left(\frac{\sigma}{\epsilon\omega}\right)^2}\right]^{\frac{1}{2}}$$

1st expression for  $\beta$

$$\beta = \frac{\mu\sigma\omega}{2} \sqrt{\frac{2}{\mu\epsilon\omega^2}} \left[1 \pm \sqrt{1 + \left(\frac{\sigma}{\epsilon\omega}\right)^2}\right]^{-\frac{1}{2}}$$

only +ve sign

$$\beta = \frac{\mu\sigma}{\sqrt{2\mu\epsilon}} \left[1 \pm \sqrt{1 + \frac{\sigma^2}{\epsilon^2\omega^2}}\right]^{-\frac{1}{2}}$$

2nd expression for good conductor,  $\frac{\sigma}{\epsilon\omega} \gg 1$

$$\beta = \left[\frac{\mu\epsilon\omega^2}{2}\right]^{\frac{1}{2}} \left[\sqrt{1 + \left(\frac{\sigma}{\epsilon\omega}\right)^2} - 1\right]^{\frac{1}{2}}$$

$$\Rightarrow \left[1 \pm \sqrt{1 + \frac{\sigma^2}{\epsilon^2\omega^2}}\right]^{\frac{1}{2}} \approx \left[1 \pm \frac{\sigma}{\epsilon\omega}\right]^{\frac{1}{2}} \approx \left(\frac{\sigma}{\epsilon\omega}\right)^{\frac{1}{2}}$$

$$\Rightarrow \alpha = \sqrt{\frac{\mu \epsilon \omega^2}{2}} \sqrt{\frac{\sigma}{\epsilon \omega}} = \sqrt{\frac{\mu \sigma \omega}{2}}$$

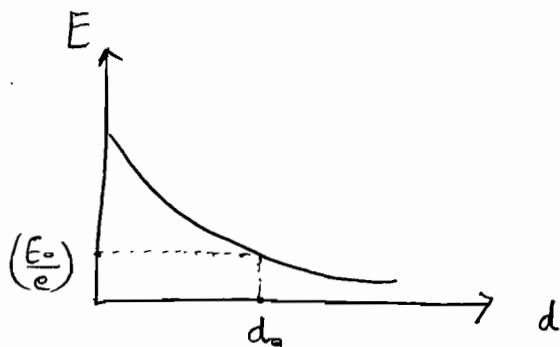
$$\beta = \frac{\mu \sigma}{\sqrt{2 \mu \epsilon}} \sqrt{\frac{\epsilon \omega}{\sigma}} = \sqrt{\frac{\mu \sigma \omega}{2}}$$

$$\Rightarrow \text{For good conductors, } \alpha = \beta = \sqrt{\frac{\mu \sigma \omega}{2}}$$

✓ As  $\sigma \rightarrow 0$ , i.e. for dielectric material

$\alpha \rightarrow \sqrt{\mu \epsilon \omega} = k$   
 $\beta \rightarrow 0$

Perfect



For a good conductor :-  $d_0$ : skin depth =  $\frac{1}{\beta} = \sqrt{\frac{2}{\mu \sigma \omega}}$

## Scalar and Vector Potentials [Electromagnetic]

Note that these Potentials are different from Electrostatic Potentials.

In electrostatics :-  $V = \frac{1}{4\pi\epsilon_0} \frac{q}{r}$  : Coulumb Potential we studied

$$dV = \int E \cdot dl, \quad \vec{E} = -\vec{\nabla}V, \quad \vec{\nabla} \times \vec{E} = 0$$

$$dV = 0 = \oint E \cdot dl$$

⊙ In electromagnetics, fields are changing w.r.t. time i.e. accelerated charge motion, hence completely different analysis, thereby giving different Potentials.

In Magnetostatics, we study	$\vec{B} = -\vec{\nabla} \phi_m$
	$\phi_m = \frac{\mu_0 I \Omega}{4\pi}$

From Maxwell equations,

$$\vec{\nabla} \cdot \vec{E} = \frac{\rho}{\epsilon_0}$$

$$\vec{\nabla} \cdot \vec{B} = 0$$

$$\vec{\nabla} \times \vec{E} = -\frac{\partial \vec{B}}{\partial t} = -\frac{\partial (\vec{\nabla} \times \vec{A})}{\partial t} = -\vec{\nabla} \times \left( \frac{\partial \vec{A}}{\partial t} \right)$$

From Biot Savart's law,  $\vec{B} = \vec{\nabla} \times \vec{A}$   
 where we define  $\vec{A}$  = EM vector Potential  
 if valid  $\rightarrow$   $\vec{A} = \frac{\mu_0}{4\pi} \int \frac{I d\vec{l}}{r}$

$$\Rightarrow \vec{\nabla} \times \left( \vec{E} + \frac{\partial \vec{A}}{\partial t} \right) = 0$$

Curl of Grad is zero

$$\Rightarrow \left[ \vec{E} + \left( \frac{\partial \vec{A}}{\partial t} \right) \right] \text{ can be written as } -\vec{\nabla} \phi$$

$\vec{E} + \frac{\partial \vec{A}}{\partial t} = -\vec{\nabla} \phi$	Electromagnetic Scalar Potential
--	-------------------------------------

Electromagnetic  
Vector Potential

$$\Rightarrow \vec{E} = -\vec{\nabla}\phi - \frac{\partial \vec{A}}{\partial t}$$

$$\vec{B} = \vec{\nabla} \times \vec{A}$$

EM Fields in terms of Potentials  
( $\vec{A}, \phi$ )

EM Potentials are mathematical constructs which are dependent on each other. They are used to measure  $\vec{E}$  and  $\vec{B}$ . Note that  $\vec{E}$  and  $\vec{B}$  are dependent upon scale. Unlike ~~potentials~~ potentials that are scale invariant. Physical measurement of  $\phi$  and  $\vec{A}$  is not, thereby, important as they are scale invariant, but their variation is important as it gives absolute value of  $\vec{E}$  &  $\vec{B}$ .  
To prove  $\vec{E}$  &  $\vec{B}$  are scale invariant even if  $\phi$  and  $A$  are changed

$$\text{let } \phi' = \phi + \left(\frac{\partial \lambda}{\partial t}\right)$$

$$\Rightarrow \phi' = \phi + \left(\frac{\partial \lambda}{\partial t}\right)$$

when  $\lambda$  is constant scalar quantity.

$$\text{let } \vec{A}' = \vec{A} + \vec{\nabla} \lambda$$

$$\Rightarrow \vec{A}' = \vec{A} + \vec{\nabla} \lambda$$

$$\begin{aligned} \vec{B}' &= \vec{\nabla} \times (\vec{A}') \\ &= \vec{\nabla} \times (\vec{A} + \vec{\nabla} \lambda) \\ &= \vec{\nabla} \times \vec{A} + \vec{\nabla} \times (\vec{\nabla} \lambda) \\ &= \vec{\nabla} \times \vec{A} \\ &= \vec{B} \end{aligned}$$

We can get at these forms

$$\vec{E}' = -\vec{\nabla} \phi' - \frac{\partial \vec{A}'}{\partial t} = -\vec{\nabla} \left( \phi + \frac{\partial \lambda}{\partial t} \right) - \frac{\partial (\vec{A} + \vec{\nabla} \lambda)}{\partial t}$$

$$= -\vec{\nabla} \phi - \frac{\partial \vec{A}}{\partial t}$$

$$= \vec{E}$$

Note that upon coupling 4 Maxwell equations, we get 2 equations only :

$$\left( \nabla^2 - \frac{1}{c^2} \frac{\partial^2}{\partial t^2} \right) \vec{E} = 0$$

$$\left( \nabla^2 - \frac{1}{c^2} \frac{\partial^2}{\partial t^2} \right) \vec{B} = 0$$

Recasting Maxwell equations in terms of  $(\vec{A}, \phi)$  :

$$\vec{\nabla} \cdot \vec{E} = \frac{\rho}{\epsilon_0} \quad \text{--- (1)}$$

$$\vec{\nabla} \cdot \vec{B} = 0 \quad \text{--- (2)}$$

$$\vec{\nabla} \times \vec{E} = -\left( \frac{\partial \vec{B}}{\partial t} \right) \quad \text{--- (3)}$$

$$\vec{\nabla} \times \vec{B} = \mu \left( \vec{J} + \epsilon \frac{\partial \vec{E}}{\partial t} \right) \quad \text{--- (4)}$$

From (4),

$$\vec{\nabla} \times (\vec{\nabla} \times \vec{A}) = \mu \left( \vec{J} + \epsilon \frac{\partial}{\partial t} \left( -\vec{\nabla} \phi - \frac{\partial \vec{A}}{\partial t} \right) \right)$$

$$\vec{\nabla} (\vec{\nabla} \cdot \vec{A}) - \vec{A} (\vec{\nabla} \cdot \vec{\nabla}) + \mu \epsilon \left( \frac{\partial^2 \vec{A}}{\partial t^2} \right) + \mu \epsilon \vec{\nabla} \left( \frac{\partial \phi}{\partial t} \right) = \mu \vec{J}$$

$$\boxed{\nabla^2 \vec{A} - \mu \epsilon \left( \frac{\partial^2 \vec{A}}{\partial t^2} \right) - \vec{\nabla} \left( \vec{\nabla} \cdot \vec{A} + \mu \epsilon \frac{\partial \phi}{\partial t} \right) = -\mu \vec{J}} \quad \text{--- (5)}$$

$$\Rightarrow \left( \nabla^2 - \mu \epsilon \frac{\partial^2}{\partial t^2} \right) \vec{A} - \vec{\nabla} L = -\mu \vec{J}$$

↑  
de Alembert

$$\text{where } L = \vec{\nabla} \cdot \vec{A} + \mu \epsilon \left( \frac{\partial \phi}{\partial t} \right)$$

$$\vec{\nabla} \cdot \left( -\vec{\nabla} \phi - \frac{\partial \vec{A}}{\partial t} \right) = \frac{\rho}{\epsilon_0}$$

From ①

$$-\nabla^2 \phi - \frac{\partial}{\partial t} (\vec{\nabla} \cdot \vec{A}) = \frac{\rho}{\epsilon_0}$$

Writing again in the same form as ⑤ by adding & subtracting  $\mu \epsilon \frac{\partial \phi}{\partial t}$

$$\nabla^2 \phi + \mu \epsilon \left( \frac{\partial^2 \phi}{\partial t^2} \right) + \frac{\partial}{\partial t} (\vec{\nabla} \cdot \vec{A}) = \frac{\rho}{\epsilon_0}$$

$$\Rightarrow \left( \nabla^2 - \mu \epsilon \frac{\partial^2}{\partial t^2} \right) \phi + \frac{\partial}{\partial t} \left[ \vec{\nabla} \cdot \vec{A} + \mu \epsilon \frac{\partial \phi}{\partial t} \right] = \frac{\rho}{\epsilon_0} \quad \text{--- ⑥}$$

$$\Rightarrow \left( \nabla^2 - \mu \epsilon \frac{\partial^2}{\partial t^2} \right) \phi + \frac{\partial}{\partial t} L = \frac{\rho}{\epsilon_0}$$

Lorentz Gauge Equation :  $L = \vec{\nabla} \cdot \vec{A} + \mu \epsilon \left( \frac{\partial \phi}{\partial t} \right) = 0$

$$\Rightarrow \begin{cases} \square^2 \vec{A} = -\mu \vec{J} \\ \square^2 \phi = -\frac{\rho}{\epsilon_0} \end{cases} \quad \text{--- ⑦}$$

Lorentz Gauge Condition is Gauge Invariant i.e.

$$\vec{\nabla} \cdot \vec{A} + \mu \epsilon \left( \frac{\partial \phi}{\partial t} \right) = 0 = \vec{\nabla} \cdot \vec{A}' + \mu \epsilon \left( \frac{\partial \phi'}{\partial t} \right) = 0$$

We can always

Choose  $\lambda$  s.t.  $\nabla^2 \lambda - \mu \epsilon \left( \frac{\partial^2 \lambda}{\partial t^2} \right) = 0$

s.t.  $\Rightarrow \underline{\underline{\square^2 \lambda = 0}}$



From ①

$$\nabla^2 \phi + \frac{\partial}{\partial t} (\nabla \cdot \vec{A}) = -\frac{\rho}{\epsilon_0}$$

If I choose  $\nabla \cdot \vec{A} = 0$ , it becomes Poisson Equation of Electrostatic &  $\phi$  becomes  $V$

$$\boxed{\nabla \cdot \vec{A} = 0 : \text{Coulumb Gauge Condition}}$$

↓  
changing EM scalar Potential to  
Electrostatic scalar Potential (Coulumb Potential)

Writing ⑤ and ⑥ under Coulumb Gauge Condition

$$\boxed{\begin{aligned} \square^2 \vec{A} - \mu \epsilon \nabla \left( \frac{\partial \phi}{\partial t} \right) &= -\mu \vec{J} \\ \nabla^2 \phi &= -\frac{\rho}{\epsilon_0} \end{aligned}} \quad \text{--- ⑧}$$

In relativistic domain,

[only Lorentz gauge is  
gauge invariant]

$$\boxed{\begin{aligned} J_\mu &= (\vec{J}, ic\rho) \\ A_\mu &= (\vec{A}, \frac{i\phi}{c}) \\ \square^2 A_\mu &= -\mu_0 J_\mu \end{aligned}} \quad \checkmark$$

$$\begin{aligned} \therefore \square^2 \vec{A} &= -\mu_0 \vec{J} \\ \square^2 \frac{i\phi}{c} &= -\frac{\rho}{c\epsilon_0} \end{aligned}$$

from  
Lorentz  
gauge

# Potentials

~~Q~~ Why do we do the Crappy Potentials?

~~A~~ By the help of these potentials, we seek that how the sources ( $\rho$  and  $\vec{J}$ ) generate electric and magnetic fields, in other words, we seek the general solution of the Maxwell equations. Given  $\rho(\vec{r}, t)$  and  $\vec{J}(\vec{r}, t)$  what are the fields  $\vec{E}(\vec{r}, t)$  and  $\vec{B}(\vec{r}, t)$ ?

In the static case, Coulomb's law and the Biot-Savart's law provide the answers. What we are looking for, then, is the generalization of these laws to time-dependent configurations.

~~Q~~ Definition of Potentials?

~~A~~ In electrostatics  $\nabla \times \vec{E}$  was 0 and that allowed us to write  $\vec{E}$  as  $-\vec{\nabla}V$  but here curl is non zero. But here we still have  $\nabla \cdot \vec{B} = 0$

$\Rightarrow$  We can still write

$$\boxed{\vec{B} = \vec{\nabla} \times \vec{A}} \quad \text{--- (1)}$$

$$\Rightarrow \vec{\nabla} \times \vec{E} = -\left(\frac{\partial \vec{B}}{\partial t}\right) = -\vec{\nabla} \times \frac{\partial \vec{A}}{\partial t}$$

$$\Rightarrow \vec{\nabla} \times \left( \vec{E} + \frac{\partial \vec{A}}{\partial t} \right) = 0$$

Now, this quantity  $\vec{E} + \left(\frac{\partial \vec{A}}{\partial t}\right)$ , instead of  $\vec{E}$  alone, satisfies condition for being gradient of a potential

$$\Rightarrow \boxed{\vec{E} + \frac{\partial \vec{A}}{\partial t} = -\vec{\nabla} \phi} \quad \text{--- (2)}$$

Now from remaining two maxwell equations,

$$\nabla \cdot \vec{E} = \frac{\rho}{\epsilon_0}$$

$$\Rightarrow \boxed{\nabla^2 \phi + \frac{\partial}{\partial t} (\nabla \cdot \vec{A}) = -\frac{\rho}{\epsilon_0}} \quad \text{--- (3)}$$

$$\nabla \times \vec{B} = \mu_0 \vec{J} + \mu_0 \epsilon_0 \left( \frac{\partial \vec{E}}{\partial t} \right)$$

$$\Rightarrow \nabla \times (\nabla \times \vec{A}) = \mu_0 \vec{J} - \mu_0 \epsilon_0 \nabla \left( \frac{\partial \phi}{\partial t} \right) - \mu_0 \epsilon_0 \frac{\partial^2 \vec{A}}{\partial t^2}$$

$$\Rightarrow \boxed{\left( \nabla^2 \vec{A} - \mu_0 \epsilon_0 \frac{\partial^2 \vec{A}}{\partial t^2} \right) - \nabla \left( \nabla \cdot \vec{A} + \mu_0 \epsilon_0 \frac{\partial \phi}{\partial t} \right) = -\mu_0 \vec{J}}$$

--- (4)

Now (3) and (4) represent all the information of Maxwell Equations in terms of Potentials.

Note that we have reduced the problem of finding 6 variables (3 components each of  $\vec{E}$  &  $\vec{B}$ ) to 4 variables ( $\phi$  and  $\vec{A}$ )

## Gauge Transformations

In order to further simplify the analysis, we note that equations (1) and (2) do not uniquely define the potentials; we are free to impose extra conditions on  $\phi$  and  $\vec{A}$ , as long as, nothing happens to  $\vec{E}$  and  $\vec{B}$ , thereby simplifying our analysis.

Suppose we have two sets of potentials ( $\phi, \vec{A}$ ) and ( $\phi', \vec{A}'$ ), that correspond to same electric and magnetic fields: By how much can they differ?

$$\text{let } \vec{A}' = \vec{A} + \alpha$$

$$\phi' = \phi + \beta$$

since  $\vec{B}$  is same

$$\nabla \times \vec{A}' = \nabla \times \vec{A} + \nabla \times \alpha \Rightarrow \nabla \times \alpha = 0$$

$$\Rightarrow \text{let } \boxed{\alpha = \nabla \lambda}$$

since  $\vec{E}$  is same,

$$-\nabla \phi - \frac{\partial \vec{A}}{\partial t} = -\nabla \phi' - \frac{\partial \vec{A}'}{\partial t}$$

$$\Rightarrow 0 = \nabla \left( \beta + \frac{\partial \lambda}{\partial t} \right)$$

$$\Rightarrow \beta = -\left( \frac{\partial \lambda}{\partial t} \right)$$

$$\Rightarrow \vec{A}' = \vec{A} + \nabla \lambda$$

$$\phi' = \phi - \left( \frac{\partial \lambda}{\partial t} \right)$$

This implies that for any scalar function  $\lambda$ , we can add  $\nabla \lambda$  to  $\vec{A}$ , provided we simultaneously subtract  $\left( \frac{\partial \lambda}{\partial t} \right)$  from  $\phi$ . None of this will affect the physical quantities  $\vec{E}$  and  $\vec{B}$ . Such changes in  $\phi$  and  $\vec{A}$  are called Gauge Transformations. They can be exploited to adjust the divergence of  $\vec{A}$ , with a view to simplify ugly equations (3) and (4)

There are two famous gauges in physics:  
Coulomb Gauge and Lorentz Gauge

## Coulumb Gauge

As done in magnetostatics, but

$$\boxed{\vec{\nabla} \cdot \vec{A} = 0}$$

This simplifies (3) to simple Poisson equation in electrodynamics

$$\nabla^2 \phi = -\left(\frac{\rho}{\epsilon_0}\right)$$

$$\phi(\mathbf{r}, t) = \frac{1}{4\pi\epsilon_0} \int \frac{\rho(\mathbf{r}', t')}{|\mathbf{r} - \mathbf{r}'|} d\tau'$$

But we shouldn't be too happy as  $\phi$  won't give  $\vec{E}$  (we have to know  $\vec{A}$  as well).

Also the (4) equation still remains quite ugly.

## Lorentz Gauge

Here we pick

$$\boxed{\vec{\nabla} \cdot \vec{A} = -\mu_0 \epsilon_0 \left(\frac{\partial \phi}{\partial t}\right)} \quad (5)$$

The virtue of Lorentz Gauge is that it puts  $\phi$  and  $\vec{A}$ , both on an equal footing.

$$\square^2 \phi = \frac{-\rho}{\epsilon_0} \quad - (6)$$

$$\square^2 \vec{A} = -\mu_0 \vec{J} \quad - (7)$$

Lorentz Gauge is usually dealt with.

## Retarded Potentials

In static case, Lorentz gauge equations will lead

to  $\uparrow$   $\nabla^2 \phi = \frac{-\rho}{\epsilon_0}$        $\nabla^2 \vec{A} = -\mu_0 \vec{J}$

[ $\rho$  and  $\vec{J}$  are independent of time]

These have standard solutions,

$$\phi(\mathbf{r}) = \frac{1}{4\pi\epsilon_0} \int \frac{\rho(\mathbf{r}')}{r-r'} d\tau'$$

$$\vec{A}(\mathbf{r}) = \frac{\mu_0}{4\pi} \int \frac{\vec{J}(\mathbf{r}')}{r-r'} d\tau'$$

However in non static case, as we know electromagnetic news travels at the speed of light, it's not the status of sources right now, but rather their conditions at some time earlier, called the retarded time  $\left[ t = \left( \frac{|\mathbf{r}-\mathbf{r}'|}{c} \right) \right]$  when the message left. So we can safely conclude (we can verify that they satisfy (6) and (7)) that,

$$\phi(\vec{\mathbf{r}}, t) = \frac{1}{4\pi\epsilon_0} \int \frac{\rho(\vec{\mathbf{r}}', \frac{ct-\sigma}{c})}{\sigma} d\tau' \quad \sigma = |\vec{\mathbf{r}}-\vec{\mathbf{r}}'|$$

$$\vec{A}(\vec{\mathbf{r}}, t) = \frac{\mu_0}{4\pi} \int \frac{\vec{J}(\vec{\mathbf{r}}', \frac{ct-\sigma}{c})}{\sigma} d\tau'$$

Because integrands are evaluated at the retarded time, they are called retarded potentials.

### A beautiful thing

Incidentally, the advanced potentials also satisfy equation (6) and (7), i.e.  $\rho(\vec{\mathbf{r}}', t + \frac{\sigma}{c})$  &  $\vec{J}(\vec{\mathbf{r}}', t + \frac{\sigma}{c})$ . This is because the d'Alembertian involves  $t^2$  (as opposed to  $t$ ), the theory itself is time-reversal invariant. And does not distinguish "past" from "future". Although the

Advanced potentials are entirely consistent with Maxwell's equations, they violate the most sacred tenet in all of the physics: THE PRINCIPLE OF CAUSALITY. They suggest that the potentials "now" depend on what the charge and the current distribution "will" be at some point in future - the effect precedes the cause. Therefore, although the advanced potentials are of some theoretical interest, they have no direct, yet known, physical significance.

So therefore as we have calculated  $\phi$  as well as  $\vec{A}$ , we can with the help of some mathematics, we are in a position to find out  $\vec{E}$  and  $\vec{B}$ .

○ In Lorentz gauge condition,  $\square^2 A = -\mu_0 \vec{J}$   
 $\square^2 \phi = -\frac{\rho}{\epsilon_0}$

Note that  $\rho$  is source of electric field and  $\vec{J}$  is the source of magnetic field.

In a source free region,  $\rho = 0$ ,  $\vec{J} = 0$

we have  $\square^2 \vec{A} = 0$

$\square^2 \phi = 0$

### Retarded Potentials

○ Looking for solutions of Maxwell equations in potential form:

$$\square^2 \vec{A} = -\mu_0 \vec{J}$$

$$\vec{A}(\vec{r}, t) = \frac{\mu_0}{4\pi} \int \frac{\vec{J}(\vec{r}', t - \frac{R}{c})}{R} dV$$

(solution of the eqn)

$$= \frac{\mu_0}{4\pi} \int \frac{\vec{J}(\vec{r}', t - \frac{R}{c})}{R} d\vec{r}'$$

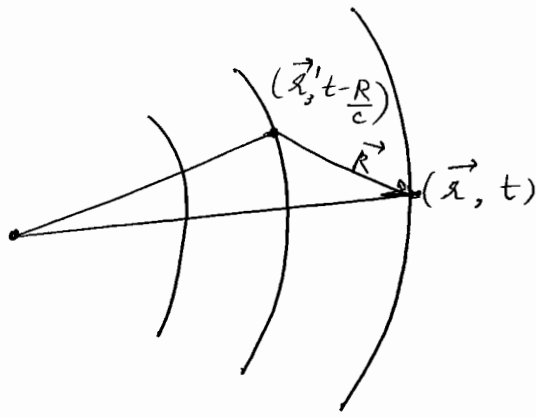
where  $R = |\vec{r} - \vec{r}'|$

$$\square^2 \phi = -\frac{\rho}{\epsilon_0}$$

$$\phi(\vec{r}, t) = \frac{1}{4\pi\epsilon_0} \int \frac{\rho(\vec{r}', t - \frac{R}{c})}{R} dV$$

Note that  $\vec{E}$ ,  $\vec{B}$  at  $(\vec{r}, t)$  are not due to present positions of charge & current. Present values of Potentials are due to past values of charge, current. Hence solutions are called retarded POTENTIALS.





Solutions are like if wavefront of EM wave is travelling. Present values of  $\vec{A}$  and  $\phi$  are due to past positions of current & charge respectively, i.e. field parameters  $(\vec{E}, \vec{B})$  are not instantaneously set up.

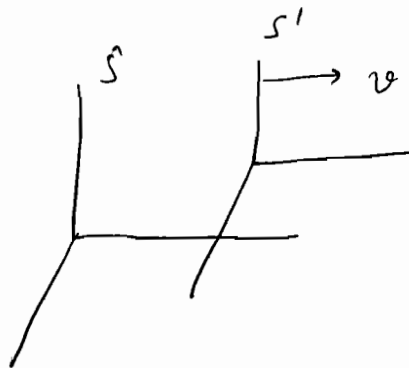
Note that if we are given  $\vec{A}$ , in order to calculate  $\vec{J}$  we can use

$$\nabla^2 \vec{A} - \frac{1}{c^2} \frac{\partial^2 \vec{A}}{\partial t^2} = -\mu_0 \vec{J}$$

But we need to mention that it is the solution under Lorentz gauge condition. ✓

## Covariance of EM waves

From STR, we know



$$x_\mu = (x, y, z, ict)$$

$$\alpha_{\mu\nu} = \begin{bmatrix} \alpha & 0 & 0 & i\beta \\ 0 & 1 & 0 & 0 \\ 0 & 0 & 1 & 0 \\ -i\beta & 0 & 0 & \alpha \end{bmatrix}$$

where  $\alpha = \frac{1}{\sqrt{1 - \frac{v^2}{c^2}}}$        $\beta = \left(\frac{v}{c}\right)$

- o  $q$  is Lorentz invariant
- o Charge density is not invariant

o Consider a stationary charge dist<sup>n</sup> in frame  $S$

In this frame, since the charge is stationary, it will appear as charge dist<sup>n</sup> but in  $S'$  it will appear in motion and will appear as current distribution.

We have 2 new 4-vectors:

$$1) \vec{J}_\mu = (\vec{J}, ic\rho)$$

$$2) \vec{A}_\mu = (\vec{A}, i\frac{\phi}{c})$$

$$J^2 - c^2\rho^2 : \text{invariant}$$

$$A^2 - \frac{\phi^2}{c^2} : \text{invariant}$$

From equation of continuity, we know,

$$\vec{\nabla} \cdot \vec{J} + \left(\frac{\partial \rho}{\partial t}\right) = 0$$

i.e. writing in 4 coordinates

$$\frac{\partial J_1}{\partial x_1} + \frac{\partial J_2}{\partial x_2} + \frac{\partial J_3}{\partial x_3} + \frac{\partial \rho \times ic}{\partial t \times ic} = 0$$

$$(x_4 = ict)$$

$$(J_4 = ic\rho)$$

$$\Rightarrow \boxed{\frac{\partial J_1}{\partial x_1} + \frac{\partial J_2}{\partial x_2} + \frac{\partial J_3}{\partial x_3} + \frac{\partial J_4}{\partial x_4} = 0}$$

We know  $\vec{J} = \rho \vec{v}$

$$\boxed{\sum_{\mu=1}^4 \frac{\partial J_\mu}{\partial x_\mu} = 0}$$

Continuity Equation

This continuity equation is invariant.

This means, even from  $S'$ ,

$$\sum_{\mu=1}^4 \frac{\partial J_{\mu}'}{\partial x_{\mu}'} = 0$$

✓ Charge sets electric field in its own frame and magnetic field in the other frame.

$$dJ_{\mu}' = \sum_{\nu} \alpha_{\mu\nu} dJ_{\nu}$$

$$dx_{\mu}' = \sum_{\nu} \alpha_{\mu\nu} dx_{\nu}$$

$$\left[ \frac{dJ_{\mu}'}{dx_{\mu}'} = \frac{dJ_{\mu}}{dx_{\mu}} \Rightarrow \underline{\text{Invariance of continuity equation}} \right]$$

To prove invariance of  $\vec{J}_{\mu}$

$$\text{Norm}(\vec{J}_{\mu}) = J^2 - c^2 \rho^2$$

$$\Rightarrow \text{we need to show } J_1^2 - c^2 \rho^2 = J_1'^2 - c^2 \rho'^2$$

To prove invariance of Lorentz Gauge Condition

$$\vec{\nabla} \cdot \vec{A} + \frac{1}{c^2} \frac{\partial \phi}{\partial t} = 0$$

$$\Rightarrow \vec{\nabla} \cdot \vec{A} + \frac{\partial(\frac{i\phi}{c})}{\partial(ict)} = 0$$

$$\Rightarrow \boxed{\vec{\nabla} \cdot \vec{A} + \frac{\partial(\frac{i\phi}{c})}{\partial x_4} = 0} \Rightarrow \underline{\sum_{\mu=1}^4 \frac{\partial A_{\mu}}{\partial x_{\mu}} = 0}$$

$$\text{define } \vec{A}_{\mu} = (\vec{A}, \frac{i\phi}{c})$$

hence Lorentz invariant

$\vec{A}$ ,  $\phi$  are not independent

$$\square^2 \vec{A} = -\mu_0 \vec{J} \quad \text{--- (1)}$$

$$\square^2 \phi = -\rho/\epsilon_0$$

$$\Rightarrow \square^2 \left( \frac{i\phi}{c} \right) = -\frac{i\rho \mu_0}{c\epsilon_0 \mu_0} = -\mu_0 (ic\rho)$$

$$\square^2 A_4 = -\mu_0 J_4 \quad \text{--- (2)}$$

Combining (1) and (2),

$$\boxed{\square^2 A_\mu = -\mu_0 J_\mu}$$

Tensor Form of  
Maxwell Equations

Note  
that

$$\square^2 : \text{Lorentz invariant} \quad \square'^2 = \square^2$$

$$\square^2 = \nabla^2 - \frac{1}{c^2} \frac{\partial^2}{\partial t^2}$$

$$= \nabla^2 + \frac{\partial^2}{\partial (ict)^2} = \sum_{\mu=1}^4 \frac{\partial^2}{\partial x_\mu^2}$$

## Covariance of Maxwell Equations

Maxwell Equations in frame  $S$  are represented in tensor form.

If they retain their form in  $S'$ , then Maxwell equations are covariant.

$$\text{Given } \square^2 A_\mu = -\mu_0 J_\mu$$

$$\text{To prove } \square'^2 A'_\mu = -\mu_0 J'_\mu$$

$$\text{we know } \left\{ \begin{array}{l} A'_\mu = \alpha_{\mu\nu} A_\nu \\ \square'^2 = \square^2 \end{array} \right\}$$

$$\square'^2 A_{\mu}' = \square^2 (\alpha_{11} A_1 + \alpha_{14} A_4)$$

$$= \alpha_{11} \square^2 A_1 + \alpha_{14} \square^2 A_4$$

$$= +\alpha_{11} (-\mu_0 J_1) + \alpha_{14} -\mu_0 J_4$$

(Using Tensor form)

$$= -\mu_0 [\alpha_{11} J_1 + \alpha_{14} J_4]$$

$$= -\mu_0 J_1'$$

Similarly prove for rest.

$$\square'^2 A_2' = -\mu_0 J_2'$$

$$\square'^2 A_3' = -\mu_0 J_3'$$

$$\square'^2 A_4' = -\mu_0 J_4'$$

Hence  $\boxed{\square'^2 A_{\mu}' = -\mu_0 J_{\mu}'}$  ✓

So 5 things we have shown :-

- 1) equation of continuity is invariant
- 2) Lorentz gauge condition is invariant
- 3) Maxwell equations are invariant
- 4)  $\square^2$  is invariant
- 5)  $J_{\mu}$  is a 4 vector

① Charge is relativistically invariant. If charge was affected by motions, then we would not have exact cancellation of electronic and nuclear charge in atoms like Hydrogen or Helium.

$$\Rightarrow \int_A E \cdot dA \equiv \int_{A'} E' \cdot dA'$$

② Surface Charge density & Volume Charge density

$$\rho' = \frac{\rho}{\sqrt{1 - \frac{v^2}{c^2}}}$$

$$\sigma' = \frac{\sigma}{\sqrt{1 - \frac{v^2}{c^2}}}$$

### Relativistic Electrodynamics

We can express some equations of electromagnetism using 4-dimensional forms:

(1) Continuity Equation

$$\nabla \cdot \vec{J} + \left( \frac{\partial \rho}{\partial t} \right) = 0$$

$$\Rightarrow \frac{\partial J_x}{\partial x} + \frac{\partial J_y}{\partial y} + \frac{\partial J_z}{\partial z} + \frac{\partial (ic\rho)}{\partial (ict)} = 0$$

$$\Rightarrow \left( \frac{\partial J_1}{\partial x_1} \right) + \left( \frac{\partial J_2}{\partial x_2} \right) + \left( \frac{\partial J_3}{\partial x_3} \right) + \left( \frac{\partial J_4}{\partial x_4} \right) = 0$$

where  $\vec{J}_\mu = (J_1, J_2, J_3, ic\rho) = (\vec{J}, ic\rho)$  : 4 vector  
Current density

The quantity  $\left( \frac{\partial J_\mu}{\partial x_\mu} \right)$  is 4-dimensional divergence of  $\vec{J}_\mu$ , represented by  $\square$

$$\square \cdot \vec{J}_\mu = 0$$

## ② Lorentz Condition

$$\nabla \cdot \vec{A} + \frac{1}{c^2} \left( \frac{\partial \phi}{\partial t} \right) = 0$$

$$\text{or } \frac{\partial A_x}{\partial x} + \frac{\partial A_y}{\partial y} + \frac{\partial A_z}{\partial z} + \frac{\partial \left( \frac{i\phi}{c} \right)}{\partial (ict)} = 0$$

$$\Rightarrow \frac{\partial A_1}{\partial x_1} + \frac{\partial A_2}{\partial x_2} + \frac{\partial A_3}{\partial x_3} + \frac{\partial A_4}{\partial x_4} = 0$$

$\therefore \vec{A}$  and  $\phi$  combine to form a four vector potential  
or electromagnetic four potential

$$A_\mu = \left( \vec{A}, \frac{i\phi}{c} \right)$$

$$\& \boxed{\square \cdot \vec{A}_\mu = 0}$$

## ③ Wave Equation for $\vec{A}$ and $\phi$

$$\nabla^2 \vec{A} = \frac{1}{c^2} \frac{\partial^2 \vec{A}}{\partial t^2} = -\mu_0 \vec{J}$$

$$\nabla^2 \phi - \frac{1}{c^2} \frac{\partial^2 \phi}{\partial t^2} = -\frac{\rho}{\epsilon}$$

$$\boxed{\square^2 \vec{A}_\mu = -\mu_0 \vec{J}_\mu}$$

## Electromagnetic Field Tensor

$F_{\mu\nu}$  is an electromagnetic field tensor

$$F_{\mu\nu} = \left( \frac{\partial A_\nu}{\partial x_\mu} - \frac{\partial A_\mu}{\partial x_\nu} \right) \quad \text{--- (1)}$$

$$F_{\mu\nu} = \begin{bmatrix} F_{11} & F_{12} & F_{13} & F_{14} \\ F_{21} & F_{22} & F_{23} & F_{24} \\ F_{31} & F_{32} & F_{33} & F_{34} \\ F_{41} & F_{42} & F_{43} & F_{44} \end{bmatrix}$$

From definition (1)  $F_{\mu\mu} = 0$  i.e. all diagonal elements are 0.

From definition (1)  $F_{\mu\nu} = -F_{\nu\mu}$  i.e. anti-symmetric matrix

$$F_{11} = 0 = F_{22} = F_{33} = F_{44}$$

$$F_{12} = -F_{21} = \left( \frac{\partial A_2}{\partial x_1} - \frac{\partial A_1}{\partial x_2} \right)$$

$$F_{13} = -F_{31} = \left( \frac{\partial A_3}{\partial x_1} - \frac{\partial A_1}{\partial x_3} \right)$$

$$F_{14} = -F_{41} = \left( \frac{\partial A_4}{\partial x_1} - \frac{\partial A_1}{\partial x_4} \right)$$

$$F_{23} = -F_{32} = \left( \frac{\partial A_3}{\partial x_2} - \frac{\partial A_2}{\partial x_3} \right)$$

$$F_{24} = -F_{42} = \left( \frac{\partial A_4}{\partial x_2} - \frac{\partial A_2}{\partial x_4} \right)$$

$$F_{34} = -F_{43} = \left( \frac{\partial A_4}{\partial x_3} - \frac{\partial A_3}{\partial x_4} \right)$$

We thus need to find 6 components to find out whole matrix

$$\left[ \frac{\left( \frac{\partial A_2}{\partial x_1} \right)}{\left( \frac{\partial A_1}{\partial x_2} \right)} - \frac{\left( \frac{\partial A_1}{\partial x_2} \right)}{\left( \frac{\partial A_2}{\partial x_1} \right)} \right]$$



$$\rightarrow \boxed{\vec{B} = \vec{\nabla} \times \vec{A}} = \begin{vmatrix} i & j & k \\ \frac{\partial}{\partial x_1} & \frac{\partial}{\partial x_2} & \frac{\partial}{\partial x_3} \\ A_1 & A_2 & A_3 \end{vmatrix}$$

$$B_1 = \frac{\partial A_3}{\partial x_2} - \frac{\partial A_2}{\partial x_3} = F_{23} \quad \text{--- (1)}$$

$$B_2 = \frac{\partial A_1}{\partial x_3} - \frac{\partial A_3}{\partial x_1} = F_{31} \quad \text{--- (2)}$$

$$B_3 = \frac{\partial A_2}{\partial x_1} - \frac{\partial A_1}{\partial x_2} = F_{12} \quad \text{--- (3)}$$

$$\rightarrow \vec{E} = -\vec{\nabla}\phi - \frac{\partial \vec{A}}{\partial t}$$

$$-\frac{i\vec{E}}{c} = \vec{\nabla}\left(\frac{i\phi}{c}\right) + \frac{i}{c}\frac{\partial \vec{A}}{\partial t}$$

$$\boxed{-\frac{i\vec{E}}{c} = \vec{\nabla}A_4 + \frac{\partial \vec{A}}{\partial x_4}}$$

$$-\frac{iE_1}{c} = \frac{\partial A_4}{\partial x_1} - \frac{\partial A_1}{\partial x_4} = F_{14} \quad \text{--- (4)}$$

$$-\frac{iE_2}{c} = \frac{\partial A_4}{\partial x_2} - \frac{\partial A_2}{\partial x_4} = F_{24} \quad \text{--- (5)}$$

$$-\frac{iE_3}{c} = \frac{\partial A_4}{\partial x_3} - \frac{\partial A_3}{\partial x_4} = F_{34} \quad \text{--- (6)}$$

$$\begin{aligned} \text{eg. } \underline{B_1}' &= F_{23}' \\ &= \sum_{p,q} a_{2p} a_{3q} F_{pq} \\ &= a_{22} a_{33} F_{23} \\ &= \underline{B_1} \end{aligned}$$

★ New transformation is defined as

$$F'_{xy} = \sum_{p,q} a_{xp} a_{yq} F_{pq}$$

where  $a_{p,q}$  are elements of Minowski Matrix

Replacing values in tensor,

$$F_{\mu\nu} = \begin{bmatrix} 0 & B_3 & -B_2 & (-i\frac{E_1}{c}) \\ -B_3 & 0 & B_1 & (-i\frac{E_2}{c}) \\ B_2 & -B_1 & 0 & (-i\frac{E_3}{c}) \\ (i\frac{E_1}{c}) & (i\frac{E_2}{c}) & (i\frac{E_3}{c}) & 0 \end{bmatrix}$$

◦ EM Field tensor is an antisymmetric tensor of rank 2.

$$F_{\mu\nu}^2 = 2 \left[ B^2 - \frac{E^2}{c^2} \right]$$

◦ Note that using this tensor, field parameters can be written in  $S'$ .

### Boundary Conditions

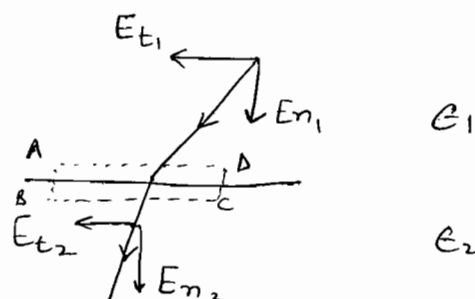
$$\begin{aligned} \circ \vec{D} &= \epsilon \vec{E} = k\epsilon_0 \vec{E} \\ &= \epsilon_0 \vec{E} + \vec{P} \end{aligned}$$

$$\begin{aligned} \circ \vec{B} &= \mu \vec{H} = \mu_0 \mu_r \vec{H} \\ &= \mu_0 [\vec{H} + \vec{M}] \end{aligned}$$

$$\oint \vec{E} \cdot d\vec{l} = 0$$

$$\circ \oint \vec{D} \cdot d\vec{s} = q_{\text{free}}$$

AB: very small



$$\oint_{ABCD} \vec{E} \cdot d\vec{l} = -E_{t_2} (BC) + E_{t_1} (AD) = 0$$

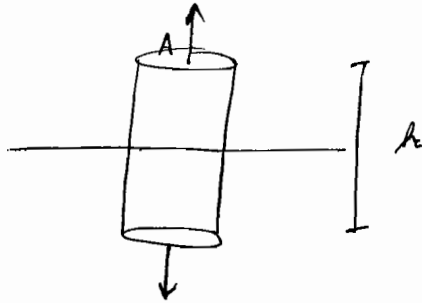
$$\Rightarrow \boxed{E_{t_1} = E_{t_2}}$$

Tangential Component of  $\vec{E}$  is continuous.

Similarly using the same,

$$\oint \vec{H} \cdot d\vec{l} = I$$

If there is no current on Boundary,  $\boxed{H_{t_1} = H_{t_2}}$



Making  $h \rightarrow 0$

$$\oint \vec{B} \cdot d\vec{s} = 0 = B_{n_1} A - B_{n_2} A = 0$$

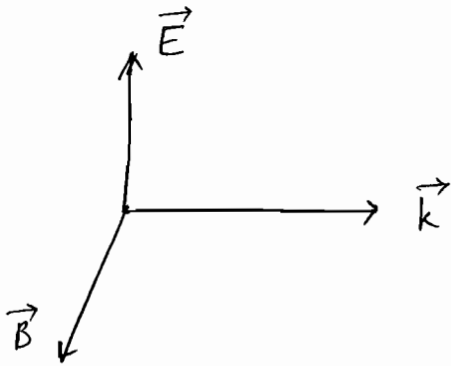
$$\Rightarrow \boxed{B_{n_1} = B_{n_2}} \quad \text{Normal Component of } \vec{B} \text{ is continuous}$$

$$\oint \vec{D} \cdot d\vec{s} = Q_{\text{free}}$$

If there is no charge on boundary,  $\boxed{D_{n_1} = D_{n_2}}$

# E&M (13)

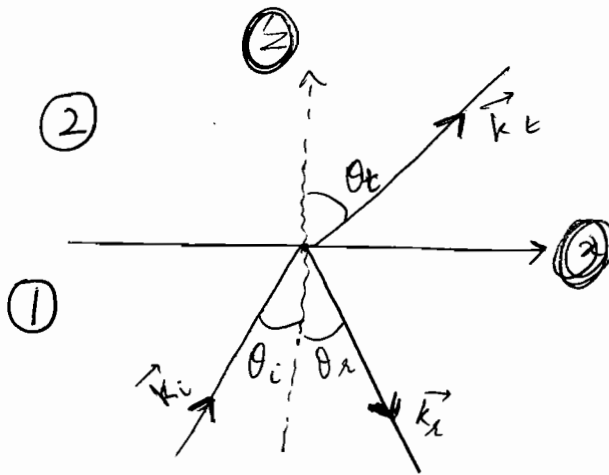
## EM Waves: Reflection & Refraction



At the interface between 2 dielectric media, we saw  $E_T$  and  $H_T$  are continuous.

We know,  $\frac{E}{H} = \frac{Z_0}{n} \Rightarrow H = \frac{n}{Z_0} E$

Consider interface between 2 media,



Boundary: xy plane

Plane of incidence: xz

Consider an EM wave incident at angle  $\theta_i$  from normal. We want to find out reflectivity of surface i.e. fraction of energy reflected

$$\text{Reflectivity} = \left( \frac{\text{Reflected Intensity}}{\text{Incident Intensity}} \right) = \left( \frac{S_r}{S_i} \right)$$

$$\text{Transmittivity} = \left( \frac{S_t}{S_i} \right)$$

where

$$S_i = E_i^2 / Z_1$$

$$S_r = E_r^2 / Z_1$$

$$S_t = E_t^2 / Z_2$$

$$R = \text{Reflectivity} = \left( \frac{E_r}{E_i} \right)^2$$

$\vec{S}$  is proportional to  $\mu$   
 [From  $\frac{1}{2} \epsilon c E^2$ , we have  $\frac{1}{2} \epsilon_0 n^2 \frac{c}{n} E_0^2$ , also]

$$T = \text{Transmittivity} = \left( \frac{Z_1}{Z_2} \right) \left( \frac{E_t}{E_i} \right)^2 = \frac{n_2}{n_1} \left( \frac{E_t}{E_i} \right)^2$$

$\therefore \mu$  is usually  
 the difference  
 is mainly of  $\epsilon_0$

$r = \left( \frac{E_r}{E_i} \right)$  &  $t = \left( \frac{E_t}{E_i} \right)$  are given by Fresnel's relations

We need to prove  $R+T=1$  : Conservation of Energy  
 (if assumed that no energy is absorbed)

$$R + T = \frac{E_r^2}{E_i^2} + \frac{Z_1}{Z_2} \left( \frac{E_t}{E_i} \right)^2 = 1$$

energy per unit time  
 reaching a particular  
 patch of area on  
 the surface =  
 energy per unit time  
 leaving the patch.

From the diagram on previous page,

$$\vec{k}_i = k_i \sin \theta_i \hat{x} + k_i \cos \theta_i \hat{z}$$

$$\vec{k}_r = k_r \sin \theta_r \hat{x} - k_r \cos \theta_r \hat{z}$$

$$\vec{k}_t = k_t \sin \theta_t \hat{x} + k_t \cos \theta_t \hat{z}$$

$$\vec{E}_i = \vec{E}_{i_0} e^{i(\vec{k}_i \cdot \vec{x} - \omega t)}$$

$$\vec{E}_r = \vec{E}_{r_0} e^{i(\vec{k}_r \cdot \vec{x} - \omega t)}$$

$$\vec{E}_t = \vec{E}_{t_0} e^{i(\vec{k}_t \cdot \vec{x} - \omega t)}$$

We know the boundary conditions,

$$E_{t_1} = E_{t_2}$$

$$\Rightarrow (\vec{E}_I)_t + (\vec{E}_R)_t = (\vec{E}_T)_t \quad \text{at } z=0 \quad \forall (x, y, t)$$

$$\Rightarrow (\vec{E}_{I_0})_t e^{i(\vec{k}_i \cdot \vec{x} - \omega t)} + (\vec{E}_{R_0})_t e^{i(\vec{k}_r \cdot \vec{x} - \omega t)} = (\vec{E}_{T_0})_t e^{i(\vec{k}_t \cdot \vec{x} - \omega t)}$$

We can put  $\vec{r} = 0$

If the equality is valid for all times, this can only happen when  $\omega_i = \omega_r = \omega_t$  — (1)

Hence EM waves frequency does not change upon reflection or refraction. velocity & wavelength may change but frequency is const.

Now we can put  $t = 0$

If the equality is valid for all  $\vec{r}$ , this can only happen when (along the interface)

$$\vec{k}_i \cdot \vec{r} = \vec{k}_r \cdot \vec{r} = \vec{k}_t \cdot \vec{r} \quad \text{--- (2)}$$

i.e. all phase parts are same

This also implies that all the 3 waves lie in the same plane

$$\Rightarrow (\vec{E}_i)_t + (\vec{E}_r)_t = (\vec{E}_t)_t$$

We have  $\vec{r} = (x\hat{x} + y\hat{y} + z\hat{z})$

### Laws of Reflection

$$\vec{k}_i \cdot \vec{r} = \vec{k}_r \cdot \vec{r} \quad \text{and} \quad \omega_i = \omega_r$$

$$\Rightarrow k_i x \sin \theta_i + k_i z \cos \theta_i = k_r x \sin \theta_r - k_r z \cos \theta_r$$

$$\text{@ } z=0 \text{ (boundary)} \quad \checkmark$$

$$k_i \sin \theta_i = k_r \sin \theta_r$$

$$v = \frac{\omega}{k} \Rightarrow k_i = \omega_i \sqrt{\mu_i \epsilon_i} = \omega_r \sqrt{\mu_r \epsilon_r} = k_r$$

$$\Rightarrow \sin \theta_i = \sin \theta_r$$

## Laws of refraction

$$\vec{k}_i \cdot \vec{z} = \vec{k}_t \cdot \vec{z}$$

$$\Rightarrow k_i \times \sin \theta_i = k_t \times \sin \theta_t \quad (\text{at } z=0)$$

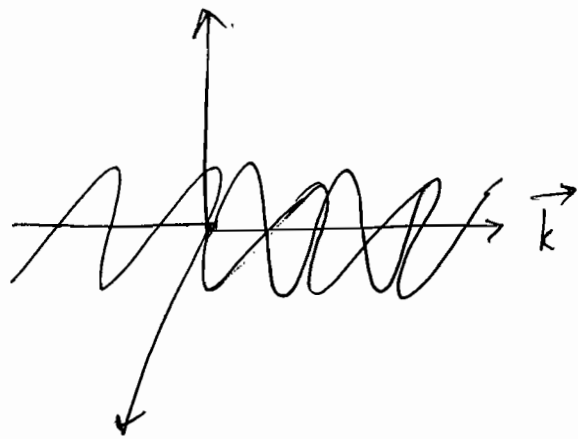
$$\Rightarrow k_i \sin \theta_i = k_t \sin \theta_t$$

$$k_i = \omega_i \sqrt{\mu_i \epsilon_i}$$

$$k_t = \omega_t \sqrt{\mu_t \epsilon_t}$$

$$\frac{\sin \theta_i}{\sin \theta_t} = \frac{k_t}{k_i} = \sqrt{\frac{\epsilon_t}{\epsilon_i}} = \sqrt{\frac{\mu_i \epsilon_t}{\mu_t \epsilon_i}} = \frac{n_t}{n_i}$$

$$\Rightarrow \boxed{n_i \sin \theta_i = n_t \sin \theta_t} \quad \text{Snell's law}$$



For a given  $\vec{k}$ ,  
Now  $\vec{E}$  can be polarized  
in 2 planes,

- (i) Parallel to plane of incidence
- (ii) Perpendicular to plane of incidence

Case 1  $E$  is Parallel to plane of incidence

This implies  $\vec{H}$  will ~~lie~~ <sup>tangential</sup> lie on boundary.

$$(\vec{H}_{i0})_t + (\vec{H}_{r0})_t = (\vec{H}_{t0})_t$$

$$\Rightarrow \boxed{H_{i0} + H_{r0} = H_{t0}} \quad [\text{all } \vec{H} \text{ are tangential only}]$$

$$\Rightarrow \frac{n_1 E_i}{Z_0} + \frac{n_1 E_r}{Z_0} = \frac{n_2 E_T}{Z_0}$$

\*  $E_r$  case में tangential components को ही equate कराटा है !!

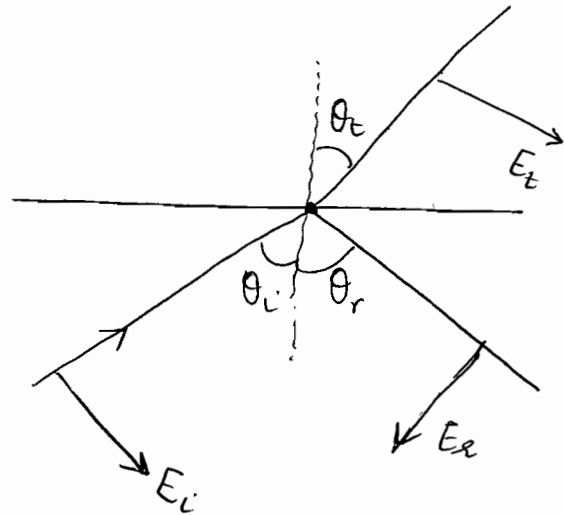
$$\Rightarrow \boxed{E_i + E_r = E_T \left( \frac{n_2}{n_1} \right)} \quad \text{--- (2)} \quad * \text{ Use } H = \left( \frac{n E}{Z_0} \right)$$

Also,

$$(\vec{E}_i)_t + (\vec{E}_r)_t = (\vec{E}_T)_t$$

$$\Rightarrow E_i \cos \theta_i - E_r \cos \theta_r = E_t \cos \theta_t$$

$$\Rightarrow \boxed{E_i - E_r = E_t \left( \frac{\cos \theta_t}{\cos \theta_i} \right)} \quad \text{--- (3)}$$



Adding (2) and (3) & Subtracting (2) and (3)

For normal incidence,  $\theta_i = \theta_r = 0$

$$E_i + E_r = E_T \left( \frac{n_2}{n_1} \right)$$

$$E_i - E_r = E_T$$

NOTE that we can take any direction of  $\vec{E}_r$  depending on  $n_1$  and  $n_2$  either  $E$  will be reversed or H.

$$\Rightarrow E_i = \left( \frac{n_1 + n_2}{2n_1} \right) E_T \quad \Rightarrow \boxed{\frac{E_T}{E_i} = \left( \frac{2n_1}{n_1 + n_2} \right)}$$

$$\Rightarrow E_r = \left( \frac{n_2 - n_1}{2n_1} \right) E_T \quad \Rightarrow \boxed{\frac{E_r}{E_i} = \left( \frac{n_2 - n_1}{n_2 + n_1} \right)}$$

$$R + T = \left( \frac{n_2 - n_1}{n_2 + n_1} \right)^2 + \frac{n_2}{n_1} \left( \frac{2n_1}{n_1 + n_2} \right)^2 = 1$$



For oblique incidence,

$$2E_i = \left( \frac{n_2}{n_1} + \frac{\cos \theta_t}{\cos \theta_i} \right) E_t$$

$$\frac{n_2}{n_1} = \frac{\sin \theta_i}{\sin \theta_t}$$

$$2E_r = \left( \frac{n_2}{n_1} - \frac{\cos \theta_t}{\cos \theta_i} \right) E_t$$

Fresnel equations

$$\left( \frac{E_t}{E_i} \right) = \frac{2}{\left( \frac{\sin \theta_i}{\sin \theta_t} + \frac{\cos \theta_t}{\cos \theta_i} \right)} = \left( \frac{2 \sin \theta_t \cos \theta_i}{\sin \theta_i \cos \theta_i + \sin \theta_t \cos \theta_t} \right) = \left[ \frac{4 \cos \theta_i \sin \theta_t}{\sin 2\theta_i + \sin 2\theta_t} \right]$$

$$\left( \frac{E_r}{E_i} \right) = \frac{\left( \frac{\sin \theta_i}{\sin \theta_t} - \frac{\cos \theta_t}{\cos \theta_i} \right)}{\left( \frac{\sin \theta_i}{\sin \theta_t} + \frac{\cos \theta_t}{\cos \theta_i} \right)} = \frac{\left[ \sin 2\theta_i - \sin 2\theta_t \right]}{\left[ \sin 2\theta_i + \sin 2\theta_t \right]} = \frac{\tan(\theta_i - \theta_t)}{\tan(\theta_i + \theta_t)} \times \left( \begin{array}{l} \text{ensure that} \\ \text{denominators} \\ \text{are written in} \\ \text{same form} \end{array} \right)$$

R + T ★ Note that factor of  $\cos \theta$  will be in addition, due to oblique incidence and Energy intensity

$$= \left( \frac{E_r}{E_i} \right)^2 + \left( \frac{E_t}{E_i} \right)^2 \left( \frac{\sin \theta_i}{\sin \theta_t} \cdot \frac{\cos \theta_t}{\cos \theta_i} \right) = S \cdot \hat{z} = S \cos \theta$$

$$= \left( \frac{E_r}{E_i} \right)^2 + \left( \frac{E_t}{E_i} \right)^2 \left( \frac{\sin \theta_i}{\sin \theta_t} \cdot \frac{\cos \theta_t}{\cos \theta_i} \right)$$

$$= \frac{(\sin 2\theta_i + \sin 2\theta_t)^2 + \sin^2 \theta_t \cos^2 \theta_i \frac{\sin \theta_i}{\sin \theta_t}}{[\sin(2\theta_i) + \sin(2\theta_t)]^2} \times \frac{\cos \theta_t}{\cos \theta_i}$$

$$= \underline{\underline{1}}$$

✓ At grazing incidence ( $\theta_i \approx 90^\circ$ ), wave is totally reflected. It creates problem for those who drive @ night on wet roads.

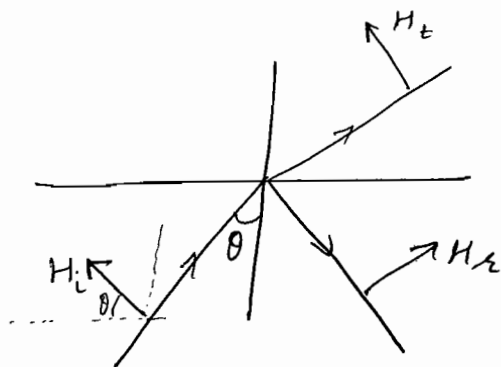
## Case 2

$E$  is perpendicular to plane of incidence.

$\Rightarrow E$  is lying on boundary

$$\Rightarrow \boxed{E_i + E_r = E_t} \quad \text{--- (1)}$$

$$-H_i \cos \theta_i + H_r \cos \theta_r = -H_t \cos \theta_t \quad \text{--- take } H_i \text{ in any direction}$$



Correspondingly we will have  $E_i$  and we arrive at some equation.

$$\Rightarrow \boxed{H_i - H_r = \frac{\cos \theta_t}{\cos \theta_r} H_t} \quad \text{--- (2)}$$

$$\Rightarrow \text{Using } \boxed{H = \frac{nE}{Z_0}}$$

$$\boxed{E_i - E_r = \frac{n_2}{n_1} \left( \frac{\cos \theta_t}{\cos \theta_i} \right) E_t} \quad \text{--- (3)}$$

Using (1) and (3),

$$\Rightarrow 2E_i = \left( 1 + \frac{\sin \theta_i \cos \theta_t}{\sin \theta_t \cos \theta_i} \right) E_t \quad \checkmark$$

$$2E_r = \left( 1 - \frac{\sin \theta_i \cos \theta_t}{\sin \theta_t \cos \theta_i} \right) E_t \quad \checkmark$$

$$\Rightarrow \left( \frac{E_r}{E_i} \right) = r = \left( \frac{\tan \theta_t - \tan \theta_i}{\tan \theta_t + \tan \theta_i} \right) \quad \checkmark$$

Same formula for normal incidence

$$\left(\frac{E_r}{E_i}\right) = \frac{-\sin(\theta_i - \theta_t)}{\sin(\theta_i + \theta_t)} \quad \checkmark$$

Fresnel equations

$$\left(\frac{E_t}{E_i}\right) = \frac{2 \sin \theta_t \cos \theta_i}{\sin(\theta_i + \theta_t)} \quad \checkmark$$

यहाँ 2A same funde:

$$r^2 + t^2 \cdot \left(\frac{\cos \theta_2}{\cos \theta_1}\right) \cdot \left(\frac{\sin \theta_1}{\sin \theta_2}\right) = 1$$

R

T

Brewster Angle

Consider Polarization Parallel to the plane of incidence

$$\left(\frac{E_r}{E_i}\right)_{\parallel} = \frac{\tan(\theta_i - \theta_t)}{\tan(\theta_i + \theta_t)} = \frac{1}{\infty} = 0$$

⊙ Now  $\theta_1 \neq \theta_2$

⇒ for  $\sin 2\theta_1 = \sin 2\theta_2$

$2\theta_1 = \pi - 2\theta_2$

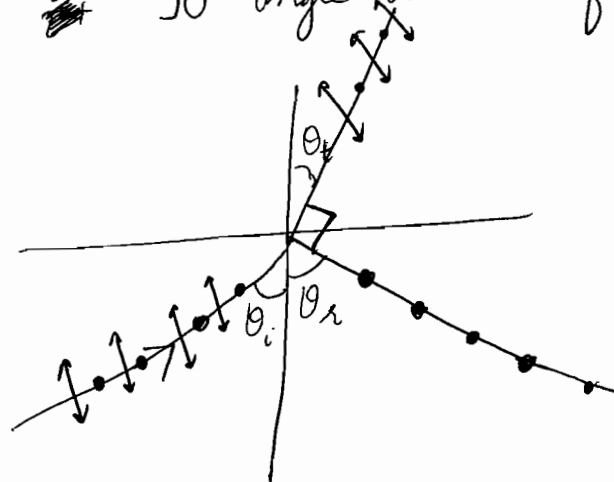
⇒  $\theta_1 + \theta_2 = \frac{\pi}{2}$

( $\vec{E}$  is  $\parallel$  to plane of incidence) if  $\theta_i + \theta_t = 90^\circ$  : Brewster Angle

If we can make  $(E_r)_{\parallel}$  zero, then we can say that light is polarized only in 1 direction. In this case,  $E_r$  is only in plane perpendicular to plane of incidence. So condition for reflected wave to have vibrations in plane  $\perp$  to plane of ~~motion~~ incidence:

$$\theta_i + \theta_t = (\pi/2)$$

⇒ ~~is~~  $90^\circ$  angle between reflected & transmitted wave.



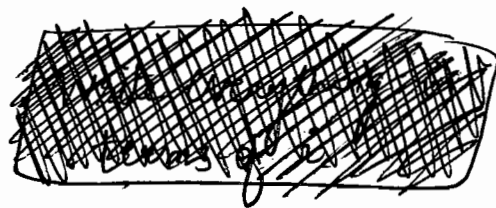
∴ The reflected wave is easily plane polarized as  $E_{\parallel}$  component is cut off by incidence at Brewster's angle.

# Total Internal Reflection

Consider Polarization to be parallel to the plane of incidence

$$\left(\frac{E_r}{E_i}\right)_{\parallel} = \frac{\left(\frac{n_2}{n_1} - \frac{\cos\theta_t}{\cos\theta_i}\right)}{\left(\frac{n_2}{n_1} + \frac{\cos\theta_t}{\cos\theta_i}\right)} = \frac{1 - \frac{n_1}{n_2} \frac{\cos\theta_t}{\cos\theta_i}}{1 + \frac{n_1}{n_2} \frac{\cos\theta_t}{\cos\theta_i}}$$

$$\left[ \begin{array}{l} \text{For critical angle } n_1 \sin\theta_i = n_2 \\ \Rightarrow \sin c = \left(\frac{n_2}{n_1}\right) \\ \& \sin\theta_t = \left(\frac{n_1}{n_2}\right) \sin\theta_i \end{array} \right]$$



$$= \frac{\cos\theta_i - \frac{n_1}{n_2} \cos\theta_t}{\cos\theta_i + \frac{n_1}{n_2} \cos\theta_t} = \frac{\cos\theta_i - \frac{n_1}{n_2} \sqrt{1 - \left(\frac{n_1}{n_2}\right)^2 \sin^2\theta_i}}{\cos\theta_i + \frac{n_1}{n_2} \sqrt{1 - \left(\frac{n_1}{n_2}\right)^2 \sin^2\theta_i}}$$

$$= \frac{\cos\theta_i - \left(\frac{n_1}{n_2}\right)^2 \sqrt{\left(\frac{n_2}{n_1}\right)^2 - \sin^2\theta_i}}{\cos\theta_i + \left(\frac{n_1}{n_2}\right)^2 \sqrt{\left(\frac{n_2}{n_1}\right)^2 - \sin^2\theta_i}}$$

$$= \frac{\cos\theta_i - \left(\frac{n_1}{n_2}\right)^2 \sqrt{\sin^2 c - \sin^2\theta_i}}{\cos\theta_i + \left(\frac{n_1}{n_2}\right)^2 \sqrt{\sin^2 c - \sin^2\theta_i}}$$

replace  $\theta_t$  by  $\theta_i$

$$\left[ \text{For } \underline{i > c} \quad \text{Put } \left(\frac{n_1}{n_2}\right)^2 \sqrt{\sin^2 i - \sin^2 c} = X \right]$$

$$= \frac{\cos\theta_i - jX}{\cos\theta_i + jX}$$

$$\left|\frac{E_r}{E_i}\right| = 1 \quad \Rightarrow \quad \underline{\text{Incident field is totally reflected.}}$$

# DISPERSION

## E&M (14)

✓ dispersion is the broadening of spectrum due to variation of  $n$  w.r.t. frequency of EM waves

✓ There are 2 types of dispersion

(1) Normal Dispersion :-  $\left(\frac{dv_p}{d\lambda}\right) > 0$  i.e.  $\left(\frac{dn}{d\lambda}\right) < 0$

$$\text{i.e. } \left(\frac{dn}{d\omega}\right) > 0$$

(2) Anomalous dispersion :-  $\left(\frac{dv_p}{d\lambda}\right) < 0$  i.e.  $\left(\frac{dn}{d\lambda}\right) > 0$

$$\text{i.e. } \left(\frac{dn}{d\omega}\right) < 0$$

Now in order to calculate  $\left(\frac{dn}{d\omega}\right)$ , we know need  $n$  as  $f(\omega)$

$$\vec{D} = \epsilon \vec{E} = \epsilon_r \epsilon_0 \vec{E} = \epsilon_0 \vec{E} + \vec{P}$$

$$\vec{P} = (\epsilon_r - 1) \epsilon_0 \vec{E} \Rightarrow \epsilon_r = 1 + \frac{P}{\epsilon_0 E}$$

We know,

$$\Rightarrow n^2 = 1 + \frac{P}{\epsilon_0 E}$$

Now to find  $n$ , we need  $\left(\frac{P}{E}\right)$ ,

We know,  $F = q(\vec{E} + \vec{v} \times \vec{B})$  : Force exerted on charge  $q$

This force causes displacement of dipoles which causes Polarization. Also  $|\vec{B}| = \frac{|\vec{E}|}{c} \Rightarrow |\vec{E}| \gg |\vec{v} \times \vec{B}|$

Lorentz Force

$\Rightarrow$  is neglected

$\Rightarrow \vec{F} = q\vec{E}$  : Periodic Oscillating Force  
hence forced oscillations.

Particles absorb energy from EM wave and hence accelerate, due to which they radiate out energy. This radiating out energy is called scattering.

$\vec{x}$ : displacement of charged particle due to EM wave

dipole Moment  $\vec{p} = e \vec{x}$

$$\vec{p} = \sum n_k \vec{p}_k$$

In a dielectric medium, restoring force acts upon the particle.

$$\Rightarrow m \ddot{x} = -kx - b\dot{x} + q E_0 e^{i(\vec{k} \cdot \vec{x} - \omega t)}$$

$$\Rightarrow \ddot{x} + \frac{k}{m} x + \frac{b}{m} \dot{x} = \left[ \frac{q E_0}{m} e^{i \vec{k} \cdot \vec{x}} \right] e^{-i\omega t}$$

Using the usual symbols,

$$\ddot{x} + 2c \dot{x} + \omega_0^2 x = f_0 e^{-i\omega t}$$

if  $D = \frac{d}{dt}$

$$\Rightarrow (D^2 + 2cD + \omega_0^2) x = f_0 e^{-i\omega t}$$

replacing D by  $-i\omega$

$$x = \frac{f_0 e^{-i\omega t}}{(\omega_0^2 - \omega^2) + (2c\omega) i}$$

$\omega$ : EM wave freq.  
 $\omega_0$ : molecular parameter

$$x = \frac{f_0}{\sqrt{(\omega_0^2 - \omega^2)^2 + 4c^2 \omega^2}} e^{-i(\omega t - \phi)}$$

$$\phi = \tan^{-1} \left( \frac{2c\omega}{\omega_0^2 - \omega^2} \right)$$

exactly same (replace  $\phi$  by  $\omega$ )

$$\vec{p} = q\vec{x}$$

$$\vec{p} = \frac{q^2 \vec{E}}{m} \frac{e^{i(\vec{k} \cdot \vec{x} - \omega t)}}{(\omega_0^2 - \omega^2) - 2c\omega i}$$

✓ \* But natural frequency  $\omega_0$  is not the same for all vibrating molecules.

Assume  $n$  dipoles per unit volume vibrating in groups and  $n_k$  dipoles having natural frequency

$\omega_k$

$$\sum_k n_k = n$$

$$\vec{p}_k = \frac{q^2 \vec{E}}{m[(\omega_k^2 - \omega^2) - 2c\omega i]}$$

$$\vec{p} = \sum_k n_k \vec{p}_k$$

$$= \sum_k \frac{n_k q^2 \vec{E}}{m[(\omega_k^2 - \omega^2) - 2c\omega i]}$$

$$\frac{p}{E} = \frac{q^2}{m} \sum \frac{n_k}{(\omega_k^2 - \omega^2) - 2c\omega i}$$

Lorentz Dispersion Eqn

$$n^2 = 1 + \frac{p}{\epsilon_0 E} \Rightarrow n^2 = 1 + \frac{e^2}{\epsilon_0 m} \sum \frac{n_k}{(\omega_k^2 - \omega^2) - 2c\omega i}$$

Hence  $n = n(\omega)$

$\omega_k$ : natural freq.  
 $\omega$ : EM freq

Case 1 Normal dispersion : damping is neglected

$$n^2 = 1 + \frac{e^2}{\epsilon_0 m} \sum \frac{n_k}{(\omega_k^2 - \omega^2)}$$

$\Rightarrow n$  becomes real

**Normal dispersion** is there is regions far from natural frequencies, where damping is very small

$$n^2 = 1 + \frac{e^2}{\epsilon_0 m} \sum \frac{n_k}{(\omega_k^2 - \omega^2)}$$

Now  $\lambda = \frac{2\pi c}{\omega}$ ,  $\lambda_k = \frac{2\pi c}{\omega_k}$

$$\Rightarrow n^2 = 1 + \frac{e^2}{m \epsilon_0} \sum \frac{n_k \lambda^2 \lambda_k^2}{(\lambda^2 - \lambda_k^2)(2\pi c)^2}$$

$$= 1 + \sum_k \frac{e^2 \lambda_k^2 n_k}{m \epsilon_0 (2\pi c)^2} \cdot \frac{\lambda^2}{\lambda^2 - \lambda_k^2}$$

$$n^2 = 1 + \sum_k \frac{A_k \lambda^2}{\lambda^2 - \lambda_k^2}$$

Drallmeier's Formula

where

$$A_k = \frac{n_k \cdot \lambda_k^2}{4\pi^2 c^2} \frac{e^2}{\epsilon_0 m}$$

Now for  $\lambda \gg \lambda_k$ , we have

$$n^2 = 1 + \sum_k A_k \cdot \frac{\lambda^2}{\lambda^2} \left[ 1 - \frac{\lambda_k^2}{\lambda^2} \right]^{-1}$$

$$= 1 + \sum_k A_k \left( 1 + \frac{\lambda_k^2}{\lambda^2} + \frac{\lambda_k^4}{\lambda^4} + \dots \right)$$

$$n^2 = A + \frac{B}{\lambda^2} + \frac{C}{\lambda^4} + \dots$$

Cauchy's Formula

### Anomalous Dispersion

If frequency of EM wave lie near a natural freq, we obtain phenomenon of anomalous dispersion.

For simplicity we assume that there is only one natural frequency  $\omega_k = \omega_0$

$$n \approx 1 + \frac{e^2}{2 m \epsilon_0} \left[ \frac{n}{\omega_0^2 - \omega^2 - i(2c\omega)} \right]$$

Now rationalize & put  $\omega \approx \omega_0$



$$\Rightarrow n^2 = 1 + \frac{e^2}{2m\epsilon_0} \sum_k \left( \frac{n_k}{\omega_k^2 - \omega^2} \right) \quad \text{Sellmire formula (binomial)}$$

Sellmire's Formula : Cauchy's Formula

$$\omega_k = \frac{2\pi c}{\lambda_k}$$

$$n = 1 + \frac{e^2}{2m\epsilon_0} \sum_k \frac{n_k}{(2\pi c)^2} \left[ \frac{1}{\lambda_k^2} - \frac{1}{\lambda^2} \right]^{-1}$$

$$= 1 + \frac{e^2}{(2m\epsilon_0)(2\pi c)^2} \sum_k \frac{n_k \cdot \lambda_k^2}{\lambda^2} \left[ 1 - \left( \frac{\lambda_k}{\lambda} \right)^2 \right]^{-1}$$

Note that  $\lambda_k < \lambda$  i.e.  $\omega_k > \omega$ . Under this damping is negligible

$$= 1 + \frac{e^2}{8\pi^2 c^2 m \epsilon_0} \sum_k n_k \lambda_k^2 \left[ 1 + \left( \frac{\lambda_k}{\lambda} \right)^2 + \left( \frac{\lambda_k}{\lambda} \right)^4 + \dots \right]$$

$$\Rightarrow n = 1 + A + \frac{B}{\lambda^2} + \frac{C}{\lambda^4}$$

$$\frac{dn}{d\omega} > 0$$

Cauchy's Formula

$$n = 1 + A \left( 1 + \frac{B}{\lambda^2} \right)$$

Coefficient of refraction

Coefficient of dispersion

Case 2 Anomalous Dispersion

$$n^2 = 1 + \frac{e^2}{m\epsilon_0} \sum \frac{n_k}{(\omega_k^2 - \omega^2) - 2c\omega i}$$

$\Rightarrow n$ : complex

Part  $n = n_+ + i n_-$

$$n^2 = 1 + \frac{e^2}{m\epsilon_0} \sum \frac{n_k ((\omega_k^2 - \omega^2) + i 2c\omega)}{(\omega_k^2 - \omega^2)^2 + 4c^2\omega^2}$$

(Using Binomial to remove factor of  $(\frac{1}{2})$ )

$$n = 1 + \frac{e^2}{2m\epsilon_0} \sum \frac{n_k (\omega_k^2 - \omega^2)}{(\omega_k^2 - \omega^2)^2 + 4c^2\omega^2} + i \frac{e^2}{2m\epsilon_0} \sum \frac{n_k}{(\omega_k^2 - \omega^2)^2 + 4c^2\omega^2}$$

For  $\omega = \omega_k$ ,  $n_+ = 1$  ✓

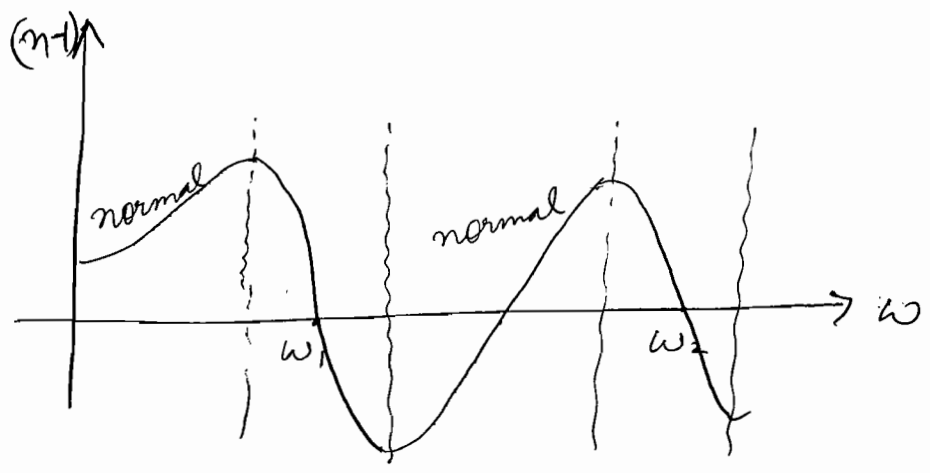
$$n_- = \frac{A}{\omega} \quad ; \quad A = \frac{2ne^2}{4c\omega m\epsilon_0} \quad \checkmark$$

for analysis see  $\omega \approx \omega_k$   
 $\omega \neq \omega_k$   
Not equal

$\Rightarrow \frac{dn}{d\omega} < 0$  : Anomalous dispersion

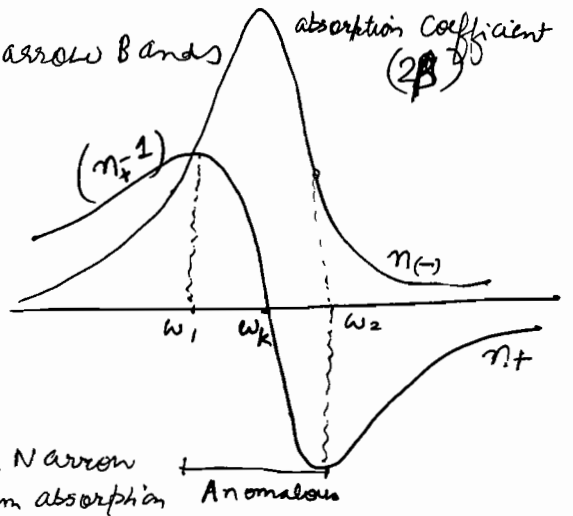
When vibration of particles ~~is~~ is matched by the EM frequency, maximum absorption takes place. (See the graph below)  
i.e.  $\frac{dn}{d\omega} < 0$  : only in neighbourhood of  $\omega_k$ .

Incident frequency in neighbourhood of absorption frequency  $\Rightarrow$  narrow band



imaginary part leads to absorption or attenuation  
Its maximum at  $\omega_k$

anomalous dispersion : Narrow Bands absorption coefficient  $(\beta)$



$\omega_1$ : Absorption Frequency

Most of the time, (refraction) <sup>index of</sup> rises gradually with increasing frequency, consistent with normal dispersion. However in immediate neighbourhood of resonance, index of refraction drops sharply, its called Anomalous Dispersion. Narrow Band of Anomalous Dispersion coincides with maximum absorption

# Scattering

Medium particles act like oscillating dipoles when EM waves are incident.

$$\vec{F} = e E_0 \sin \omega t$$

$$\vec{x} = \frac{e E_0 \sin(\omega t - \theta)}{m \sqrt{(\omega_k^2 - \omega)^2 + 4c^2 \omega^2}}$$

$$\vec{p} = e \vec{x} = \frac{e^2 E_0 \sin(\omega t - \theta)}{m \sqrt{(\omega_k^2 - \omega)^2 + 4c^2 \omega^2}}$$

Since its an oscillating dipole,

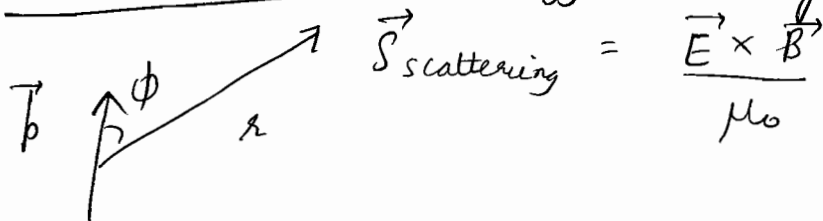
$$p = p_0 \sin(\omega t - \theta)$$

$$\text{where } p_0 = \frac{e^2 E_0}{m \sqrt{(\omega_k^2 - \omega)^2 + 4c^2 \omega^2}}$$

Dipole's Natural frequency:  $\omega_k$

Incident frequency:  $\omega$

At  $\omega_k \approx \omega$ : dipole absorbs maximum energy which is radiates out: energy scattering


$$\vec{S}_{\text{scattering}} = \frac{\vec{E} \times \vec{B}}{\mu_0}$$

Scattering Energy at  $(\theta, \phi)$ : Poynting Vector

Scattering  $\left\{ \begin{array}{l} \text{Rayleigh} \\ \text{Thomson} \end{array} \right.$  Scattering of Bound Charged Particles  
Restoring force acts.  $e^-$  do not leave atom.

Scattering of free electrons

Nuclei treated at rest. only  $e^-$  are displaced.

$$I_{\text{scattering}} \propto \frac{P_0^2}{r^2} (1 + \cos^2 \phi)$$

$$I_{\text{scattering}} \propto \frac{\omega^4}{r^2} (1 + \cos^2 \phi)$$

$$I_{\text{scattering}} \propto \frac{1}{\lambda^4} \quad (\text{Rayleigh Scattering})$$

★ In case of perfect conductors, the incident wave is totally reflected with a  $180^\circ$  phase shift. That's why excellent conductors make good mirrors. In practice, you paint a thin coating of silver onto the back of a pane of glass - the glass has nothing to do with the reflection; it's just there to support the silver and to keep it from tarnishing. Since the skin depth of silver at optical frequencies  $\approx 100 \text{ \AA}$ , we don't need a very thick layer.

★ Since attenuated field falls off as  $e^{-\beta x}$   
 attenuated intensity falls off as  $e^{-2\beta x}$   
 " $2\beta$ " is called Absorption Coefficient.

Note that  $k = \left(\frac{\omega}{c}\right) n$

$$\Rightarrow n(\omega) = \frac{k(\omega)}{\left(\frac{c}{\omega}\right)}$$

$$= \frac{\beta(\omega)}{\left(\frac{c}{\omega}\right)}$$

$$n(\omega) = \frac{k(\omega)}{\left(\frac{c}{\omega}\right)}$$

$$= \alpha \left(\frac{c}{\omega}\right)$$

Induction

2 types of EM induction:

✓ 1) Self Induction : effect in same coil✓ 2) Mutual Induction : effect on other coil

For self Induction:

$$\phi \propto I$$

$$N\phi \propto I$$

$$N\phi = LI$$

✓ 
$$L = \frac{N\phi}{I}$$

$$-N \frac{d\phi}{dt} = -L \left( \frac{dI}{dt} \right)$$

✓ 
$$\text{Emf } e = -L \left( \frac{dI}{dt} \right)$$

2 definitions  
of  
induction L

$$\text{Work Done by the circuit} = e i dt = -L \left( \frac{di}{dt} \right) i dt$$

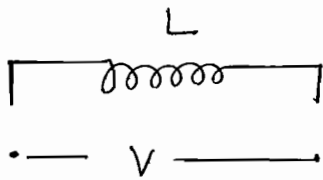
$$W = \int dw = \int -L \left( \frac{di}{dt} \right) i dt = \int -L i di = -\frac{L I^2}{2}$$

since Work Done is  $\ominus$ ve,  $\Rightarrow$  Work is done upon system $\Rightarrow$  It is stored as energy.

$$\Rightarrow \text{Energy stored} = \frac{1}{2} L I^2$$

○ In D.C. :  $\frac{dI}{dt} = 0 \Rightarrow$  No role of inductor in D.C. circuit.But when circuit is switched on or switched off, transient currents flow; here inductor plays an important role.

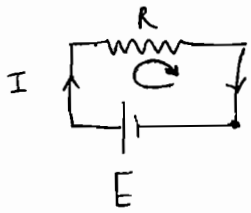
Signs



drop in voltage  $V = L \left( \frac{dI}{dt} \right)$



: Battery or Accumulator  
It has some emf \$E\$.



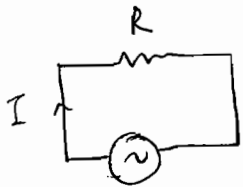
$$E - IR = 0 \Rightarrow I = \left( \frac{E}{R} \right)$$



: A.C. source

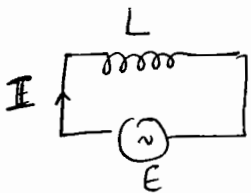
$$E = E_0 \sin \omega t = E_0 e^{j\omega t}$$

( $j$ : used since 'i' is used for current)



$$I = \left( \frac{E}{R} \right) = \left( \frac{E_0}{R} \right) e^{j\omega t} = I_0 e^{j\omega t}$$

(Note that phase is same)



$$L \left( \frac{dI}{dt} \right) = E$$

$$I = \frac{E_0}{L} \int e^{j\omega t} dt = \frac{E_0}{L} \frac{e^{j\omega t}}{j\omega}$$

$$= \frac{E_0}{j\omega L} e^{j\omega t}$$

$$\boxed{Z = X_L = j\omega L}$$

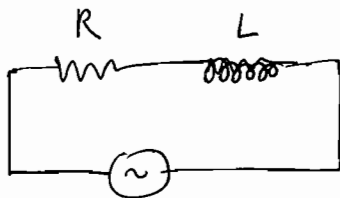
Inductive Impedance  
or  
Reactance

We know  $j = e^{j\pi/2}$

$$\Rightarrow I = \frac{E_0}{\omega L} \frac{e^{j\omega t}}{e^{j\pi/2}} = \frac{E_0}{\omega L} e^{j(\omega t - \pi/2)}$$

✓ In Purely inductive circuit, current LAGS by  $(\pi/2)$  w.r.t. EMF.

Inductor is equivalent to resistance of value  $j\omega L$



aim is to find I, Power.

Big word is  $Z$ : impedance

It can be  $R$ : resistance

$j\omega L$ : inductive reactance

$\frac{1}{j\omega C}$ : capacitive reactance

In the above RL circuit

$$Z = R + j\omega L = \sqrt{R^2 + \omega^2 L^2} \angle \phi = \sqrt{R^2 + \omega^2 L^2} e^{j \tan^{-1}(\frac{\omega L}{R})}$$

$$\text{We know } z = x + iy = r e^{i \tan^{-1}(y/x)} = r \angle \phi$$

$$E_R = -IR$$

$$E_L = -L \left( \frac{dI}{dt} \right) = j\omega L I$$

$$\text{We know, } E_R + E_L = E$$

$$\Rightarrow E = IR + j\omega L I$$

$$\Rightarrow I = \left( \frac{E}{R + j\omega L} \right)$$

$$\underline{IR + L \left( \frac{dI}{dt} \right) = E_0 \sin \omega t} \quad \text{--- AC}$$

$$\underline{IR + L \left( \frac{dI}{dt} \right) = E_0} \quad \text{--- DC}$$

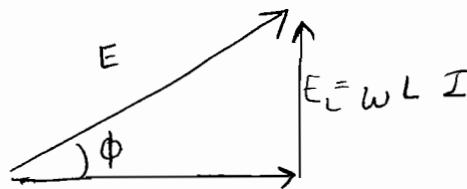
For A.C.

$$I = \frac{E_0}{\sqrt{R^2 + \omega^2 L^2}} \frac{e^{j\omega t}}{e^{j\phi}} = \frac{E_0 e^{j(\omega t - \phi)}}{\sqrt{R^2 + \omega^2 L^2}}$$

$\Rightarrow$  Current lags by angle (or phase difference) =  $\phi$

$$= \tan^{-1} \left( \frac{\omega L}{R} \right)$$

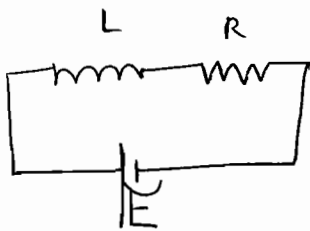
do not forget  
 $\tan^{-1}$



$$E_R = IR$$

For Switching on

For D.C. (Transient Response)



$$E_R + E_L = E$$

$$IR + L \left( \frac{dI}{dt} \right) = E$$

$$L \frac{dI}{dt} = E - IR$$

$$\frac{dI}{E - IR} = \frac{dt}{L}$$

$$\frac{\ln(E - IR)}{-R} = \frac{t}{L} + C_1$$

$$\Rightarrow \ln(E - IR) = -\frac{Rt}{L} - RC_1$$

$$\text{At } t = 0, I = 0 \Rightarrow \ln(E) = -RC_1 \Rightarrow C_1 = -\frac{\ln E}{R}$$



$$\Rightarrow \ln(E - IR) - \ln E = -\frac{R}{L}t$$

$$\Rightarrow e^{-\frac{Rt}{L}} = \frac{E - IR}{E} = 1 - \frac{IR}{E}$$

$$\Rightarrow \boxed{I = \frac{E}{R} (1 - e^{-Rt/L})}$$

at  $t = \infty$ ,  $I = \left(\frac{E}{R}\right)$

Power across  $Z = I_Z V_Z$  (We know)

$\Rightarrow$  Power<sub>R</sub> =  $I^2 R$  : Instantaneous Power across R

& Power<sub>L</sub> =  $I L \left(\frac{dI}{dt}\right)$  : Instantaneous Power across L

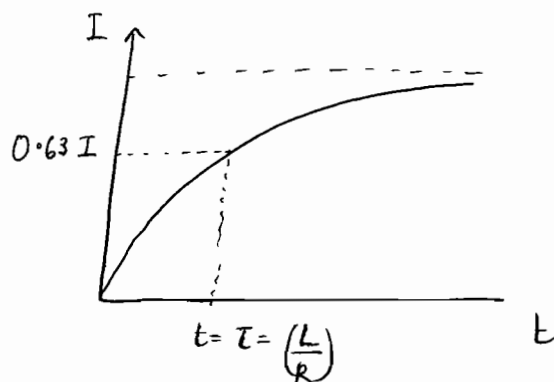
$$I = I_0 (1 - e^{-Rt/L})$$

at  $t = \left(\frac{L}{R}\right)$   $I = I_0 (1 - e^{-1})$

$$\frac{1}{e} = 0.37$$

$$1 - \frac{1}{e} = 0.63$$

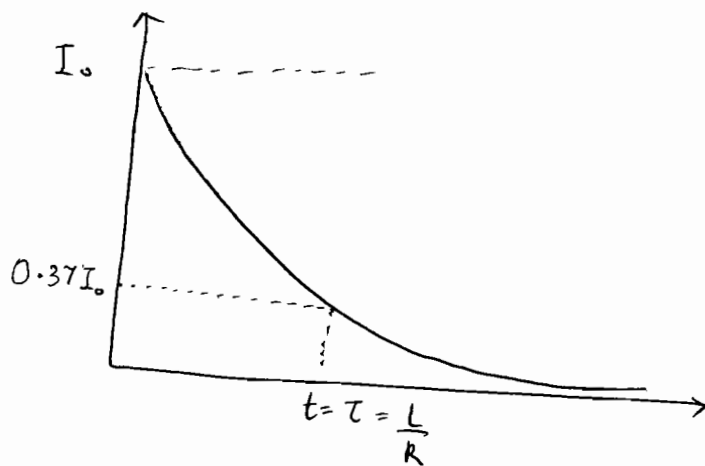
$\Rightarrow$  at  $t = \tau = \frac{L}{R}$ ,  $I = 0.63 I_0$



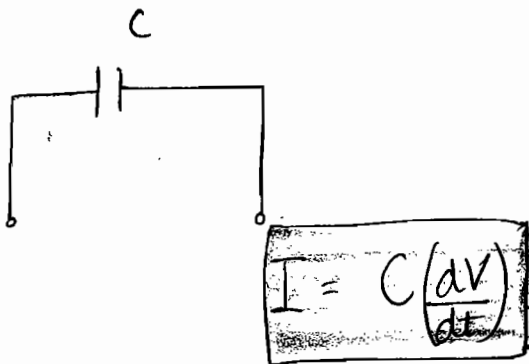
For switching off

$$IR + L \frac{dI}{dt} = 0$$

$$I = I_0 e^{-\frac{Rt}{L}}$$



Capacitance



$$Q = CE$$

$$I = \frac{dQ}{dt} = C \left( \frac{dE}{dt} \right)$$

For A.C.  $I = CE \cdot j\omega e^{j\omega t} = \frac{E_0}{\left( \frac{1}{j\omega C} \right)} e^{j\omega t}$

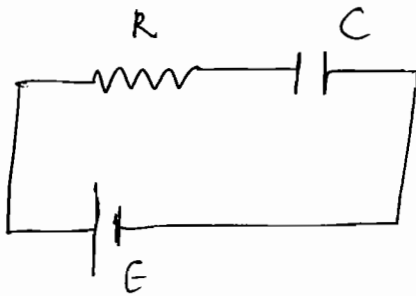
$$Z = X_C = \frac{1}{j\omega C}$$

For D.C.,  $Z$  is infinite, hence totally block D.C. except during switching on & switching off.

$$I = \frac{E_0}{\left(\frac{1}{\omega C}\right)} e^{\pi/2} e^{j\omega t} = I_0 e^{j(\omega t + \pi/2)}$$

⇒ For purely capacitive circuit, current leads by phase  $(\pi/2)$  wrt. EMF.

Switch On



$$IR + \frac{Q}{C} = E$$

$$R \frac{dQ}{dt} + \frac{Q}{C} = E$$

$$\Rightarrow \frac{dQ}{E - \frac{Q}{C}} = \frac{dt}{R}$$

$$\Rightarrow \ln\left(E - \frac{Q}{C}\right) \cdot (-C) = \frac{t}{R} + C_1 \rightarrow \text{do not take derivative and}$$

$$\Rightarrow \ln\left(E - \frac{Q}{C}\right) = -\frac{t}{RC} - \frac{C_1}{C} \text{ rather calculate } Q.$$

$$\text{at } t=0, Q=0$$

$$\Rightarrow \ln E = -\frac{C_1}{C}$$

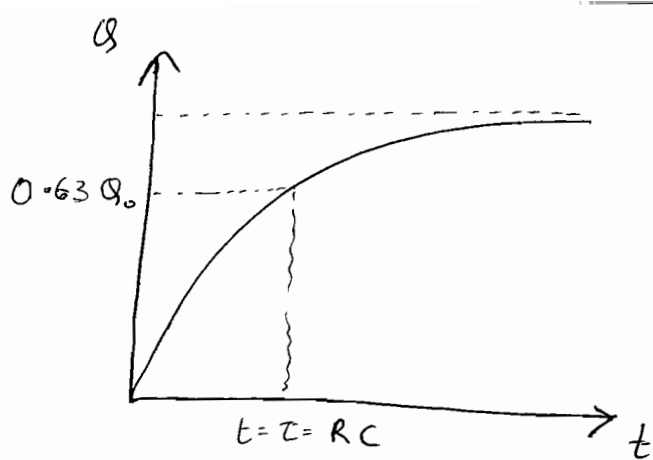
$$\Rightarrow \ln\left(1 - \frac{Q}{EC}\right) = -\frac{t}{RC}$$

$$\Rightarrow \boxed{Q = EC \left(1 - e^{-t/RC}\right)}$$

$$Q = Q_{\max} = EC \text{ at } t = \infty$$

$$I = \left(\frac{dQ}{dt}\right)$$

calculate  $I$  in the end only



Similarly for switch off

$$Q = C V_0 e^{-(t/RC)}$$

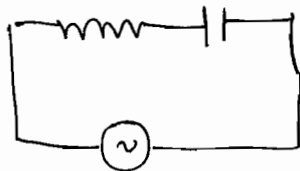
$$I_c = -\frac{V_0}{R} e^{-(t/RC)}$$

At  $t = \tau = RC$

$$\text{Power}_R = I^2 R$$

$$\text{Power}_{cap} = \frac{Q}{C} \left( \frac{dQ}{dt} \right)$$

R-Circuit AC



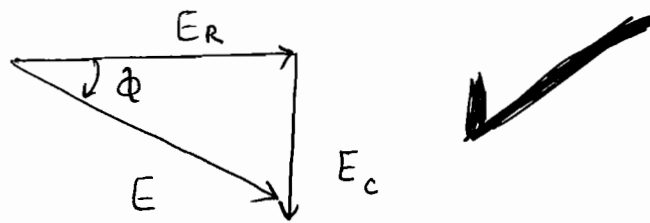
$$Z = R + \frac{1}{j\omega C} = R - \frac{j}{\omega C}$$

$$= \sqrt{R^2 + \frac{1}{\omega^2 C^2}} e^{-\tan^{-1}\left(\frac{1}{\omega RC}\right)}$$

$$I = \frac{E_0}{\sqrt{R^2 + \frac{1}{\omega^2 C^2}}} e^{j\left(\omega t + \tan^{-1}\left(\frac{1}{\omega RC}\right)\right)}$$

⇒ Current is leading by phase  $\tan^{-1}\left(\frac{1}{\omega RC}\right)$

$$E_c = I X_c = \frac{I}{j\omega C} = \frac{I}{\omega C} e^{-j\pi/2}$$



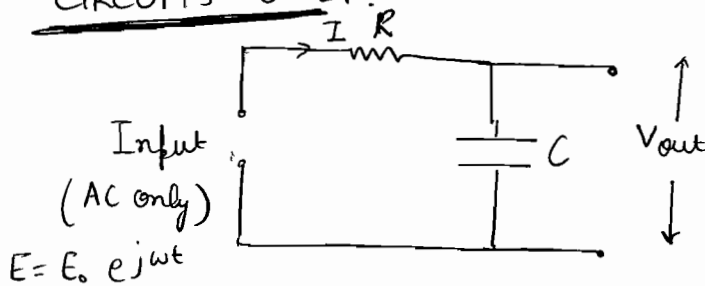
$$\phi = \tan^{-1}\left(\frac{1}{\omega RC}\right)$$

ICE  
 ↑  
 Capacitor  
 Current  $\underline{I_c}$   
 leads Voltage  
 $\underline{I_c} \underline{E_c}$

$$\text{Energy stored in Capacitor} = \frac{1}{2} \frac{Q^2}{C} = \frac{1}{2} CV^2$$

RL and RC circuits can be used as integrator & differentiator circuits under different conditions in AC

CIRCUITS ONLY.



INTEGRATOR

$$E_R = \frac{IR}{R} \quad E_C = \frac{Q}{C} = \frac{1}{C} \int i dt = \frac{1}{RC} \int E_R dt$$

If  $\underline{V_R} = V_i$   $\Rightarrow$   $E_{out} = E_C = \frac{1}{RC} \int V_{In} dt$

Integrator circuit output  $\propto \int (\text{Input})$

For differentiator circuit output  $\propto \frac{d(\text{input})}{dt}$

Our aim is to make  $V_R \approx V_i$

We know  $V_i = V_R + V_C$  if  $V_C \ll V_R \Rightarrow V_i \approx V_R$

$$V_c = \frac{I}{\omega C} \ll IR$$

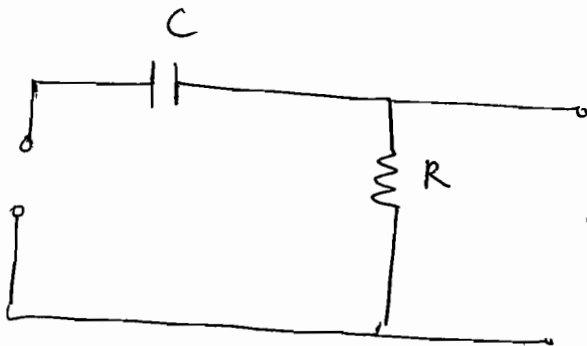
$$\Rightarrow \boxed{R \gg \frac{1}{\omega C}}$$

$$\text{i.e. } \boxed{\omega \gg \frac{1}{RC}}$$

R and C are constants

hence if  $\omega$  of ~~source~~ <sup>chosen</sup> in such a way that  $\omega \gg \frac{1}{RC}$

$\Rightarrow$  the output is integrated input.



DIFFERENTIATOR

$$\begin{aligned} V_o &= IR \\ &= C \left( \frac{dV_i}{dt} \right) R \\ &= RC \left( \frac{dV_i}{dt} \right) \end{aligned}$$

$$I = \left( \frac{dQ}{dt} \right), \quad E_c = \frac{Q}{C}$$

$$V_o = RI = R \left( \frac{dQ}{dt} \right) = RC \left( \frac{dE_c}{dt} \right) = \frac{RC}{RC} \frac{dV_c}{dt}$$

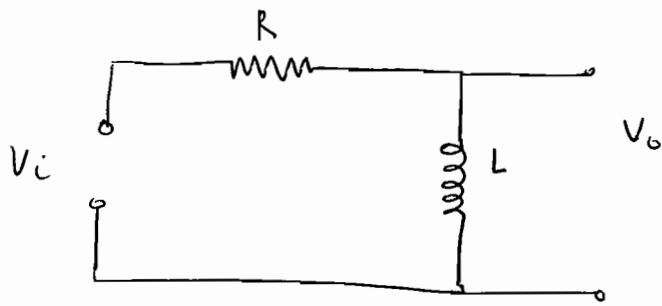
$$= RC \frac{dV_i}{dt} \quad \text{if } \underline{V_i \approx V_c} \Rightarrow \text{differentiator}$$

$$V_i \approx V_c \text{ if } \frac{1}{\omega C} \gg R$$

$$\Rightarrow \boxed{\omega \ll \frac{1}{RC}} \quad \checkmark$$

if output across R, and  $\omega$  chosen st.  $\omega \ll \frac{1}{RC}$ ,

circuit acts as differentiator.



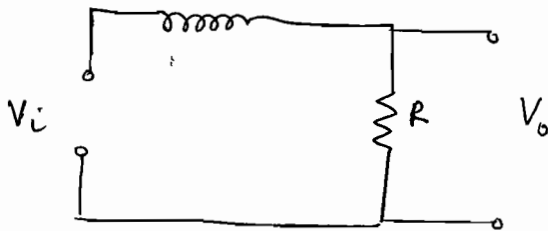
## DIFFERENTIATOR

$$V_o = V_L = L \frac{dI}{dt} = L \frac{d}{dt} \frac{V_R}{R} = \frac{L}{R} \frac{d}{dt} V_R = \frac{L}{R} \frac{dV_i}{dt}$$

if  $V_R \approx V_i$ ; if  $R \gg \omega L$

$\Rightarrow$  differentiator if  $\omega \ll \left(\frac{R}{L}\right)$

In all the 4 cases, we have to have a large ~~impedance~~ impedance of the component placed at this position.



## INTEGRATOR

$$V_o = IR = \frac{R}{L} \int V_L dt$$

if  $V_L \approx V_{in} \Rightarrow$  integrator  
 $\Rightarrow \omega L \gg R \Rightarrow$

$$\omega \gg \frac{R}{L}$$

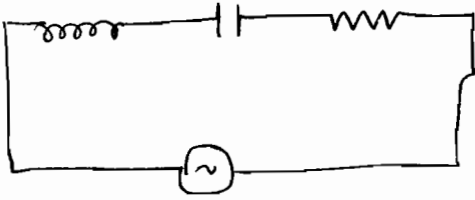
$\rightarrow$  RL circuit is choke coil, used for regulation of AC.

Its a coil of large inductance  $L$  and low resistance  $R$ .  
 Since Power factor  $\cos \phi = \frac{R}{\sqrt{R^2 + L^2 \omega^2}} = \frac{R}{\omega L} \Rightarrow$  power absorbed by coil is very less.

Its used in AC currents for the purpose of adjusting current to any required value w/o wastage of energy.

# LCR ckt

# EM (16)



@ resonance, impedance is minimum & ckt. is called Acceptor circuit

Acceptor  
Circuit

or  
Series LCR Circuit

$$E = V_L + V_R + V_C$$

$$I = \frac{E}{Z} = \frac{E}{R + j\omega L + \frac{1}{j\omega C}} = \frac{E}{\sqrt{R^2 + \left(\omega L - \frac{1}{\omega C}\right)^2}} e^{j\left(\omega t - \tan^{-1}\left(\frac{\omega L - \frac{1}{\omega C}}{R}\right)\right)}$$

$$\phi = \tan^{-1}\left(\frac{\omega L - \frac{1}{\omega C}}{R}\right)$$

✓ If  $\phi > 0$  : inductive ckt      $\omega L > \frac{1}{\omega C} \Rightarrow \omega > \frac{1}{\sqrt{LC}}$

$\Rightarrow$  current lags emf

✓ If  $\phi < 0$  : capacitive ckt      $\omega L < \frac{1}{\omega C} \Rightarrow \omega < \frac{1}{\sqrt{LC}}$

$\Rightarrow$  current leads emf

✓ If  $\phi = 0$  :  $\omega L = \frac{1}{\omega C} \Rightarrow \omega = \frac{1}{\sqrt{LC}} = \omega_0$

$\Rightarrow$  resonance : resistive ckt      $\phi = 0$   
in phase current.

Net Reactive Component = 0

$Z = Z_{min} = R$       $\Rightarrow$  I is maximum

$\Rightarrow$  Power delivered is maximum  
where frequency of source matches  
with frequency of circuit.



✓ It is typically used in tuning of - radio / transistor.  
By adjusting L and C, we are trying to match frequency of circuit with frequency of source.

### Power Factor

✓ Cosine of  $\phi$  =  $\cos \phi$  : Power Factor

✓  $V = V_0 \sin \omega t$   
 $I = I_0 \sin (\omega t \pm \phi)$

$$\begin{aligned} P &= VI \\ &= V_0 I_0 \sin(\omega t) \sin(\omega t - \phi) \\ &= V_0 I_0 [\sin \omega t] [\sin \omega t \cos \phi - \cos \omega t \sin \phi] \\ &= V_0 I_0 \sin^2 \omega t \cos \phi - V_0 I_0 \sin \omega t \cos \omega t \sin \phi \end{aligned}$$

Avg. Power

$$= \frac{1}{2} V_0 I_0 \cos \phi$$

$$P_{\text{avg}} = V_{\text{rms}} I_{\text{rms}} \cos \phi$$

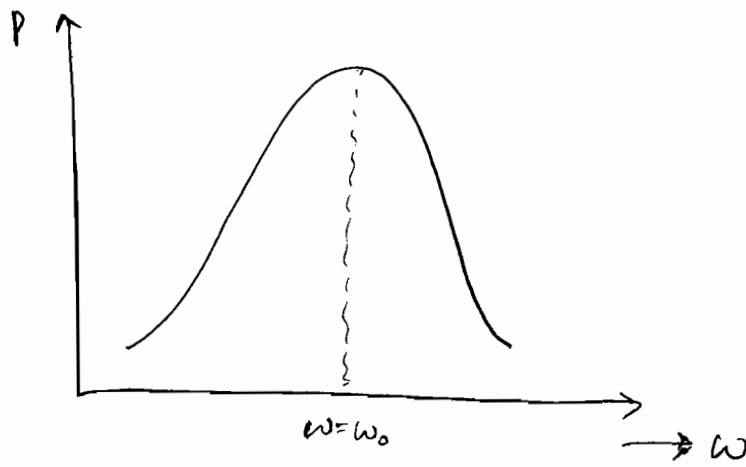
### Quality Factor

In resonance  $\phi = 0$ ,  $\cos \phi = 1$

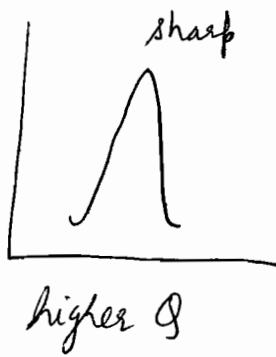
$$P_{\text{avg}} = V_{\text{rms}} I_{\text{rms}} = \frac{1}{2} (V_0) \left( \frac{V_0}{Z} \right) \left( \frac{R}{Z} \right) = \frac{V_0^2 R}{2Z^2}$$

⇒  $P_{\text{avg}} = 0$  for purely ~~resistive~~ inductive or capacitive ckt.

At resonance,  $P_{\text{avg}} = \frac{V_0^2}{2R}$



Quality factor is measurement of sharpness of resonance.



$$Q = \left( \frac{\omega_0}{\text{Bandwidth}} \right) = \left( \frac{\omega_0}{\omega_2 - \omega_1} \right)$$

where  $\omega_1, \omega_2$  : half power frequencies.

$$\frac{V_0^2 R}{2 \left[ R^2 + \left( \omega L - \frac{1}{\omega C} \right)^2 \right]} = \frac{1}{2} \frac{V_0^2}{2R}$$

$$\left( \omega L - \frac{1}{\omega C} \right)^2 = R^2$$

$$\Rightarrow \omega L - \frac{1}{\omega C} = \pm R = 0$$

$$\Rightarrow \omega^2 LC \pm R\omega C - 1 = 0 \Rightarrow \omega = \frac{\pm R C \pm \sqrt{R^2 C^2 + 4LC}}{2LC}$$

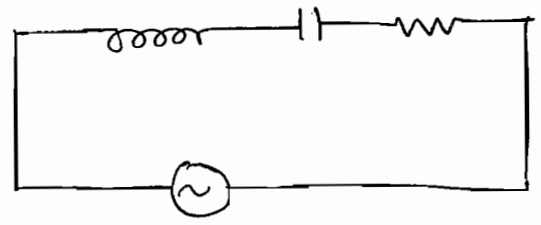
$$Q = \frac{\omega_0 L}{R} = \frac{1}{\omega_0 RC}$$

$$\Rightarrow \omega_1 = \frac{\sqrt{R^2 C^2 + 4LC} - RC}{2LC}$$

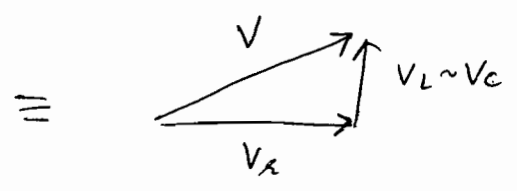
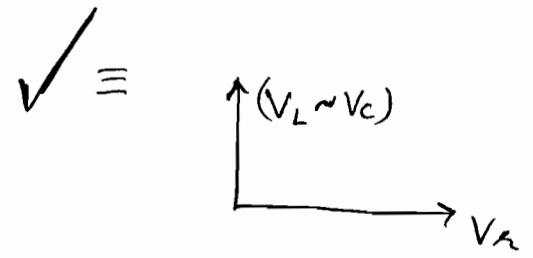
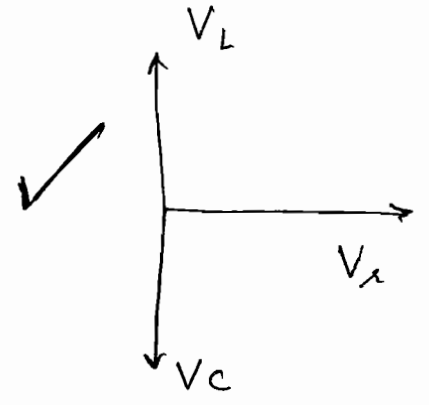
$$\omega_2 = \frac{\sqrt{R^2 C^2 + 4LC} + RC}{2LC}$$

$$\Rightarrow \omega_2 - \omega_1 = \frac{2RC}{2LC} = \left(\frac{R}{L}\right)$$

$$\Rightarrow Q = \frac{1}{\sqrt{LC}} \frac{2L}{R} = \frac{1}{R} \sqrt{\frac{L}{C}}$$



- $V_L = I X_L = j I \omega L$
- $V_C = I X_C = \frac{I}{j \omega C} = -j \frac{I}{\omega C}$
- $V_R = I R$



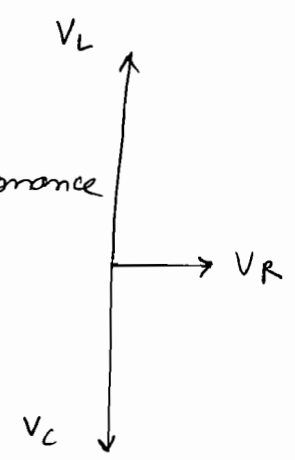
$$V = \sqrt{(V_L \sim V_C)^2 + V_R^2}$$

### Quality Factor

def<sup>1</sup>: sharpness of resonance

def<sup>2</sup>: voltage magnifying power at resonance

ie.  $Q = \frac{V_L \text{ (or } V_C)}{V_{\text{applied}}}$  at resonance



$$= \frac{I \omega_0 L}{IR}$$

$$= \left( \frac{\omega_0 L}{R} \right) = \underline{\underline{\frac{1}{\omega_0 CR}}}$$

From def<sup>1</sup>:  $\omega_2 - \omega_1 = (R/L) = \text{Bandwidth}$

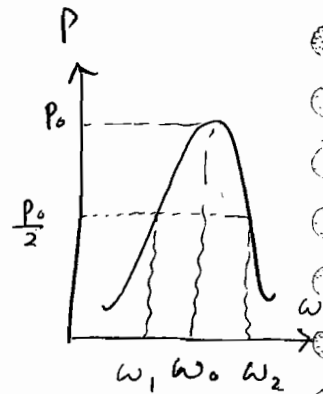
### Half Power Frequencies

$$\omega_0 - \omega_1 = \frac{R}{2L}$$

$$\Rightarrow \omega_1 = \left( \omega_0 - \frac{R}{2L} \right)$$

$$\omega_2 - \omega_0 = \frac{R}{2L}$$

$$\Rightarrow \omega_2 = \left( \omega_0 + \frac{R}{2L} \right)$$



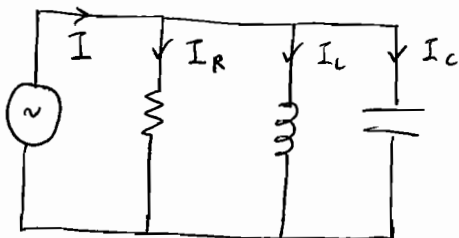
For any complex circuit, to find  $\omega_0$

Find  $Z_{net}$

Write  $Z_{net}$  as  $Z_+ + jZ_-$

Put  $Z_- = 0$  and find  $\omega_0$ .

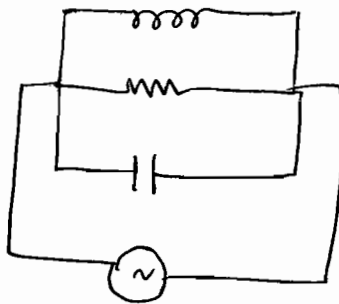
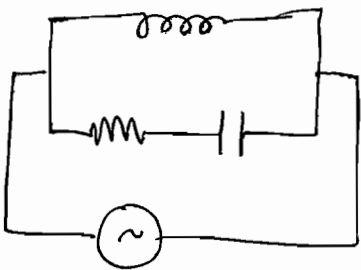
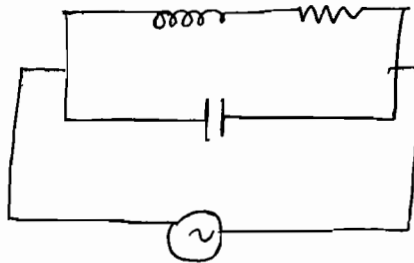
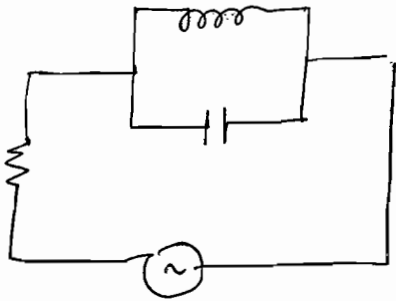
### Parallel Resonant LCR,



Rejector Circuit  
Parallel LCR Circuit

⊛ Rejector Ckt means Current at Resonance becomes minimum while  $I_C$  and  $I_L$  becomes maximum and opposite. Hence used in filtering circuit. Use to filter out resonant frequency.

Note that parallel ckt implies only L and C are in parallel. R can be in series or parallel. But all the results are derived for the latter case.



✓ We are interested in this

$Y = \frac{1}{Z}$  is called Admittance ....

$$I = \frac{V}{Z} = VY$$

$$\frac{1}{Z} = \frac{1}{Z_1} + \frac{1}{Z_2} + \frac{1}{Z_3} \Rightarrow \frac{V}{Z} = \frac{V}{Z_1} + \frac{V}{Z_2} + \frac{V}{Z_3}$$

$$\Rightarrow I = I_1 + I_2 + I_3$$

Y is written as  $g + jb$   
 ↑ conductance      susceptance

At resonant condition, set  $b = 0$

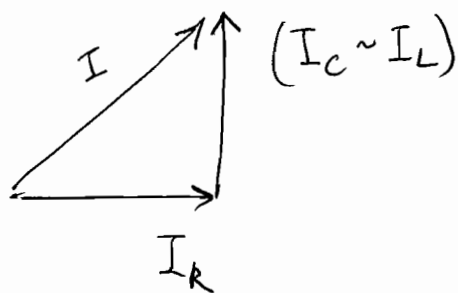
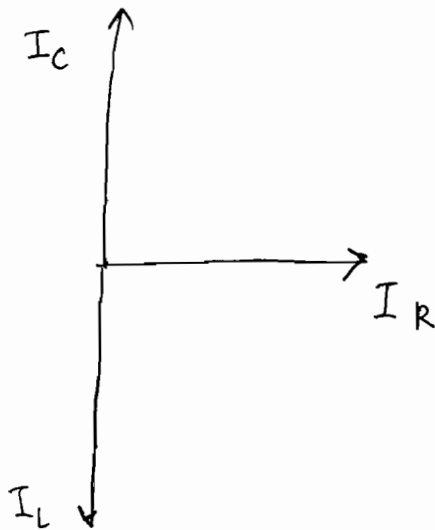
i.e. maximum impedance i.e. minimum current. This parallel ckt. does not allow the resonant frequency from source to pass in the circuit. Therefore, called rejector ckt. for such a freq.

↑↑  
 i.e. Admittance is minimum at resonance

$$I_L = \frac{V}{j\omega L} = \frac{V_0}{\omega L} e^{j(\omega t - \pi/2)}$$

$$I_C = Vj\omega C = V_0\omega C e^{j(\omega t + \pi/2)}$$

$$I_R = \frac{V}{R} = \frac{V_0}{R} e^{j\omega t}$$



$$I = \sqrt{I_R^2 + (I_C \sim I_L)^2}$$

At resonance,  $I_C = I_L \Rightarrow I = I_R$  : minimum

Also  $I_C, I_L \gg I_R$  at resonance

Hence, Quality Factor =  $\left[ \frac{I_{\text{applied}}}{I_L \text{ (or } I_C)} \right]_{\text{at resonance}}$  : Current Magnifying Power

$$I_L = \frac{V_0}{\omega_0 L} = I_C = V_0 \omega_0 C \quad (\text{at resonance})$$

$$\Rightarrow Q = \frac{V_0/R}{V_0/\omega_0 L} = \left( \frac{\omega_0 L}{R} \right) = \frac{1}{\omega_0 CR}$$

✓ Hence same Q, same BW, same Power Factor

$$\frac{1}{Z} = \frac{1}{j\omega L} + j\omega C + \frac{1}{R} = \frac{1}{R} + j \left( \omega C - \frac{1}{\omega L} \right)$$

$$= g + jb$$

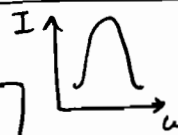
set  $b=0$  for resonance

✓ This is because, synchronization is required for resonance.  
 For synchronization phase should be same. No lead or lag. Hence put complex component which is responsible for phase = 0

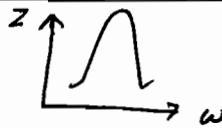
$$\Rightarrow \omega_0 C = \frac{1}{\omega_0 L} \Rightarrow \boxed{\omega_0 = \frac{1}{\sqrt{LC}}}$$

Bandwidth

① Series Ckt



② Parallel Ckt



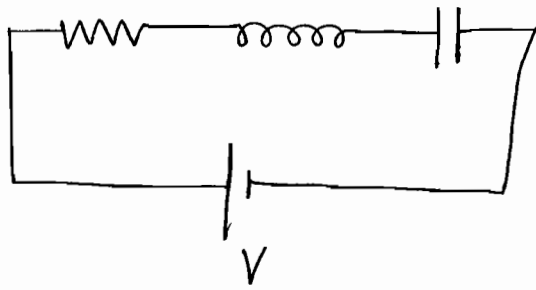
$$I = \frac{I_0}{\sqrt{2}}$$

$$Z = \left( \frac{Z_0}{\sqrt{2}} \right)$$

Power in AC ckt

$$\text{True Power} = \text{Apparent Power} * \text{Power Factor}$$

# Transient Response in LCR



→ Note that using algebraically solution is simple here. In AC it was very difficult, hence we resorted to Phasor Analysis.

$$V = V_L + V_R + V_C$$

$$V = L \left( \frac{dI}{dt} \right) + IR + \frac{q}{C}$$

→ Charge will grow across C and I will flow (increase) in L

$$\Rightarrow \frac{d^2 I}{dt^2} + \frac{R}{L} \left( \frac{dI}{dt} \right) + \frac{I}{LC} = 0$$

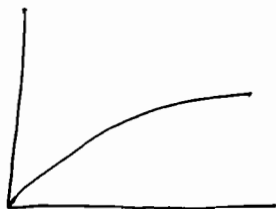
$$\Rightarrow \frac{d^2 I}{dt^2} + 2c \left( \frac{dI}{dt} \right) + \omega_0^2 = 0$$

$$2c = \left( \frac{R}{L} \right)$$

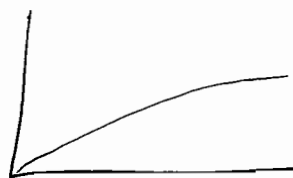
$$\omega_0^2 = \left( \frac{1}{LC} \right)$$

Typical, control equation.

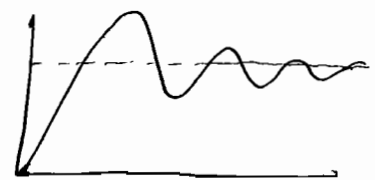
We know the solution for 3 cases.



Overdamped  
( $c > \omega_0$ )



Critically  
damped ( $c = \omega_0$ )



$c < \omega_0$   
Underdamped



$$C > \omega_0$$

$$\underline{R > R_0}$$

$$C = \omega_0$$

$$\frac{R}{2L} = \frac{1}{\sqrt{LC}}$$

$$\underline{R = R_0}$$

$$C < \omega_0$$

$$\Rightarrow \frac{R}{2L} < \frac{1}{\sqrt{LC}}$$

$$\Rightarrow R < 2\sqrt{\frac{L}{C}}$$

$$\text{ie. } \underline{R < R_0}$$

$$R_0 = 2\sqrt{\frac{L}{C}}$$

$$\checkmark I = I_0 e^{-\left(\frac{R}{2L}\right)t} \sin(\sqrt{\omega_0^2 - C^2}t + \phi)$$

$$\checkmark \omega_d = \sqrt{\frac{1}{LC} - \left(\frac{R}{2L}\right)^2}$$

✓ Oscillatory Circuit means that there are 2 sources of energy stored. Hence, the 2 exchange energy thereby giving rise to Oscillations.

Note that the equation written in current can be written for charge also. Hence again 3 solutions.

$$L \frac{d^2 q}{dt^2} + \frac{dq}{dt} R + \frac{q}{C} = V$$

$$\Rightarrow \frac{d^2 q}{dt^2} + \left(\frac{R}{L}\right) \frac{dq}{dt} + \left(\frac{1}{LC}\right) q = \left(\frac{V}{L}\right)$$

solution :

~~(solution of L.H.S)~~  
(solution of L.H.S) + VC

Put  $(q - VC) = q'$   
Now we find solution for  $q'$   
 $\checkmark \Rightarrow q = q' + VC$

2011

$$R_0 = 2 \sqrt{\frac{10 \times 10^3}{0.1}} = 200\sqrt{10}$$

or

$$\omega = \frac{1}{\sqrt{LC}} > \gamma < C \quad \text{for oscillations}$$

$$\frac{1}{\sqrt{LC}} > \frac{R}{2L}$$

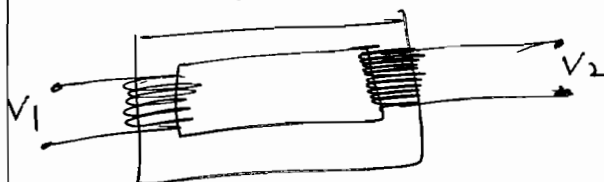
$$\text{i.e. } R < 2\sqrt{\frac{L}{C}}$$

$$200 < 200\sqrt{10}$$

Hence oscillatory

$$\omega = \sqrt{\frac{1}{LC} - \left(\frac{R}{2L}\right)^2}$$

### Principle of Transformer



$$\Phi_1 = \Phi_2$$

$$\Rightarrow \left(\frac{d\Phi_1}{dt}\right) = \left(\frac{d\Phi_2}{dt}\right)$$

$$\Rightarrow \frac{N_1}{N_1} \left(\frac{d\Phi_1}{dt}\right) = \frac{N_2}{N_2} \left(\frac{d\Phi_2}{dt}\right)$$

$$\Rightarrow \frac{E_1}{N_1} = \frac{E_2}{N_2}$$

$$\Rightarrow E_2 = E_1 \left(\frac{N_2}{N_1}\right)$$

Also

$$P_1 = P_2$$

$$\Rightarrow E_1 i_1 = E_2 i_2$$

$$\Rightarrow i_2 = i_1 \left(\frac{E_1}{E_2}\right)$$

if  $N_2 > N_1$  coils in secondary  $\rightarrow$  step up!!

2011

$$\text{Quality Factor} = \frac{2\pi (\text{Energy stored})}{(\text{Energy loss per cycle})}$$

$$q = q_0 e^{-ct} \sin(\omega dt + \theta)$$

$$\text{Energy} \propto \frac{q^2}{2C}$$

$$E = \left[\frac{q_0^2}{2C}\right] e^{-2ct}$$

$$\left(\frac{dE}{dt}\right) = -2cE.$$

$$Q = 2\pi \frac{E}{2cE * \tau} = \left(\frac{\omega}{2c}\right)$$

✓ Transmission Coefficient  $t = \frac{E_t}{E_i}$

while

$$\text{Transmittivity} = T$$

✓ Attempt BIG QUESTIONS. Attempt those with lesser sub parts.

Q10/Tut 7

$$\frac{1}{Z} = \frac{1}{Z_1} + \frac{1}{Z_2}$$

$$= \frac{1}{R_1 + j\omega L} + \frac{1}{R_2 - \frac{j}{\omega C}}$$

$$= \frac{R_1 - j\omega L}{R_1^2 + \omega^2 L^2} + \frac{R_2 + j/\omega C}{R_2^2 + \frac{1}{\omega^2 C^2}}$$

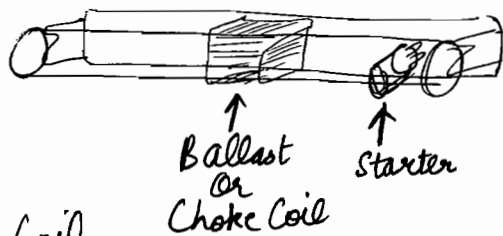
$$= \left( \frac{R_1}{R_1^2 + \omega^2 L^2} + \frac{R_2}{R_2^2 + \frac{1}{\omega^2 C^2}} \right) - j \left( \frac{\omega L}{R_1^2 + \omega^2 L^2} - \frac{1/\omega C}{R_2^2 + \frac{1}{\omega^2 C^2}} \right)$$

$$= g + jb$$

Put  $b=0$

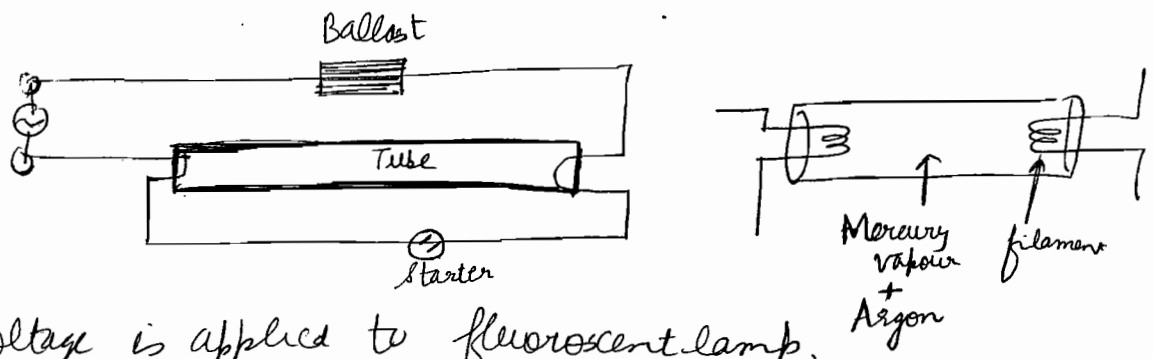
## Use of choke coil in tubes

- ① Tubes are technically called Fluorescent lamps. Small ones are called CFL or Compact Fluorescent lamps.
- ② A fluorescent lamp is a gas discharge lamp that uses electricity to excite mercury vapour. The excited mercury vapour atoms produce short-wave UV light. UV light causes a phosphor to fluoresce, producing visible light.
- ③ A fluorescent lamp converts electrical power into useful light more efficiently than an incandescent lamp (household bulb).
- ④ Fluorescent lamps are negative differential resistance devices, so as more current flows through them, the electrical resistance of the fluorescent lamp drops, allowing even more current to flow. Connected to a constant voltage power supply, a fluorescent lamp will self-destruct due to uncontrolled current flow. To prevent this, an auxiliary device called BALLAST is used. Simplest ~~Ballast~~ device for AC use is an inductor placed in series, consisting of a winding on a laminated core. The inductance of this winding limits the flow of AC current. Ballast Coil is also called



Choke Coil

- ① Common Fluorescent Ballast also used an additional device called starter. The starter is there to light the fluorescent lamp initially.



When voltage is applied to fluorescent lamp, here's what happens:

- ① starter (which is simply a timed switch) allows current to flow through the filaments at the ends of the tube
- ② the current causes the starter's contacts to heat up & open, thus interrupting the flow of current. The tube lights
- ③ since the lighted fluorescent tube has low resistance, the ballast now serves as a current limiter.

When you turn on fluorescent tube, the starter is a closed switch. The filament at the ends of the tube are heated by electricity, and they create a cloud of electrons inside the tube. The fluorescent starter is a time delay switch that opens after a second or two. When it opens, the voltage across the tube allows the stream of electrons to flow across the tube and ionize the mercury vapour.

Without the starter, a steady stream of electrons is never created (as not enough  $e^-$  are produced due to low heating) between the two filaments, and the lamp flickers.

Without the ballast or choke coil, the circuit is a short circuit b/w the filaments, and this short ckt. contains a lot of current. The current either vaporizes the filaments or causes the lamp to explode. Note that  $L$  is high but  $\cos \phi$  is low  $\Rightarrow$  low current and low wastage of power.

For a series LCR :

$$P_{\text{delivered}} = V_{\text{rms}} I_{\text{rms}} \cos \phi = \frac{V_0}{\sqrt{2}} \cdot \frac{I_0}{\sqrt{2}} \cos \phi$$

$$= \frac{V_0}{2} \cdot \frac{V_0}{Z} \cdot \frac{R}{Z} = \frac{\left( \frac{V_0^2 R}{2 Z^2} \right)}$$

$$P_{\text{consumed}} = I_{\text{rms}}^2 R = \frac{I_0^2 R}{2} = \left( \frac{V_0^2 R}{2 Z^2} \right)$$

@ Resonance & Half Power Points

$$P = \frac{P_0}{2} = \left( \frac{V_0^2 R}{4 Z^2} \right)$$

$$@ I = \frac{I_0}{\sqrt{2}}, \quad P = \left( \frac{V_0}{\sqrt{2}} \right)^2 \frac{R}{2} = \frac{I_0^2 R}{4} = \left( \frac{V_0^2 R}{4 Z^2} \right)$$

Resonance पर हो न हो, जो power deliver होगी, उसे 'R' ही खासगा न !!

For a parallel LCR :

$$@ \text{Resonance : } Z = Z_{\text{min}} \\ Y = Y_{\text{max}}$$

$$@ \text{Bandwidth points } Z = \left( \frac{Z_0}{\sqrt{2}} \right)$$

$$Y = \sqrt{2} Y_0$$

$$\vec{\nabla} \cdot \vec{E} = 0 \quad \vec{\nabla} \cdot \vec{B} = 0 \quad \vec{\nabla} \times \vec{E} = - \left( \frac{\partial \vec{B}}{\partial t} \right) \quad \vec{\nabla} \times \vec{B} = \mu_0 \sigma \vec{E} + \mu_0 \epsilon_0 \left( \frac{\partial \vec{E}}{\partial t} \right)$$

$$\vec{\nabla} \times (\vec{\nabla} \times \vec{E}) = \vec{\nabla} (\vec{\nabla} \cdot \vec{E}) - \vec{\nabla}^2 \vec{E} = + \frac{\partial}{\partial t} \left( \mu_0 \sigma \vec{E} + \mu_0 \epsilon_0 \frac{\partial \vec{E}}{\partial t} \right)$$

$$\boxed{\vec{\nabla}^2 \vec{E} = \mu_0 \sigma \left( \frac{\partial \vec{E}}{\partial t} \right) + \mu_0 \epsilon_0 \left( \frac{\partial^2 \vec{E}}{\partial t^2} \right)}$$